

24-Bit, 96-kHz/192-kHz, 6-In/8-Out Audio Codec with Differential Input/Output

FEATURES

- **24-BIT $\Delta\Sigma$ ADC AND DAC**
- **SIX-CHANNEL ADC:**
 - High Performance: Differential and Single-Ended, $f_s = 48$ kHz
 - THD+N: –93 dB (Differential), –93 dB (Single-Ended)
 - SNR: 107 dB (Differential), 104 dB (Single-Ended)
 - Dynamic Range: 107 dB (Differential), 104 dB (Single-Ended)
 - Sampling Rate: 8 kHz to 96 kHz
 - System Clock: 256 f_s , 384 f_s , 512 f_s , 768 f_s
 - Differential Voltage Input: 2 V_{RMS}
 - Single-Ended Voltage Input: 1 V_{RMS}
 - Decimation Filter:
 - Passband Ripple: ± 0.035 dB
 - Stop Band Attenuation: –75 dB
 - On-Chip, High-Pass Filter: 0.96 Hz at $f_s = 48$ kHz
 - Overflow Flag
- **EIGHT-CHANNEL DAC:**
 - High Performance: Differential, $f_s = 48$ kHz
 - THD+N: –94 dB
 - SNR: 112 dB
 - Dynamic Range: 112 dB
 - Sampling Rate: 8 kHz to 192 kHz
 - System Clock: 128 f_s , 192 f_s , 256 f_s , 384 f_s , 512 f_s , 768 f_s
 - Differential Voltage Output: 8 V_{PP}
 - Analog Low-Pass Filter Included
 - 4x/8x Oversampling Digital Filter:
 - Passband Ripple: ± 0.0018 dB
 - Stop Band Attenuation: –75 dB
 - Zero Flag
- **FLEXIBLE MODE CONTROL:**
 - Four-Wire SPI™, Two-Wire I²C™ Compatible Serial Control Interface or Hardware Control
- **MULTIPLE FUNCTIONS VIA SPI OR I²C I/F:**
 - Audio I/F Mode/Format Select for ADC and DAC
 - Digital Attenuation and Soft Mute for ADC and DAC
 - Digital De-Emphasis: 32 kHz, 44.1 kHz, 48 kHz for DAC
 - Data Polarity Control for ADC and DAC
 - Power Down ADC/DAC Independently
- **MULTI FUNCTIONS VIA H/W CONTROL:**
 - Audio I/F Mode/Format Select
 - Digital De-Emphasis Filter: 44.1 kHz for DAC
- **EXTERNAL RESET PIN:**
 - ADC/DAC Simultaneous
- **AUDIO INTERFACE MODE:**
 - ADC/DAC Independent Master/Slave
- **AUDIO DATA FORMAT:**
 - ADC/DAC Independent I²S™, Left-Justified, Right-Justified, DSP, TDM
- **POWER SUPPLIES:** 5 V for Analog and 3.3 V for Digital
- **PACKAGE:** HTQFP-64
- **OPERATING TEMPERATURE RANGE:**
 - Consumer Grade: –40°C to +85°C
 - Automotive Audio Grade: –40°C to +105°C

APPLICATIONS

- CAR AUDIO EXTERNAL AMPLIFIERS
- CAR AUDIO AVN APPLICATIONS
- HOME THEATERS
- AV RECEIVERS



Please be aware that an important notice concerning availability, standard warranty, and use in critical applications of Texas Instruments semiconductor products and disclaimers thereto appears at the end of this data sheet.

PowerPAD is a trademark of Texas Instruments Incorporated.

SPI is a trademark of Motorola.

I²C, I²S are trademarks of NXP Semiconductors.

All other trademarks are the property of their respective owners.

DESCRIPTION

The PCM3168A and PCM3168A-Q1 are high-performance, single-chip, 24-bit, 6-in/8-out, audio coders/decoders (codecs) with single-ended and differential selectable analog inputs and differential outputs. The six-channel, 24-bit analog-to-digital converter (ADC) employs a delta-sigma ($\Delta\Sigma$) modulator and supports 8-kHz to 96-kHz sampling rates and a 16-bit/24-bit width digital audio output word on the audio interface. The eight-channel, 24-bit digital-to-analog converter (DAC) employs a $\Delta\Sigma$ modulator and supports 8-kHz to 192-kHz sampling rates and a 16-bit/24-bit width digital audio input word on the audio interface. Each audio interface supports I²S, left-justified, right-justified, and DSP formats with 16-bit/24-bit word width. In addition, the PCM3168A and PCM3168A-Q1 support the time-division-multiplexed (TDM) format.

The PCM3168A and PCM3168A-Q1 can be controlled through a four-wire, SPI-compatible interface, or two-wire, I²C-compatible serial interface in software, which provides access to all functions including digital attenuation, soft mute, de-emphasis, and so forth. Also, hardware control mode provides a subset of user-programmable functions through four control pins. The PCM3168A and PCM3168A-Q1 are available in a 12-mm × 12-mm (10-mm × 10-mm body) HTQFP-64 PowerPAD™ package.



This integrated circuit can be damaged by ESD. Texas Instruments recommends that all integrated circuits be handled with appropriate precautions. Failure to observe proper handling and installation procedures can cause damage.

ESD damage can range from subtle performance degradation to complete device failure. Precision integrated circuits may be more susceptible to damage because very small parametric changes could cause the device not to meet its published specifications.

ORDERING INFORMATION

For the most current package and ordering information see the Package Option Addendum at the end of this document, or see the TI web site at www.ti.com.

ABSOLUTE MAXIMUM RATINGS⁽¹⁾

Over operating free-air temperature range (unless otherwise noted).

| | PCM3168A, PCM3168A-Q1 | UNIT |
|--|----------------------------|------|
| Supply voltage: VCCAD1, VCCAD2, VCCDA1, VCCDA2 | −0.3 to +6.5 | V |
| Supply voltage: VDD1, VDD2 | −0.3 to +4.0 | V |
| Ground voltage differences: AGNDAD1, AGNDAD2, AGNDDA1, AGNDDA2, DGND1, DGND2 | ±0.1 | V |
| Supply voltage differences: VCCAD1, VCCAD2, VCCDA1, VCCDA2 | ±0.1 | V |
| Supply voltage differences: VDD1, VDD2 | ±0.1 | V |
| Digital input voltage: RST, MS, MC, MDI, SCK | −0.3 to +6.5 | V |
| Digital input voltage: BCKAD/DA, LRCKAD/DA, DIN1/2/3/4, DOUT1/2/3, MODE, OVF, ZERO, MDO | −0.3 to (VDD + 0.3) < +4.0 | V |
| Analog input voltage: VIN1-6±, VCOMAD/DA, VOUT1-8±, VREFAD1/2 | −0.3 to (VCC + 0.3) < +6.5 | V |
| Input current (all pins except supplies) | ±10 | mA |
| Ambient temperature range (under bias) | −40 to +125 | °C |
| Storage temperature | −55 to +150 | °C |
| Junction temperature | +150 | °C |
| Lead temperature (soldering, 5s) | +260 | °C |
| Package temperature (IR reflow, peak) | +260 | °C |

(1) Stresses beyond those listed under absolute maximum ratings may cause permanent damage to the device. These are stress ratings only and functional operation of the device at these or any other conditions beyond those indicated under recommended operating conditions is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

RECOMMENDED OPERATING CONDITIONS

Over operating free-air temperature range (unless otherwise noted).

| PARAMETER | PCM3168A, PCM3168A-Q1 | | | UNIT | |
|---------------------------------|---|-------|-----------------------|------------------|----|
| | MIN | TYP | MAX | | |
| Analog supply voltage, VCC | 4.5 | 5.0 | 5.5 | V | |
| Digital supply voltage, VDD | 3.0 | 3.3 | 3.6 | V | |
| Digital Interface | LVTTTL compatible | | | | |
| Digital input clock frequency | Sampling frequency, LRCKAD/LRCKDA ⁽¹⁾ | 8 | 96/192 ⁽¹⁾ | kHz | |
| | System clock frequency, SCKI | 2.048 | 36.864 | MHz | |
| Analog input level | Single-ended | | 1 | V _{RMS} | |
| | Differential | | 2 | V _{RMS} | |
| Analog output voltage | Differential | | 8 | V _{PP} | |
| Analog output load resistance | To ac-coupled GND | 5 | | kΩ | |
| | To dc-coupled GND | 15 | | kΩ | |
| Analog output load capacitance | | | 50 | pF | |
| Digital output load capacitance | | | 20 | pF | |
| Operating free-air temperature | PCM3168A Consumer grade | −40 | +25 | +85 | °C |
| | PCM3168A-Q1 Automotive audio grade | −40 | +25 | +105 | °C |

(1) 192 kHz is supported only for DAC.

ELECTRICAL CHARACTERISTICS: Digital Input/Output

All specifications at $T_A = +25^\circ\text{C}$, $V_{CCAD1} = V_{CCAD2} = V_{CCDA1} = V_{CCDA2} = 5\text{ V}$, $V_{DD1} = V_{DD2} = 3.3\text{ V}$, $f_S = 48\text{ kHz}$, $SCKI = 512\text{ f}_S$, 24-bit data, Sampling Mode = Auto for ADC and DAC, and Interface Mode = Slave for ADC and DAC, unless otherwise noted.

| PARAMETER | | TEST CONDITIONS | PCM3168A, PCM3168A-Q1 | | | UNIT |
|---|-----------------------------------|--|------------------------------------|--------------|--------|------|
| | | | MIN | TYP | MAX | |
| DATA FORMAT | | | | | | |
| Audio data interface format | | | I ² S, LJ, RJ, DSP, TDM | | | |
| Audio data word length | | | | 16, 24 | | Bits |
| Audio data format | | | MSB first, twos complement | | | |
| Sampling frequency, ADC | f _S | | 8 | 48 | 96 | kHz |
| Sampling frequency, DAC | | | 8 | 48 | 192 | kHz |
| System clock frequency | | 128 f _S , 192 f _S , 256 f _S , 384 f _S , 512 f _S , 768 f _S | 2.048 | | 36.864 | MHz |
| INPUT LOGIC | | | | | | |
| Input logic level | V _{IH} ⁽¹⁾⁽²⁾ | | 2.0 | | VDD | VDC |
| | V _{IL} ⁽¹⁾⁽²⁾ | | | | 0.8 | VDC |
| Input logic level | V _{IH} ⁽³⁾⁽⁴⁾ | | 2.0 | | 5.5 | VDC |
| | V _{IL} ⁽³⁾⁽⁴⁾ | | | | 0.8 | VDC |
| Input logic level | I _{IH} ⁽²⁾⁽³⁾ | V _{IN} = VDD | | | ±10 | μA |
| | I _{IL} ⁽²⁾⁽³⁾ | V _{IN} = 0 V | | | ±10 | μA |
| Input logic level | I _{IH} ⁽¹⁾⁽⁴⁾ | V _{IN} = VDD | | +65 | +100 | μA |
| | I _{IL} ⁽¹⁾⁽⁴⁾ | V _{IN} = 0 V | | | ±10 | μA |
| OUTPUT LOGIC | | | | | | |
| Output logic level | V _{OH} ⁽⁵⁾ | I _{OUT} = -4 mA | 2.4 | | | VDC |
| | V _{OL} ⁽⁵⁾⁽⁶⁾ | I _{OUT} = +4 mA | | | 0.4 | VDC |
| REFERENCE INPUT/OUTPUT | | | | | | |
| VREFAD1 output voltage | | | | VCCAD1 | | V |
| VREFAD2 output voltage | | | | AGNDAD1 | | V |
| VCOMAD output voltage | | | | 0.5 × VCCAD1 | | V |
| VCOMAD output impedance | | | | 10 | | kΩ |
| Allowable VCOMAD output source/sink current | | | | | 1 | μA |
| VCOMDA output voltage | | | | 0.5 × VCCDA1 | | V |
| VCOMDA output impedance | | | | 7.5 | | kΩ |
| Allowable VCOMDA output source/sink current | | | | | 1 | μA |

- (1) BCKAD, BCKDA, LRCKAD, and LRCKDA (in slave mode, Schmitt trigger input with 50-kΩ typical internal pull-down resistor).
- (2) DIN1/2/3/4 and MDO/ADR1/MD1. (Except SPI mode, Schmitt trigger input).
- (3) SCKI, MDI/SDA/DEMP, and MC/SCL/FMT (Schmitt trigger input, 5-V tolerant).
- (4) RST and MS/ADR0/MD0 (Schmitt trigger input with 50-kΩ typical internal pull-down resistor, 5-V tolerant).
- (5) BCKAD, BCKDA, LRCKAD, and LRCKDA (in master mode), DOUT1/2/3, ZERO, OVF, and MDO/ADR1/MD1 (in SPI mode).
- (6) SDA (in I²C mode, open-drain low output).

ELECTRICAL CHARACTERISTICS: ADC Characteristics

All specifications at $T_A = +25^\circ\text{C}$, $V_{\text{CAD1}} = V_{\text{CAD2}} = V_{\text{CDA1}} = V_{\text{CDA2}} = 5\text{ V}$, $V_{\text{DD1}} = V_{\text{DD2}} = 3.3\text{ V}$, $f_S = 48\text{ kHz}$, $\text{SCKI} = 512 f_S$, 24-bit data, Sampling Mode = Auto for ADC and DAC, and Interface Mode = Slave for ADC and DAC, unless otherwise noted.

| PARAMETER | TEST CONDITIONS | PCM3168A, PCM3168A-Q1 | | | UNIT |
|--|--|-----------------------|------------------------------|---------|------------------|
| | | MIN | TYP | MAX | |
| ADC CHARACTERISTICS | | | | | |
| Resolution | | 16 | 24 | | Bits |
| Full-scale input voltage | $V_{\text{IN}} = 0\text{ dB}$, Single-ended | | $0.2 \times V_{\text{CAD1}}$ | | V_{RMS} |
| | $V_{\text{IN}} = 0\text{ dB}$, Differential | | $0.4 \times V_{\text{CAD1}}$ | | V_{RMS} |
| Center voltage | | | $0.5 \times V_{\text{CAD1}}$ | | V |
| Input impedance | | | 45 | | k Ω |
| Common-mode rejection ratio | | | 80 | | dB |
| DC ACCURACY | | | | | |
| Gain mismatch channel-to-channel | Full-scale input, V_{IN} | | ± 2.0 | ± 6 | % of FSR |
| Gain error | Full-scale input, V_{IN} | | ± 2.0 | ± 6 | % of FSR |
| Bipolar zero error | High-pass filter bypass, V_{IN} | | ± 1.0 | | % of FSR |
| DYNAMIC PERFORMANCE⁽¹⁾⁽²⁾ | | | | | |
| THD+N, $V_{\text{IN}} = -1\text{ dB}$ | $f_S = 48\text{ kHz}$, Differential | | -93 | -87 | dB |
| | $f_S = 96\text{ kHz}$, Differential | | -93 | | dB |
| | $f_S = 48\text{ kHz}$, Single-ended | | -93 | | dB |
| | $f_S = 96\text{ kHz}$, Single-ended | | -93 | | dB |
| Dynamic range | $f_S = 48\text{ kHz}$, A-weighted, differential | 100 | 107 | | dB |
| | $f_S = 96\text{ kHz}$, A-weighted, differential | | 107 | | dB |
| | $f_S = 48\text{ kHz}$, A-weighted, single-ended | | 104 | | dB |
| | $f_S = 96\text{ kHz}$, A-weighted, single-ended | | 104 | | dB |
| S/N ratio | $f_S = 48\text{ kHz}$, A-weighted, differential | 100 | 107 | | dB |
| | $f_S = 96\text{ kHz}$, A-weighted, differential | | 107 | | dB |
| | $f_S = 48\text{ kHz}$, A-weighted, single-ended | | 104 | | dB |
| | $f_S = 96\text{ kHz}$, A-weighted, single-ended | | 104 | | dB |
| Channel separation (between one channel and others) | $f_S = 48\text{ kHz}$, Differential | 98 | 104 | | dB |
| | $f_S = 96\text{ kHz}$, Differential | | 104 | | dB |
| | $f_S = 48\text{ kHz}$, Single-ended | | 101 | | dB |
| | $f_S = 96\text{ kHz}$, Single-ended | | 101 | | dB |

(1) In differential mode at $V_{\text{IN}x\pm}$ pin, $f_{\text{IN}} = 1\text{ kHz}$, using Audio Precision System II, RMS mode with 20-kHz low-pass filter and 400-Hz high-pass filter.

(2) $f_S = 48\text{ kHz}$: $\text{SCKI} = 512 f_S$ (single), $f_S = 96\text{ kHz}$: $\text{SCKI} = 256 f_S$ (dual), $f_S = 192\text{ kHz}$: $\text{SCKI} = 128 f_S$ (quad).

ELECTRICAL CHARACTERISTICS: ADC Characteristics (continued)

All specifications at $T_A = +25^\circ\text{C}$, $V_{CCAD1} = V_{CCAD2} = V_{CCDA1} = V_{CCDA2} = 5\text{ V}$, $V_{DD1} = V_{DD2} = 3.3\text{ V}$, $f_S = 48\text{ kHz}$, $SCKI = 512 f_S$, 24-bit data, Sampling Mode = Auto for ADC and DAC, and Interface Mode = Slave for ADC and DAC, unless otherwise noted.

| PARAMETER | TEST CONDITIONS | PCM3168A, PCM3168A-Q1 | | | UNIT |
|-------------------------------------|--|-----------------------|------------------------|--------------------|------|
| | | MIN | TYP | MAX | |
| DIGITAL FILTER PERFORMANCE | | | | | |
| Passband (single) | | | | $0.454 \times f_S$ | Hz |
| Passband (dual) | | | | $0.454 \times f_S$ | Hz |
| Stop band (single) | | $0.555 \times f_S$ | | | Hz |
| Stop band (dual) | | $0.597 \times f_S$ | | | Hz |
| Passband ripple | $< 0.454 \times f_S, 0.454 \times f_S$ | | | ± 0.035 | dB |
| Stop band attenuation | $> 0.555 \times f_S, 0.597 \times f_S$ | -75 | | | dB |
| Group delay time (single) | | | $27/f_S$ | | sec |
| Group delay time (dual) | | | $17/f_S$ | | sec |
| High-pass filter frequency response | -3 dB | | $0.02 \times f_S/1000$ | | Hz |

ELECTRICAL CHARACTERISTICS: DAC Characteristics

All specifications at $T_A = +25^\circ\text{C}$, $V_{CCAD1} = V_{CCAD2} = V_{CCDA1} = V_{CCDA2} = 5\text{ V}$, $V_{DD1} = V_{DD2} = 3.3\text{ V}$, $f_S = 48\text{ kHz}$, $SCKI = 512 f_S$, 24-bit data, Sampling Mode = Auto for ADC and DAC, and Interface Mode = Slave for ADC and DAC, unless otherwise noted.

| PARAMETER | TEST CONDITIONS | PCM3168A, PCM3168A-Q1 | | | UNIT |
|--|---|-----------------------|-----------|---------|----------|
| | | MIN | TYP | MAX | |
| DAC CHARACTERISTICS | | | | | |
| Resolution | | 16 | 24 | | Bits |
| DC ACCURACY | | | | | |
| Gain mismatch channel-to-channel | | | ± 2.0 | ± 6 | % of FSR |
| Gain error | | | ± 2.0 | ± 6 | % of FSR |
| Bipolar zero error | | | ± 1.0 | | % of FSR |
| DYNAMIC PERFORMANCE⁽¹⁾⁽²⁾ | | | | | |
| THD+N, $V_{OUT} = 0\text{ dB}$ | $f_S = 48\text{ kHz}$ | | -94 | -88 | dB |
| | $f_S = 96\text{ kHz}$ | | -94 | | dB |
| | $f_S = 192\text{ kHz}$ | | -94 | | dB |
| Dynamic range | $f_S = 48\text{ kHz}$, EIAJ, A-weighted | 105 | 112 | | dB |
| | $f_S = 96\text{ kHz}$, EIAJ, A-weighted | | 112 | | dB |
| | $f_S = 192\text{ kHz}$, EIAJ, A-weighted | | 112 | | dB |
| S/N ratio | $f_S = 48\text{ kHz}$, EIAJ, A-weighted | 105 | 112 | | dB |
| | $f_S = 96\text{ kHz}$, EIAJ, A-weighted | | 112 | | dB |
| | $f_S = 192\text{ kHz}$, EIAJ, A-weighted | | 112 | | dB |
| Channel separation (between one channel and others) | $f_S = 48\text{ kHz}$ | 102 | 108 | | dB |
| | $f_S = 96\text{ kHz}$ | | 108 | | dB |
| | $f_S = 192\text{ kHz}$ | | 108 | | dB |

- (1) In differential mode at $V_{OUTx\pm}$ pin, $f_{OUT} = 1\text{ kHz}$, using Audio Precision System II, RMS mode with 20-kHz low-pass filter and 400-Hz high-pass filter.
(2) $f_S = 48\text{ kHz}$: $SCKI = 512 f_S$ (single), $f_S = 96\text{ kHz}$: $SCKI = 256 f_S$ (dual), $f_S = 192\text{ kHz}$: $SCKI = 128 f_S$ (quad).

ELECTRICAL CHARACTERISTICS: DAC Characteristics (continued)

All specifications at $T_A = +25^\circ\text{C}$, $V_{CCAD1} = V_{CCAD2} = V_{CCDA1} = V_{CCDA2} = 5\text{ V}$, $V_{DD1} = V_{DD2} = 3.3\text{ V}$, $f_S = 48\text{ kHz}$, $SCKI = 512 f_S$, 24-bit data, Sampling Mode = Auto for ADC and DAC, and Interface Mode = Slave for ADC and DAC, unless otherwise noted.

| PARAMETER | TEST CONDITIONS | PCM3168A, PCM3168A-Q1 | | | UNIT |
|--|---|---|------------------------|--------------------|---------------|
| | | MIN | TYP | MAX | |
| ANALOG OUTPUT | | | | | |
| Output voltage | Differential | | $1.6 \times V_{CCDA1}$ | | V_{PP} |
| Center voltage | | | $0.5 \times V_{CCDA1}$ | | V |
| Load impedance | To ac-coupled GND ⁽³⁾ | 5 | | | k Ω |
| | To dc-coupled GND ⁽³⁾ | 15 | | | k Ω |
| Low-pass filter frequency response | $f = 20\text{ kHz}$ | | -0.04 | | dB |
| | $f = 44\text{ kHz}$ | | -0.18 | | dB |
| DIGITAL FILTER PERFORMANCE⁽⁴⁾ Sharp roll-off | | | | | |
| Passband (single, dual) | | | | $0.454 \times f_S$ | Hz |
| Passband (quad) | | | | $0.432 \times f_S$ | Hz |
| Stop band (single, dual) | | $0.546 \times f_S$ | | | Hz |
| Stop band (quad) | | $0.569 \times f_S$ | | | Hz |
| Passband ripple | $< 0.454 \times f_S, 0.432 \times f_S$ | | | ± 0.0018 | dB |
| Stop band attenuation | $> 0.546 \times f_S, 0.569 \times f_S$ | -75 | | | dB |
| DIGITAL FILTER PERFORMANCE Slow roll-off | | | | | |
| Passband | | | | $0.328 \times f_S$ | Hz |
| Stop band | | $0.673 \times f_S$ | | | Hz |
| Passband ripple | $< 0.328 \times f_S$ | | | ± 0.0013 | dB |
| Stop band attenuation | $> 0.673 \times f_S$ | -75 | | | dB |
| DIGITAL FILTER PERFORMANCE⁽⁴⁾ | | | | | |
| Group delay time (single, dual) | | | $28/f_S$ | | sec |
| Group delay time (quad) | | | $19/f_S$ | | sec |
| De-emphasis error | | | ± 0.1 | | dB |
| POWER-SUPPLY REQUIREMENTS | | | | | |
| Voltage range | $V_{CCx1/2}$ | 4.5 | 5.0 | 5.5 | VDC |
| | $V_{DD1/2}$ | 3.0 | 3.3 | 3.6 | VDC |
| Supply current | I_{CC} | $f_S = 48\text{ kHz/ADC}, f_S = 48\text{ kHz/DAC}$ | 162 | 210 | mA |
| | | $f_S = 96\text{ kHz/ADC}, f_S = 192\text{ kHz/DAC}$ | 162 | | mA |
| | | Full power-down ⁽⁵⁾ | | 300 | μA |
| | I_{DD} | $f_S = 48\text{ kHz/ADC}, f_S = 48\text{ kHz/DAC}$ | 106 | 130 | mA |
| | | $f_S = 96\text{ kHz/ADC}, f_S = 192\text{ kHz/DAC}$ | 127 | | mA |
| | | Full power-down ⁽⁵⁾ | | 50 | μA |
| Power dissipation | $f_S = 48\text{ kHz/ADC}, f_S = 48\text{ kHz/DAC}$ | | 1160 | 1480 | mW |
| | $f_S = 96\text{ kHz/ADC}, f_S = 192\text{ kHz/DAC}$ | | 1230 | | mW |
| | $f_S = 48\text{ kHz/ADC}, \text{Power-down/DAC}$ | | 660 | | mW |
| | Power-down/ADC, $f_S = 48\text{ kHz/DAC}$ | | 633 | | mW |
| | Full power-down ⁽⁵⁾ | | 1.67 | | mW |

(3) Allowable minimum input resistance of differential to single-ended converter with D to S Gain = G is calculated as $(1 + 2G)/(1 + G) \times 5k$ for ac-coupled and $(1 + 0.9G)/(1 + G) \times 15k$ for dc-coupled connection, refer to [Figure 62](#) and [Figure 63](#) of the [Application Information](#) section.

(4) Exclude single and dual at $128 f_S, 192 f_S$ system clock and quad at $256 f_S$ to $768 f_S$ system clock, and specifications for quad, single, and dual are respectively applied in reverse for them.

(5) Halt SCKI, BCKAD, BCKDA, LRCKAD, and LRCKDA.

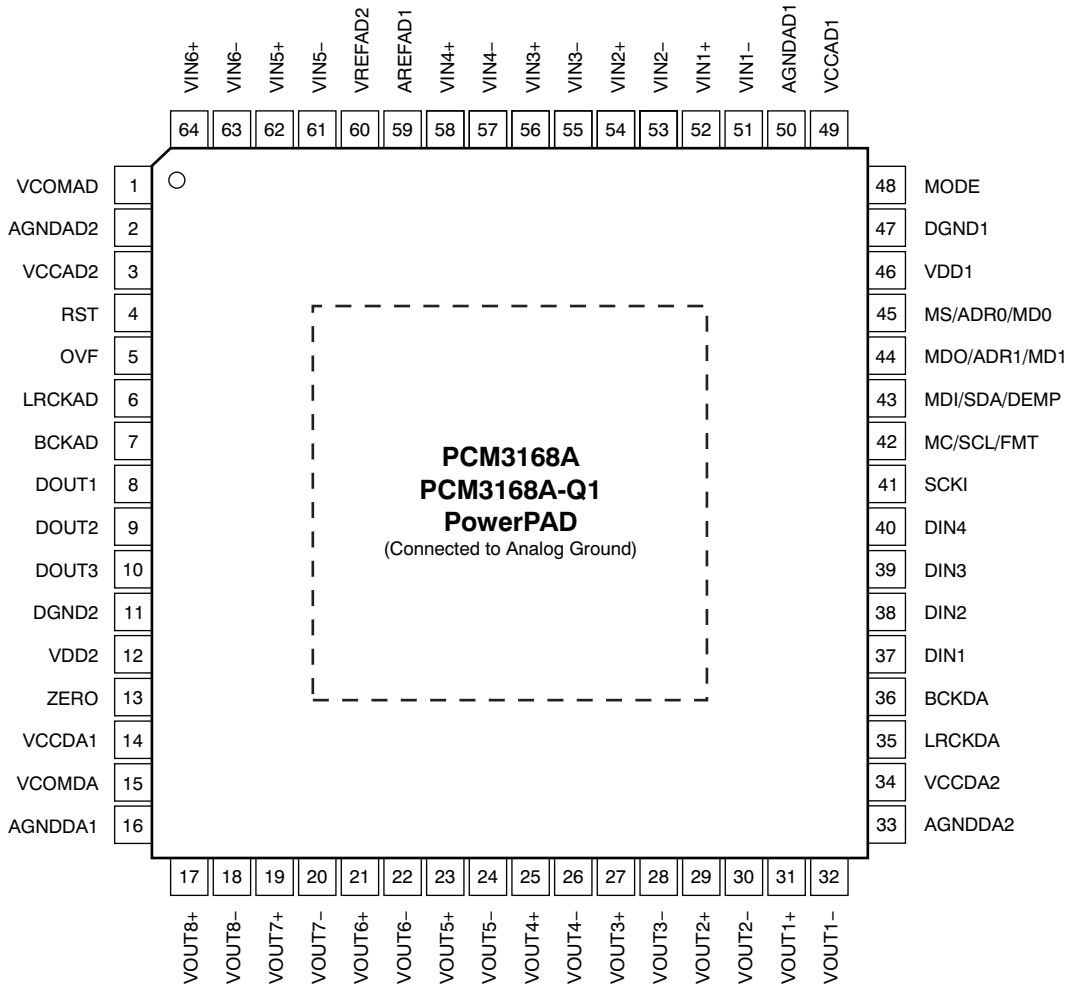
ELECTRICAL CHARACTERISTICS: DAC Characteristics (continued)

All specifications at $T_A = +25^\circ\text{C}$, $V_{CCAD1} = V_{CCAD2} = V_{CCDA1} = V_{CCDA2} = 5\text{ V}$, $V_{DD1} = V_{DD2} = 3.3\text{ V}$, $f_S = 48\text{ kHz}$, $SCKI = 512\text{ f}_S$, 24-bit data, Sampling Mode = Auto for ADC and DAC, and Interface Mode = Slave for ADC and DAC, unless otherwise noted.

| PARAMETER | TEST CONDITIONS | PCM3168A, PCM3168A-Q1 | | | UNIT |
|--------------------------|------------------------------------|-----------------------|-----|------|--------------------|
| | | MIN | TYP | MAX | |
| TEMPERATURE RANGE | | | | | |
| Operating temperature | PCM3168A Consumer grade | -40 | | +85 | $^\circ\text{C}$ |
| | PCM3168A-Q1 Automotive audio grade | -40 | | +105 | $^\circ\text{C}$ |
| Thermal resistance | θ_{JA} HTQFP-64 | | +21 | | $^\circ\text{C/W}$ |

DEVICE INFORMATION

HTQFP-64 (12 mm x 12 mm)
(10-mm x 10-mm body, 0.5-mm pitch)
(TOP VIEW)



TERMINAL FUNCTIONS

| TERMINAL | | I/O | PULL-DOWN | 5-V TOLERANT | DESCRIPTION |
|--------------|-----|-----|-----------|--------------|--|
| NAME | PIN | | | | |
| VCOMAD | 1 | — | No | No | ADC analog common voltage decoupling |
| AGNDAD2 | 2 | — | No | No | Analog ground 2 for ADC |
| VCCAD2 | 3 | — | No | No | ADC analog power supply 2, +5 V |
| RST | 4 | I | Yes | Yes | Reset and power-down control input with active low |
| OVF | 5 | O | No | No | Overflow flag output for ADC |
| LRCKAD | 6 | I/O | Yes | No | Audio data word clock input/output for ADC |
| BCKAD | 7 | I/O | Yes | No | Audio data bit clock input/output for ADC |
| DOUT1 | 8 | O | No | No | Audio data digital output for ADC1 and ADC2 |
| DOUT2 | 9 | O | No | No | Audio data digital output for ADC3 and ADC4 |
| DOUT3 | 10 | O | No | No | Audio data digital output for ADC5 and ADC6 |
| DGND2 | 11 | — | No | No | Digital ground 2 |
| VDD2 | 12 | — | No | No | Digital power supply 2, +3.3 V |
| ZERO | 13 | O | No | No | Zero detect flag output for DAC |
| VCCDA1 | 14 | — | No | No | DAC analog power supply 1, +5 V |
| VCOMDA | 15 | — | No | No | DAC voltage common decoupling |
| AGNDDA1 | 16 | — | No | No | Analog ground 1 for DAC |
| VOUT8+ | 17 | O | No | No | Positive analog output from DAC8 |
| VOUT8– | 18 | O | No | No | Negative analog output from DAC8 |
| VOUT7+ | 19 | O | No | No | Positive analog output from DAC7 |
| VOUT7– | 20 | O | No | No | Negative analog output from DAC7 |
| VOUT6+ | 21 | O | No | No | Positive analog output from DAC6 |
| VOUT6– | 22 | O | No | No | Negative analog output from DAC6 |
| VOUT5+ | 23 | O | No | No | Positive analog output from DAC5 |
| VOUT5– | 24 | O | No | No | Negative analog output from DAC5 |
| VOUT4+ | 25 | O | No | No | Positive analog output from DAC4 |
| VOUT4– | 26 | O | No | No | Negative analog output from DAC4 |
| VOUT3+ | 27 | O | No | No | Positive analog output from DAC3 |
| VOUT3– | 28 | O | No | No | Negative analog output from DAC3 |
| VOUT2+ | 29 | O | No | No | Positive analog output from DAC2 |
| VOUT2– | 30 | O | No | No | Negative analog output from DAC2 |
| VOUT1+ | 31 | O | No | No | Positive analog output from DAC1 |
| VOUT1– | 32 | O | No | No | Negative analog output from DAC1 |
| AGNDDA2 | 33 | — | No | No | Analog ground 2 for DAC |
| VCCDA2 | 34 | — | No | No | DAC analog power supply 2, +5 V |
| LRCKDA | 35 | I/O | Yes | No | Audio data word clock input/output for DAC |
| BCKDA | 36 | I/O | Yes | No | Audio data bit clock input/output for DAC |
| DIN1 | 37 | I | No | No | Audio data input for DAC1 and DAC2 |
| DIN2 | 38 | I | No | No | Audio data input for DAC3 and DAC4 |
| DIN3 | 39 | I | No | No | Audio data input for DAC5 and DAC6 |
| DIN4 | 40 | I | No | No | Audio data Input for DAC7 and DAC8 |
| SCKI | 41 | I | No | Yes | System clock input |
| MC/SCL/FMT | 42 | I | No | Yes | Clock for SPI, clock for I ² C, format select for hardware control mode |
| MDI/SDA/DEMP | 43 | I/O | No | Yes | Input data for SPI, data for I ² C ⁽¹⁾ , de-emphasis control for hardware control mode |

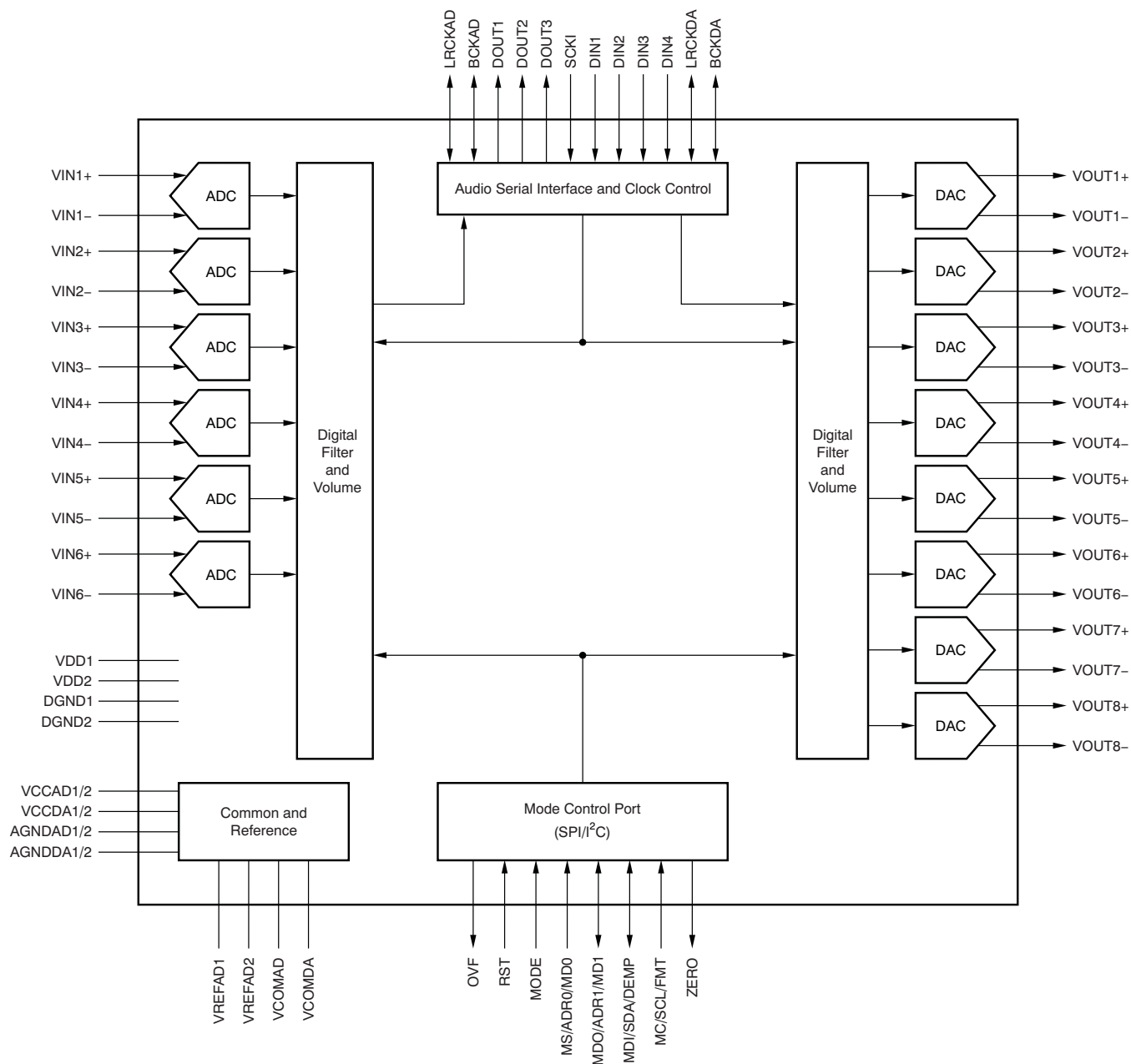
(1) Open-drain configuration in I²C.

TERMINAL FUNCTIONS (continued)

| TERMINAL | | I/O | PULL-DOWN | 5-V TOLERANT | DESCRIPTION |
|--------------|-----|-----|-----------|--------------|---|
| NAME | PIN | | | | |
| MDO/ADR1/MD1 | 44 | I/O | No | No | Output data for SPI ⁽²⁾ , address select 1 for I ² C, mode select 1 for hardware control mode |
| MS/ADR0/MD0 | 45 | I | Yes | Yes | Chip select for SPI, address select 0 for I ² C, mode select 0 for hardware control mode |
| VDD1 | 46 | — | No | No | Digital power supply 1, +3.3 V |
| DGND1 | 47 | — | No | No | Digital ground 1 |
| MODE | 48 | I | No | No | Control port mode selection. Tied to VDD: SPI, pull-up: H/W single-ended input, pull-down: H/W and differential input, tied to DGND: I ² C |
| VCCAD1 | 49 | — | No | No | ADC analog power supply 1, +5 V |
| AGNDAD1 | 50 | — | No | No | Analog ground 1 for ADC |
| VIN1– | 51 | I | No | No | Negative analog input to ADC1 |
| VIN1+ | 52 | I | No | No | Positive analog input to ADC1 |
| VIN2– | 53 | I | No | No | Negative analog input to ADC2 |
| VIN2+ | 54 | I | No | No | Positive analog input to ADC2 |
| VIN3– | 55 | I | No | No | Negative analog input to ADC3 |
| VIN3+ | 56 | I | No | No | Positive analog input to ADC3 |
| VIN4– | 57 | I | No | No | Negative analog input to ADC4 |
| VIN4+ | 58 | I | No | No | Positive analog input to ADC4 |
| VREFAD1 | 59 | — | No | No | ADC analog reference voltage 1 decoupling |
| VREFAD2 | 60 | — | No | No | ADC analog reference voltage 2 decoupling |
| VIN5– | 61 | I | No | No | Negative analog input to ADC5 |
| VIN5+ | 62 | I | No | No | Positive analog input to ADC5 |
| VIN6– | 63 | I | No | No | Negative analog input to ADC6 |
| VIN6+ | 64 | I | No | No | Positive analog input to ADC6 |

(2) 3-state (Hi-Z) operation in SPI.

FUNCTIONAL BLOCK DIAGRAM



TYPICAL CHARACTERISTICS ADC Digital Filter

All specifications at $T_A = +25^\circ\text{C}$, $V_{CCAD1} = V_{CCAD2} = V_{CCDA1} = V_{CCDA2} = 5\text{ V}$, $V_{DD1} = V_{DD2} = 3.3\text{ V}$, $f_S = 48\text{ kHz}$, $SCKI = 512 f_S$, 24-bit data, Sampling Mode = Auto for ADC and DAC, and Interface Mode = Slave for ADC and DAC, unless otherwise noted.

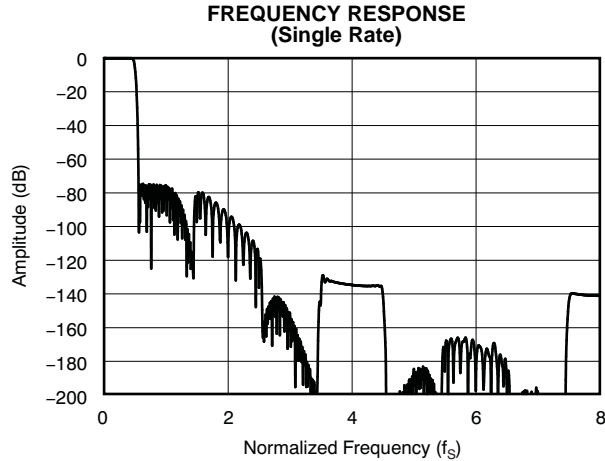


Figure 1.

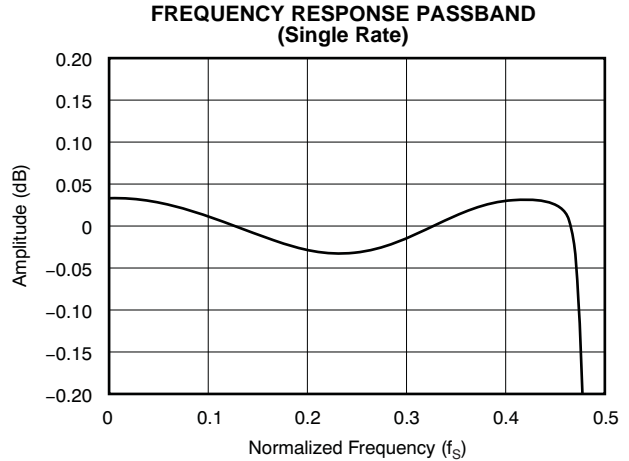


Figure 2.

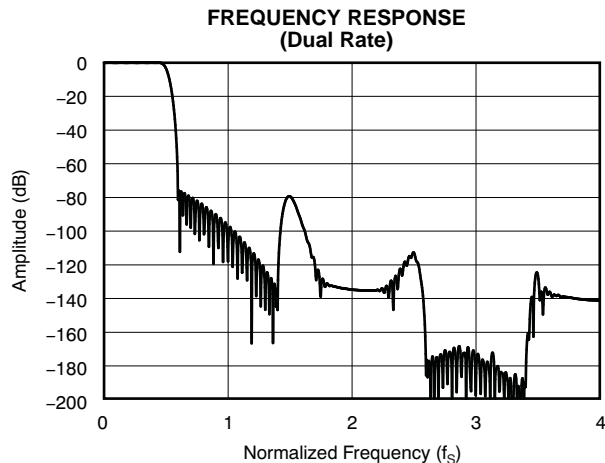


Figure 3.

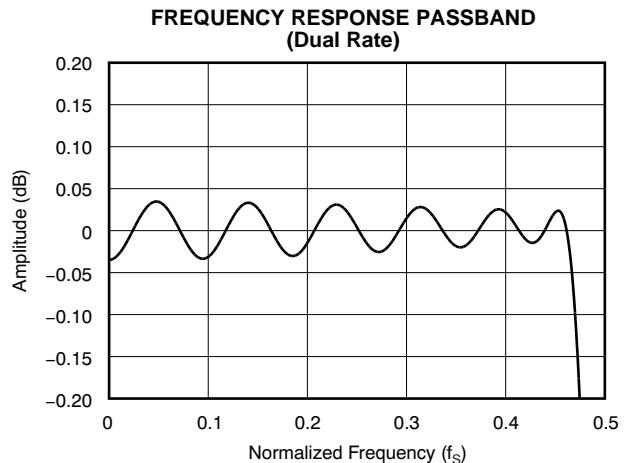


Figure 4.

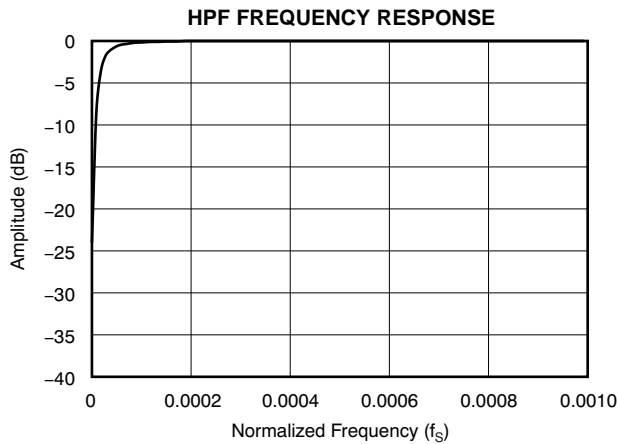


Figure 5.

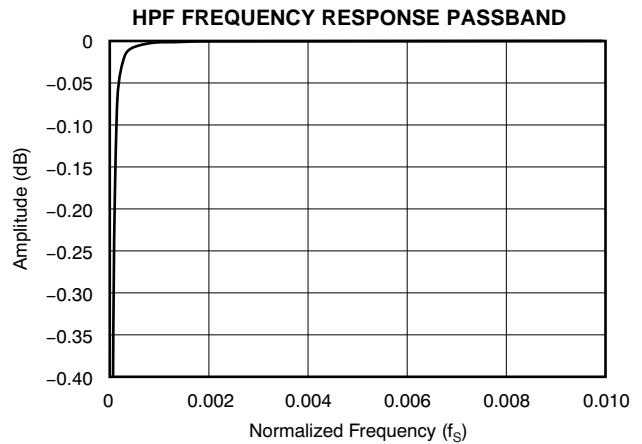


Figure 6.

DAC Digital Filter

All specifications at $T_A = +25^\circ\text{C}$, $V_{CCAD1} = V_{CCAD2} = V_{CCDA1} = V_{CCDA2} = 5\text{ V}$, $V_{DD1} = V_{DD2} = 3.3\text{ V}$, $f_S = 48\text{ kHz}$, $SCK1 = 512 f_S$, 24-bit data, Sampling Mode = Auto for ADC and DAC, and Interface Mode = Slave for ADC and DAC, unless otherwise noted.

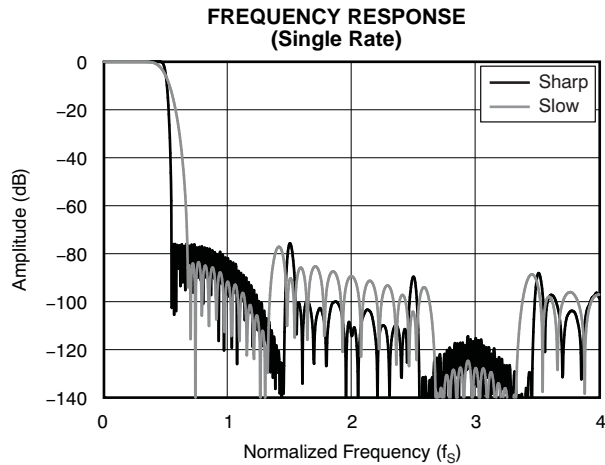


Figure 7.

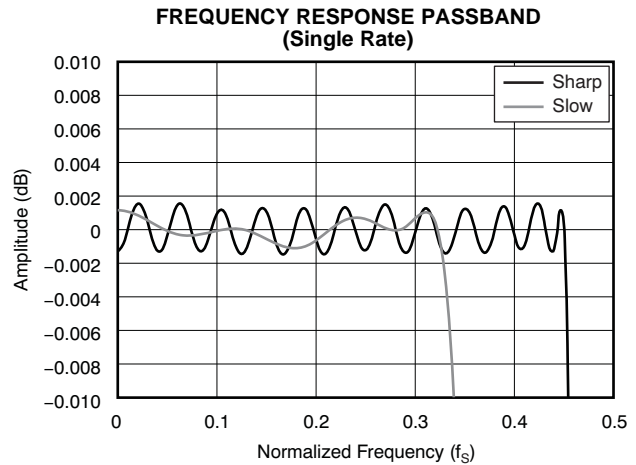


Figure 8.

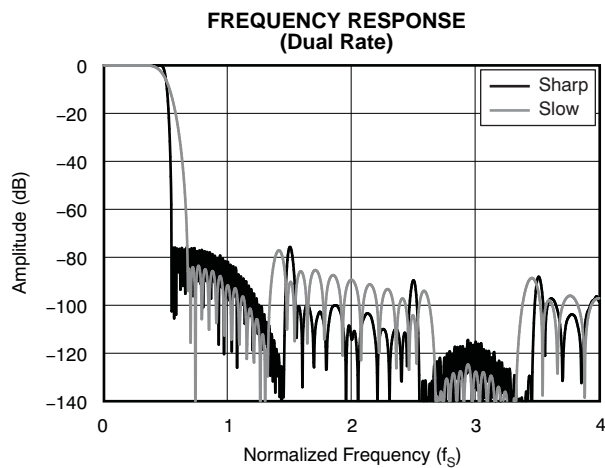


Figure 9.

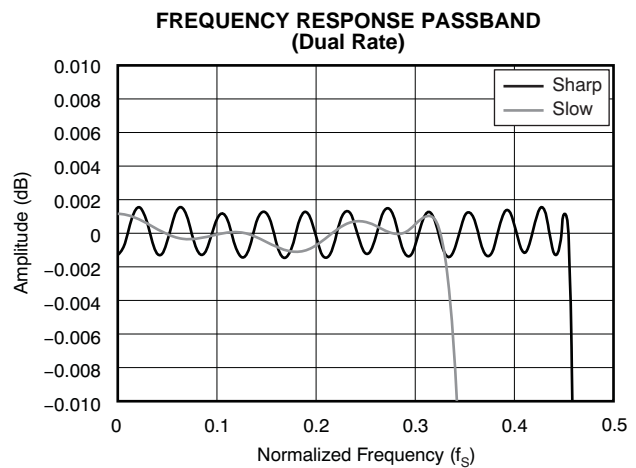


Figure 10.

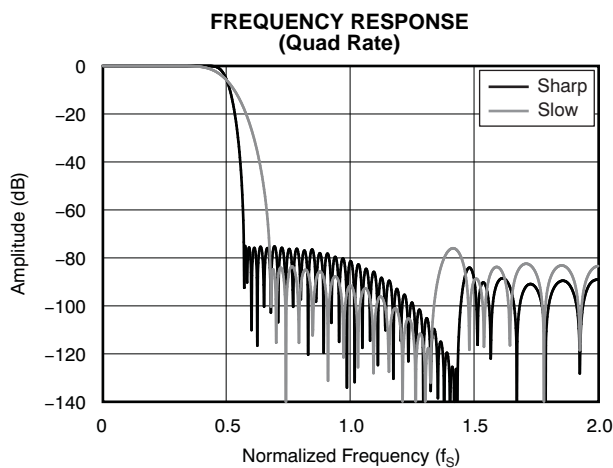


Figure 11.

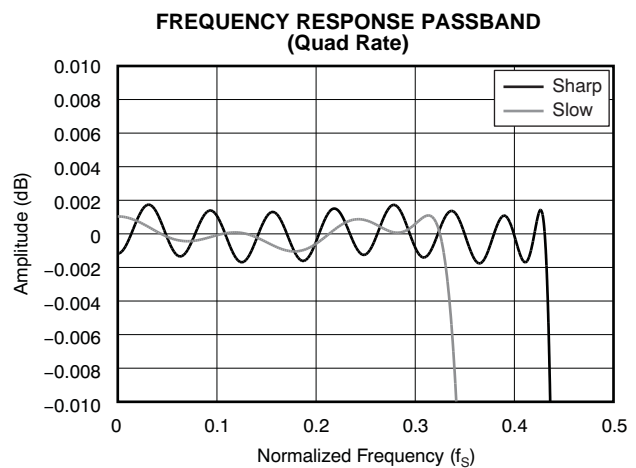


Figure 12.

DAC Digital Filter (continued)

All specifications at $T_A = +25^\circ\text{C}$, $V_{CAD1} = V_{CAD2} = V_{CCDA1} = V_{CCDA2} = 5\text{ V}$, $V_{DD1} = V_{DD2} = 3.3\text{ V}$, $f_S = 48\text{ kHz}$, $SCKI = 512 f_S$, 24-bit data, Sampling Mode = Auto for ADC and DAC, and Interface Mode = Slave for ADC and DAC, unless otherwise noted.

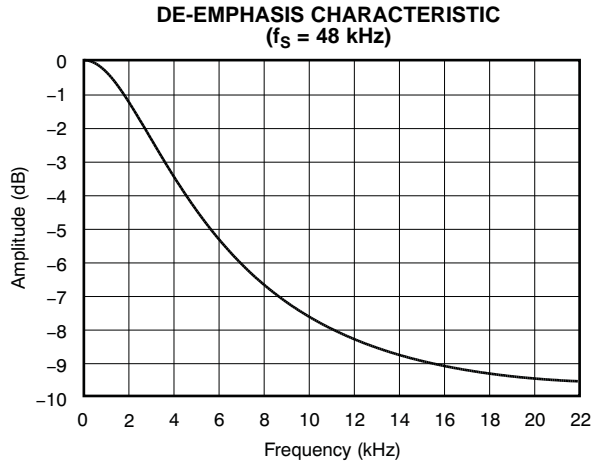


Figure 13.

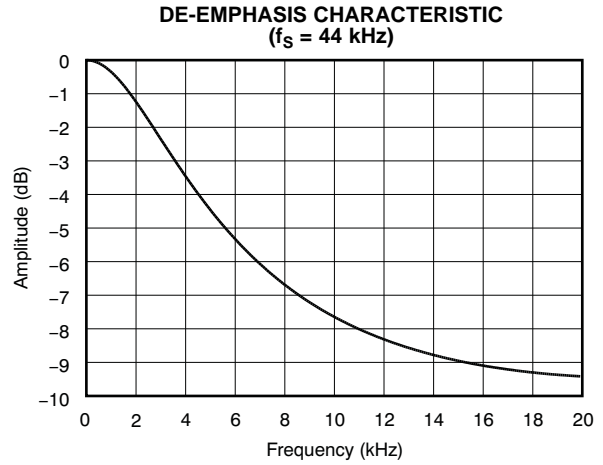


Figure 14.

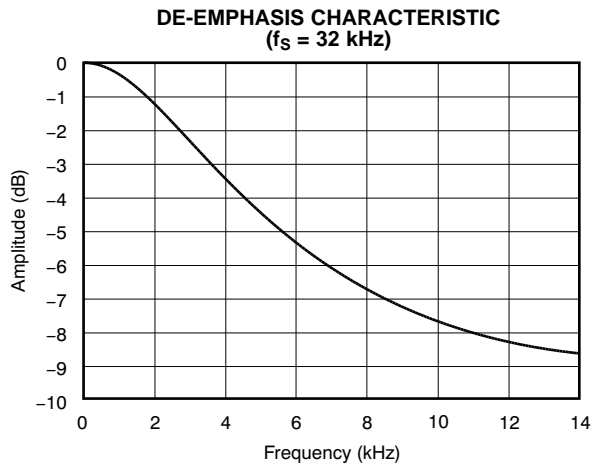


Figure 15.

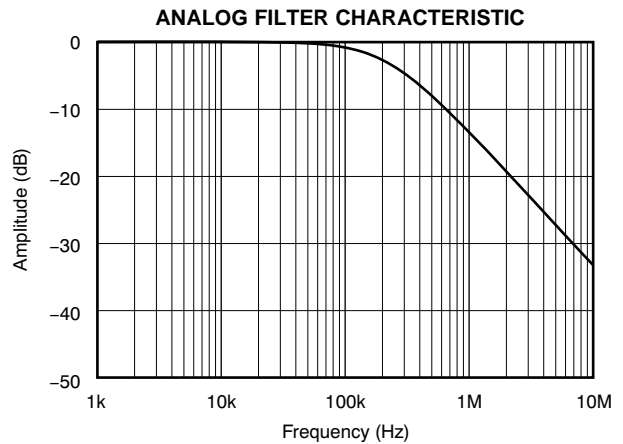


Figure 16.

ADC Performance

All specifications at $T_A = +25^\circ\text{C}$, $V_{\text{CAD1}} = V_{\text{CAD2}} = V_{\text{CDA1}} = V_{\text{CDA2}} = 5\text{ V}$, $V_{\text{DD1}} = V_{\text{DD2}} = 3.3\text{ V}$, $f_s = 48\text{ kHz}$, $\text{SCKI} = 512 f_s$, 24-bit data, Sampling Mode = Auto for ADC and DAC, and Interface Mode = Slave for ADC and DAC, unless otherwise noted.

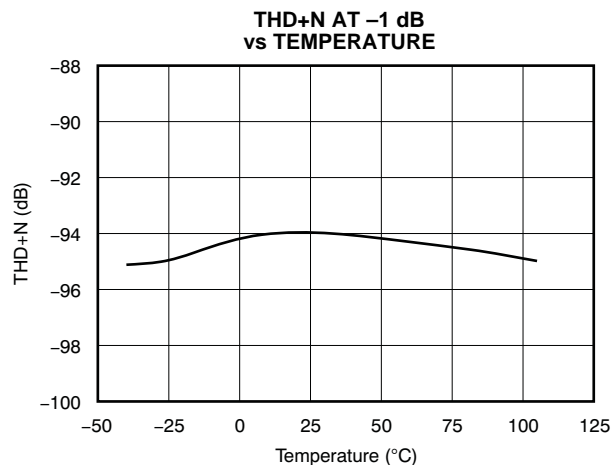


Figure 17.

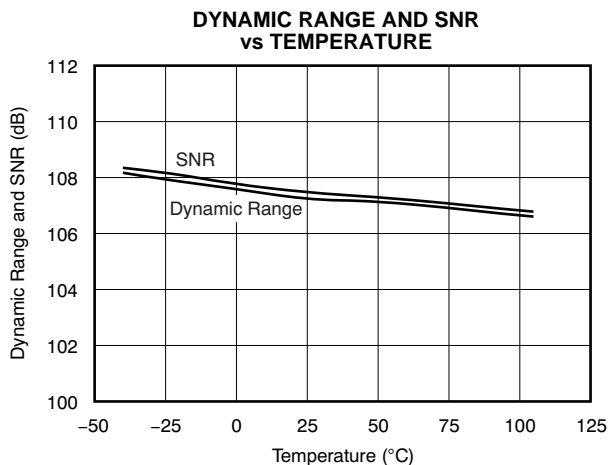


Figure 18.

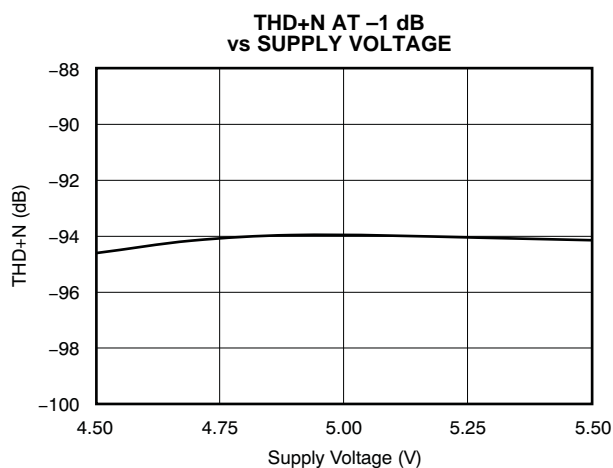


Figure 19.

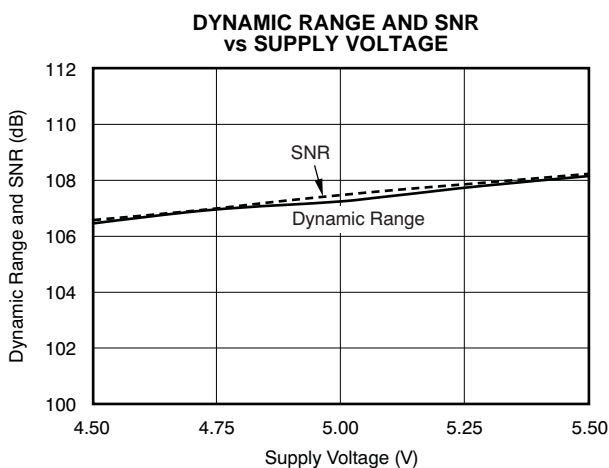


Figure 20.

DAC Performance

All specifications at $T_A = +25^\circ\text{C}$, $V_{CCAD1} = V_{CCAD2} = V_{CCDA1} = V_{CCDA2} = 5\text{ V}$, $V_{DD1} = V_{DD2} = 3.3\text{ V}$, $f_S = 48\text{ kHz}$, $SCKI = 512 f_S$, 24-bit data, Sampling Mode = Auto for ADC and DAC, and Interface Mode = Slave for ADC and DAC, unless otherwise noted.

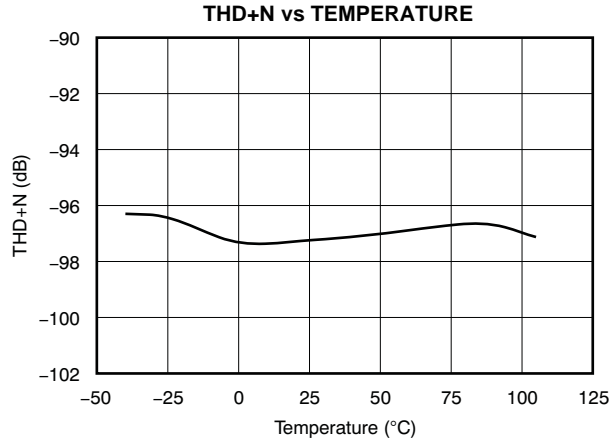


Figure 21.

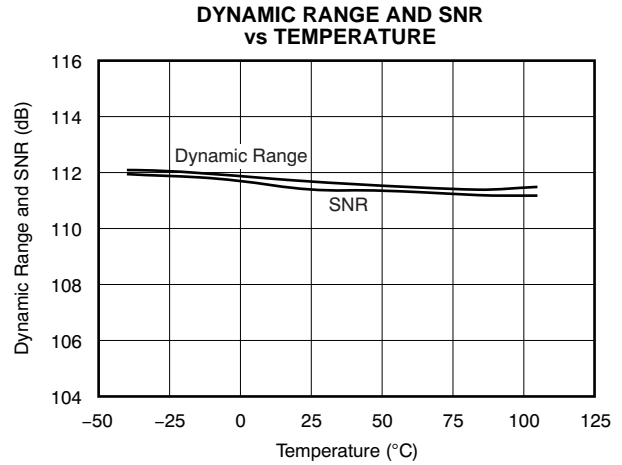


Figure 22.

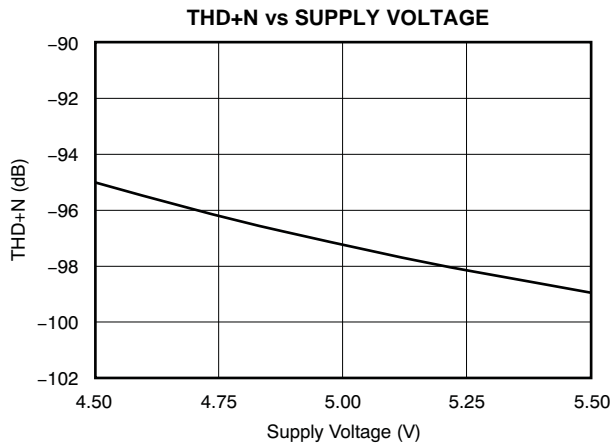


Figure 23.

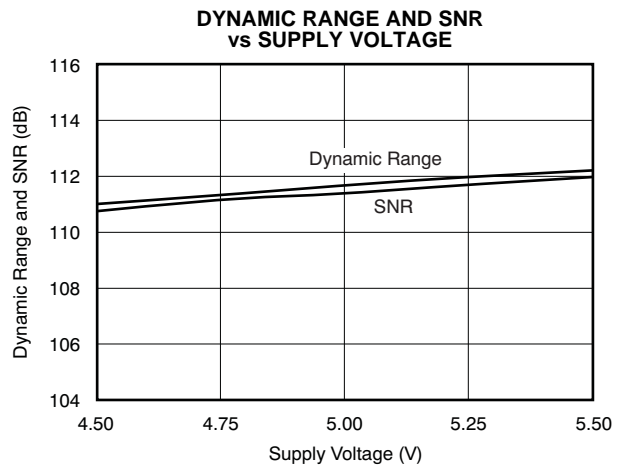


Figure 24.

Output Spectrum

All specifications at $T_A = +25^\circ\text{C}$, $V_{CCAD1} = V_{CCAD2} = V_{CCDA1} = V_{CCDA2} = 5\text{ V}$, $V_{DD1} = V_{DD2} = 3.3\text{ V}$, $f_S = 48\text{ kHz}$, $SCKI = 512 f_S$, 24-bit data, Sampling Mode = Auto for ADC and DAC, and Interface Mode = Slave for ADC and DAC, unless otherwise noted.

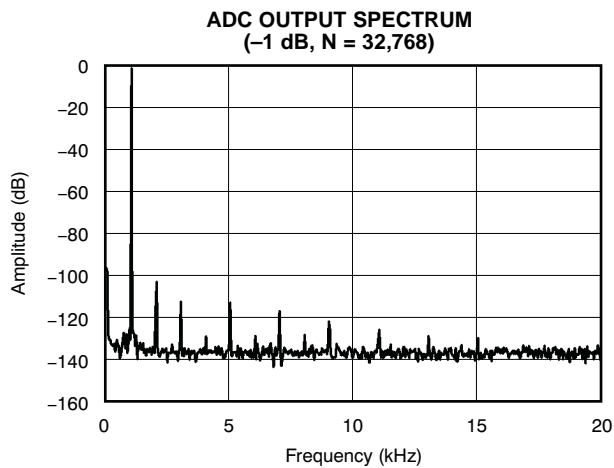


Figure 25.

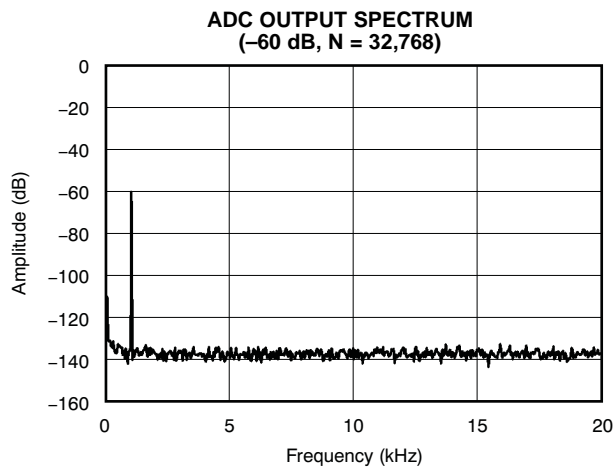


Figure 26.

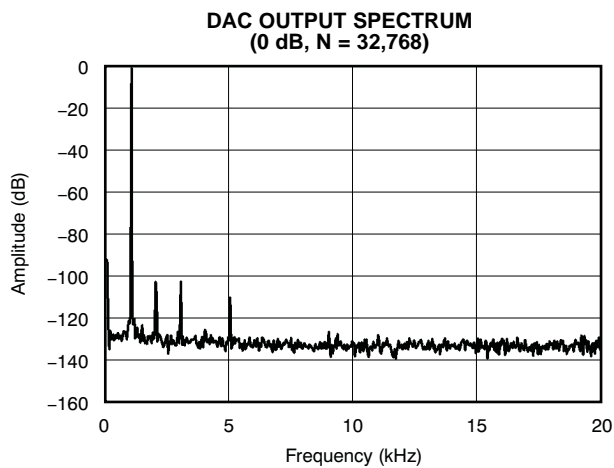


Figure 27.

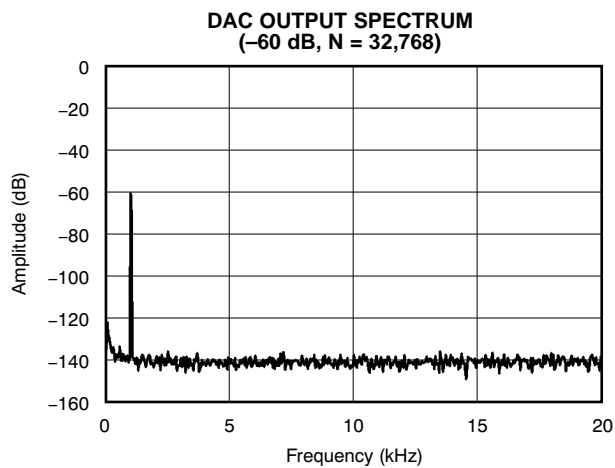


Figure 28.

Power-Supply

All specifications at $T_A = +25^\circ\text{C}$, $V_{CCAD1} = V_{CCAD2} = V_{CCDA1} = V_{CCDA2} = 5\text{ V}$, $V_{DD1} = V_{DD2} = 3.3\text{ V}$, $f_S = 48\text{ kHz}$, $SCKI = 512 f_S$, 24-bit data, Sampling Mode = Auto for ADC and DAC, and Interface Mode = Slave for ADC and DAC, unless otherwise noted.

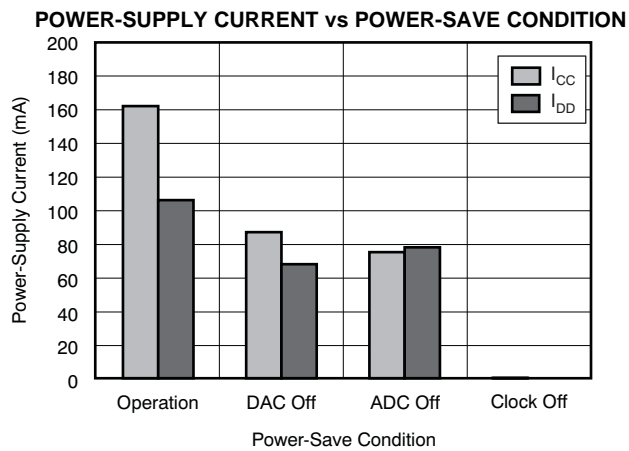


Figure 29.

PRODUCT OVERVIEW

The PCM3168A and PCM3168A-Q1 are high-performance, multi-channel codecs targeted for automotive audio applications such as external amplifiers, as well as home multi-channel audio applications (for example, home theaters and A/V receivers). The PCM3168A and PCM3168A-Q1 consist of six-channel analog-to-digital converters (ADCs) and eight-channel digital-to-analog converters (DACs). The ADC input is selectable between single-ended and differential inputs. The DAC output type is fixed with a differential configuration. The PCM3168A and PCM3168A-Q1 support 24-bit linear PCM input and output data in standard audio formats (left-justified, right-justified, and I²S), DSP and TDM formats, and various sample frequencies from 8 kHz to 192 kHz (the ADC configuration supports only up to 96 kHz). The TDM format is useful to save interface bus line numbers for multi-channel audio data communication between the codec and digital audio processor. The PCM3168A and PCM3168A-Q1 offer three modes for device control: two-wire I²C software, four-wire SPI software, and hardware modes.

ANALOG INPUTS

The PCM3168A and PCM3168A-Q1 include six ADCs, each with individual pairs of differential voltage input pins, as shown in [Table 1](#). Additionally, the PCM3168A and PCM3168A-Q1 have the capability of single-ended inputs. The full-scale input voltage is $(0.2 \times V_{CCAD1}) V_{RMS}$ at the single-ended input mode and $(0.4 \times V_{CCAD1}) V_{RMS}$ at the differential input mode. The input mode is selected by the MODE pin in hardware control mode or by register settings in the software control mode. In single-ended mode, VINx+ pins are used and VINx– pins must be terminated with AGNDAD1/2 via a capacitor or terminated with VCOMAD.

Table 1. Pin Assignments in Differential and Single-Ended Input Modes

| CHANNEL | DIFFERENTIAL INPUT MODE | SINGLE-ENDED INPUT MODE |
|----------|-------------------------|-------------------------|
| 1 (ADC1) | VIN1+, VIN1– | VIN1+ |
| 2 (ADC2) | VIN2+, VIN2– | VIN2+ |
| 3 (ADC3) | VIN3+, VIN3– | VIN3+ |
| 4 (ADC4) | VIN4+, VIN4– | VIN4+ |
| 5 (ADC5) | VIN5+, VIN5– | VIN5+ |
| 6 (ADC6) | VIN6+, VIN6– | VIN6+ |

ANALOG OUTPUTS

The PCM3168A and PCM3168A-Q1 include eight DACs, each with individual pairs of differential voltage inputs pins, as shown in [Table 2](#). The full-scale output voltage is $(1.6 \times V_{CCDA1}) V_{PP}$ in differential mode. DC-coupled loads are allowed in addition to ac-coupled loads if the load resistance conforms to the specification.

Table 2. Pin Assignments for Differential Output

| CHANNEL | DIFFERENTIAL OUTPUT |
|----------|---------------------|
| 1 (DAC1) | VOUT1+, VOUT1– |
| 2 (DAC2) | VOUT2+, VOUT2– |
| 3 (DAC3) | VOUT3+, VOUT3– |
| 4 (DAC4) | VOUT4+, VOUT4– |
| 5 (DAC5) | VOUT5+, VOUT5– |
| 6 (DAC6) | VOUT6+, VOUT6– |
| 7 (DAC7) | VOUT7+, VOUT7– |
| 8 (DAC8) | VOUT8+, VOUT8– |

VOLTAGE REFERENCES

The PCM3168A and PCM3168A-Q1 include two internal references for the six-channel ADCs; these references correspond to the outputs VREFAD1 and VREFAD2. Both reference pins should be connected with an analog ground via decoupling capacitors. In addition, the PCM3168A and PCM3168A-Q1 include two pins for common-mode voltage output (VCOMDA for DACs and VCOMAD for ADCs). These pins should be also connected with an analog ground via decoupling capacitors. Furthermore, both common pins can be used to bias external high-impedance circuits, if they are required.

SYSTEM CLOCK INPUT

The PCM3168A and PCM3168A-Q1 require an external system clock input applied at the SCKI input for ADC and DAC operation. The system clock operates at an integer multiple of the sampling frequency, or f_s . The multiples supported in ADC operation include $256 f_s$, $384 f_s$, $512 f_s$, and $768 f_s$; the multiples supported in DAC operation include $128 f_s$, $192 f_s$, $256 f_s$, $384 f_s$, $512 f_s$, and $768 f_s$. Details for these system clock multiples are shown in Table 3. Figure 30 and Table 4 show the SCKI timing requirements.

Table 3. System Clock Frequencies for Common Audio Sampling Rates

| DEFAULT SAMPLING MODE | SAMPLING FREQUENCY f_s (kHz) | SYSTEM CLOCK FREQUENCY (MHz) | | | | | |
|--------------------------|-----------------------------------|------------------------------|------------------------|-----------|------------------------|-----------|-----------|
| | | $128 f_s^{(1)}$ | $192 f_s^{(1)}$ | $256 f_s$ | $384 f_s$ | $512 f_s$ | $768 f_s$ |
| Single rate | 8 | N/A | N/A | 2.0480 | 3.0720 ⁽²⁾ | 4.0960 | 6.1440 |
| | 16 | 2.0480 ⁽¹⁾ | 3.0720 ⁽¹⁾ | 4.0960 | 6.1440 ⁽²⁾ | 8.1920 | 12.2880 |
| | 32 | 4.0960 ⁽¹⁾ | 6.1440 ⁽¹⁾ | 8.1920 | 12.2880 ⁽²⁾ | 16.3840 | 24.5760 |
| | 44.1 | 5.6488 ⁽¹⁾ | 8.4672 ⁽¹⁾ | 11.2896 | 16.9344 ⁽²⁾ | 22.5792 | 33.8688 |
| | 48 | 6.1440 ⁽¹⁾ | 9.2160 ⁽¹⁾ | 12.2880 | 18.4320 ⁽²⁾ | 24.5760 | 36.8640 |
| Dual rate | 88.2 | 11.2896 ⁽³⁾ | 16.9344 ⁽³⁾ | 22.5792 | 33.8688 | N/A | N/A |
| | 96 | 12.2880 ⁽³⁾ | 18.4320 ⁽³⁾ | 24.5760 | 36.8640 | N/A | N/A |
| Quad rate ⁽³⁾ | 176.4 ⁽³⁾ | 22.5792 ⁽³⁾ | 33.8688 ⁽³⁾ | N/A | N/A | N/A | N/A |
| | 192 ⁽³⁾ | 24.5760 ⁽³⁾ | 36.8640 ⁽³⁾ | N/A | N/A | N/A | N/A |

- (1) Supported only by DAC operation
- (2) Requires 50% duty cycle for stable ADC performance.
- (3) Supported only by DAC operation

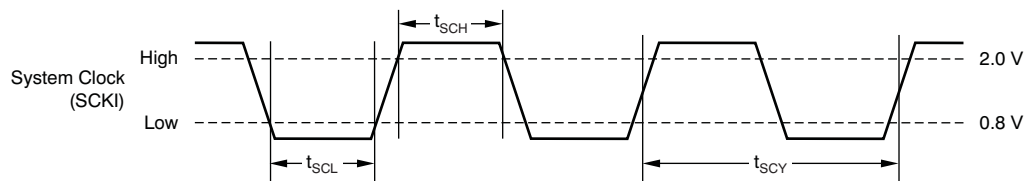


Figure 30. System Clock Timing Requirements

Table 4. Timing Requirements for Figure 30

| SYMBOL | PARAMETER | MIN | MAX | UNIT |
|-----------|-------------------------------|-----|-----|------|
| t_{SCY} | System clock pulse cycle time | 27 | | ns |
| t_{SCH} | System clock pulse width high | 10 | | ns |
| t_{SCL} | System clock pulse width low | 10 | | ns |
| t_{DTY} | System clock pulse duty cycle | 40 | 60 | % |

SAMPLING MODE

The PCM3168A and PCM3168A-Q1 support two sampling modes (single rate and dual rate) in ADC operation, and three sampling modes (single rate, dual rate, and quad rate) in DAC operation. In single rate mode, the ADC and DAC operate at an oversampling frequency of $\times 128$ (except when $SCKI = 128 f_s$ and $192 f_s$). This mode is supported for sampling frequencies less than 50 kHz. In dual rate mode, the ADC and DAC operate at an

oversampling frequency of x64; this mode is supported for sampling frequencies less than 100 kHz. In quad rate mode, the DAC operates at an oversampling frequency of x32. The sampling mode is automatically selected according to the ratio of system clock frequency and sampling frequency by default (for example, single rate for 512 f_s and 768 f_s , dual rate for 256 f_s and 384 f_s , and quad rate for 128 f_s and 192 f_s), but manual selection is also possible for specified combinations through the serial mode control resistor.

Table 5 and Figure 31 show the relation between the oversampling rate (OSR) of the $\Delta\Sigma$ modulator, noise-free shaped bandwidth, and each sampling mode setting for ADC operation. Table 6 and Figure 32 describe the relation between the oversampling rate of the digital filter and $\Delta\Sigma$ modulator, noise-free shaped bandwidth, and each sampling mode setting for DAC operation.

Table 5. ADC Modulator OSR and Noise-Free Shaped Bandwidth for Each Sampling Mode

| SAMPLING MODE REGISTER SETTING | SYSTEM CLOCK RATE (f_s) | NOISE-FREE SHAPED BANDWIDTH (kHz) | | MODULATOR OSR |
|--------------------------------|-----------------------------|-----------------------------------|----------------|---------------|
| | | $f_s = 48$ kHz | $f_s = 96$ kHz | |
| Auto | 512, 768 | 40 | N/A | x128 |
| | 256, 384 | 20 | 40 | x64 |
| Single | 512, 768 | 40 | N/A | x128 |
| | 256, 384 | 40 | N/A | x128 |
| Dual | 256, 384 | 20 | 40 | x64 |

Table 6. DAC Digital Filter OSR, Modulator OSR, and Noise-Free Shaped Bandwidth for Each Sampling Mode

| SAMPLING MODE REGISTER SETTING | SYSTEM CLOCK RATE (f_s) | NOISE-FREE SHAPED BANDWIDTH | | | DIGITAL FILTER OSR | MODULATOR OSR |
|--------------------------------|-----------------------------|-----------------------------|----------------|-----------------|--------------------|---------------|
| | | $f_s = 48$ kHz | $f_s = 96$ kHz | $f_s = 192$ kHz | | |
| Auto | 512, 768 | 40 | N/A | N/A | x8 | x128 |
| | 256, 384 | 20 | 40 | N/A | x8 | x64 |
| | 128, 192 ⁽¹⁾⁽²⁾ | 10 | 20 | 40 | x4 | x32 |
| Single | 512, 768 | 40 | N/A | N/A | x8 | x128 |
| | 256, 384 | 40 | N/A | N/A | x8 | x128 |
| | 128, 192 ⁽¹⁾⁽²⁾ | 20 | N/A | N/A | x4 | x64 |
| Dual | 256, 384 | 20 | 40 | N/A | x8 | x64 |
| | 128, 192 ⁽¹⁾⁽²⁾ | 20 | 40 | N/A | x4 | x64 |
| Quad | 128, 192 ⁽¹⁾⁽²⁾ | 10 | 20 | 40 | x4 | x32 |

(1) Supported only by DAC operation.

(2) Quad mode filter characteristic is applied.

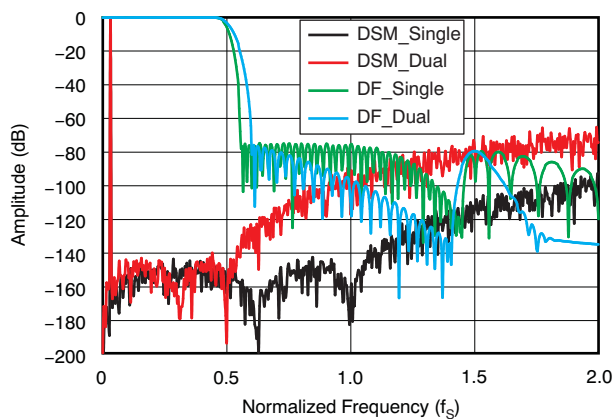


Figure 31. ADC $\Delta\Sigma$ Modulator and Digital Filter Characteristic

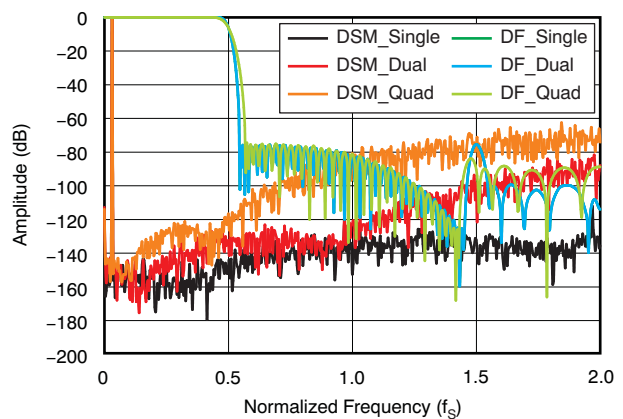


Figure 32. DAC $\Delta\Sigma$ Modulator and Digital Filter Characteristic

RESET OPERATION

The PCM3168A and PCM3168A-Q1 have both an internal power-on reset circuit and an external reset circuit. The sequences for both reset circuits are illustrated in Figure 33, Table 7, and Figure 34. Figure 33 and Table 7 describe the timing chart at the internal power-on reset. Initialization is triggered automatically at the point where VDD exceeds 2.2 V typical, and the internal reset is released after 3846 SCKI clock cycles from power-on if RST is kept high and SCKI is provided. VOUT from the DACs are forced to the VCOMDA level initially ($= 0.5 \times VCCDA1$) and settles at a specified level according to the rising VCC. If synchronization among SCKI, BCKAD/DA, and LRCKAD/DA is maintained, VOUT starts to output with a fade-in sequence after $t_{DACDLY1}$ from the internal reset release; VOUT then provides an output that corresponds to DIN after ($3846 \text{ SCKI} + t_{DACDLY1} + t_{DACDLY2}$) from power-on. Meanwhile, DOUT from the ADCs begins to output with a fade-in sequence after $t_{ADC DLY1}$ from the internal reset release; DOUT then provides output corresponding to VIN after ($3846 \text{ SCKI} + t_{ADC DLY1} + t_{ADC DLY2}$) from power-on. If the synchronization is not held, the internal reset is not released and both operating modes are maintained at reset and power-down states; after the synchronization forms again, both the DAC and ADC return to normal operation with the above sequences.

Figure 34 illustrates a timing chart at the external reset. RST accepts an external forced reset by RST = low, and provides a device reset and power-down state that makes the lowest power dissipation state available in the PCM3168A and PCM3168A-Q1. If RST goes from high to low under synchronization among SCKI, BCKAD/DA, and LRCKAD/DA, the internal reset is asserted, all registers and memory are reset, and finally the PCM3168A and PCM3168A-Q1 enter into all power-down states. At the same time, VOUT is immediately forced into the AGNDDA1 level and DOUT becomes '0'. To begin normal operation again, toggle RST high; the same power-up sequence as power-on reset shown in Figure 33 is performed.

The PCM3168A and PCM3168A-Q1 do not require particular power-on sequences for VCC and VDD; it allows VDD on and then VCC on, or VCC on and then VDD on. From the viewpoint of the Absolute Maximum Ratings, however, simultaneous power-on is recommended for avoiding unexpected responses on VOUTx and DOUTx. Figure 33 illustrates the response for VCC on with VDD on.

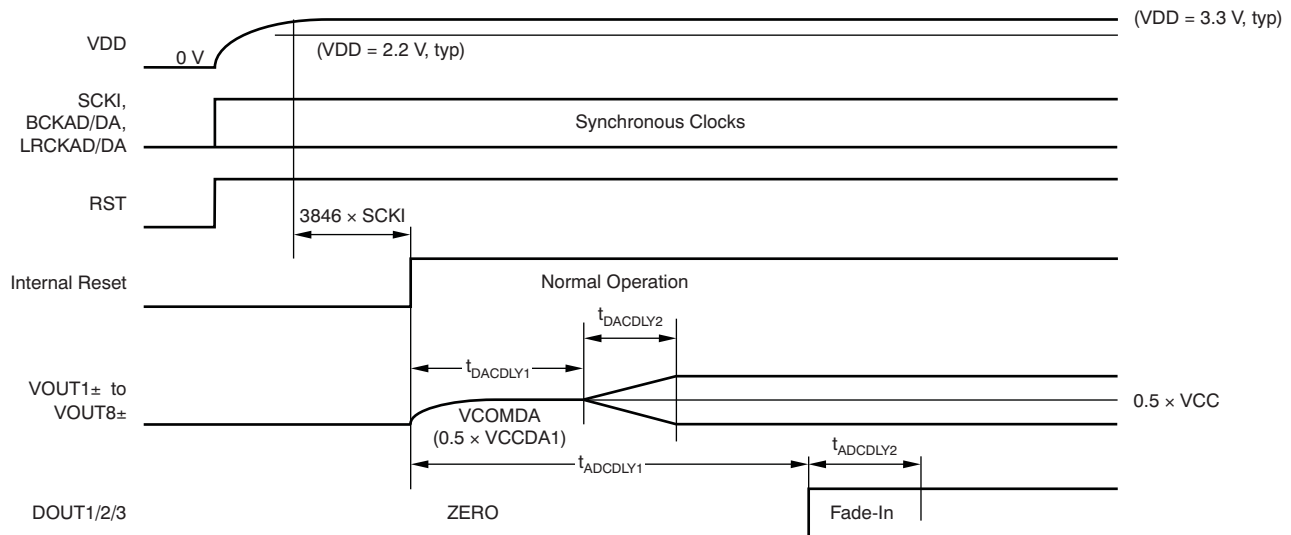
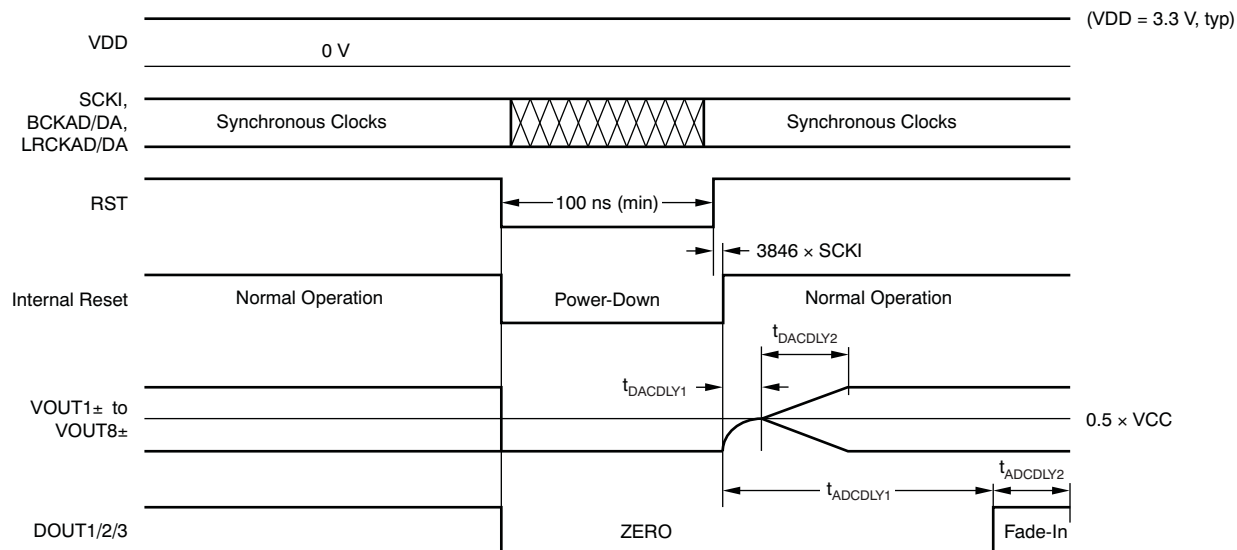


Figure 33. Power-On-Reset Timing Requirements

Table 7. Timing Requirements for Figure 33

| SYMBOL | DESCRIPTION | SINGLE | DUAL | QUAD | UNIT |
|----------------|---|--------|------|-------|------------------|
| $t_{DACDLY1}$ | DAC delay time internal reset release to VOUT start | 3600 | 7200 | 14400 | Period of LRCKDA |
| $t_{DACDLY2}$ | DAC fade-in/fade-out time | 2048 | 4096 | 8192 | Period of LRCKDA |
| $t_{ADC DLY1}$ | ADC delay time internal reset release to DOUT start | 4800 | 9600 | N/A | Period of LRCKAD |
| $t_{ADC DLY2}$ | ADC fade-in/fade-out time | 2048 | 4096 | N/A | Period of LRCKAD |


Figure 34. External Reset Timing Requirements

AUDIO SERIAL PORT OPERATION

The PCM3168A and PCM3168A-Q1 audio serial ports consist of 11 signals: BCKDA, BCKAD, LRCKDA, LRCKAD, DIN1, DIN2, DIN3, DIN4, DOUT1, DOUT2, and DOUT3. The PCM3168A and PCM3168A-Q1 also support audio interface mode, slave mode, and master mode. The BCKAD/DA is a bit clock input at the slave mode and an output at the master mode. The LRCKAD/DA is a left/right word clock or frame synchronization clock input at slave mode and output at master mode. The DIN1/2/3/4 are the audio data inputs for the DAC. The DOUT1/2/3 are the audio data outputs from the ADC. BCKAD, LRCKAD and DOUT1/2/3 are used for the ADC, and BCKDA, LRCKDA and DIN1/2/3/4 are used for the DAC.

AUDIO DATA INTERFACE FORMATS AND TIMING

The PCM3168A and PCM3168A-Q1 support eight audio data interface formats for the ADC and DAC separately in both master and slave modes: 24-bit I²S, 24-bit left-justified, 24-bit right-justified, 16-bit right-justified, 24-bit left-justified mode DSP, 24-bit I²S mode DSP, 24-bit left-justified mode TDM, and 24-bit I²S mode TDM format. The PCM3168A and PCM3168A-Q1 also support two audio data interface formats for the DAC and slave mode: 24-bit left-justified mode high-speed TDM and 24-bit I²S mode high-speed TDM format. In the case of I²S, left-justified, and right-justified data formats, 64 BCKs, 48 BCKs, and 32 BCKs per LRCK period are supported, but 48 BCKs are limited in slave mode and 32 BCKs are limited in slave mode 16-bit right-justified only. In the case of TDM data format in single rate, BCKAD/DA, LRCKAD/DA, DOUT1, and DIN1 are used. In the case of TDM data format in dual rate, BCKAD/DA, LRCKAD/DA, DOUT1/2, and DIN1/2 are used. In the case of high-speed TDM format in dual rate, BCKDA, LRCKDA, and DIN1 are used. In the case of high-speed TDM format in quad rate, BCKDA, LRCKDA, and DIN1/2 are used. TDM format and high-speed TDM format are supported only at $SCKI = 512 f_s$, $256 f_s$, $128 f_s$, and $f_{BCK} \leq f_{SCKI}$. The audio data formats are selected by MC/SCL/FMT in hardware control mode and registers 65 and 81 in software control mode. All data must be in binary twos complement, MSB first.

Figure 35 through Figure 41 show 10 audio interface data formats. Table 8 summarizes the applicable formats and describes the relationships among them and the respective restrictions with mode control.

Table 8. Audio Data Interface Formats and Sampling Rate, Bit Clock, and System Clock Restrictions

| CONTROL MODE | FORMAT | I/F MODE | DATA BITS | MAX LRCK FREQUENCY (f _s) | SCKI RATE (xf _s) | BCK RATE (xf _s) | APPLICABLE PINS |
|------------------|--|-----------------------------------|-----------|--------------------------------------|--------------------------------------|--|-------------------------|
| Software control | I ² S/Left-Justified | Master/Slave | 24 | 96 kHz (ADC) 192 kHz (DAC) | 256 to 768 (ADC) 128 to 768 (DAC) | 64, 48 (slave) ⁽¹⁾ | DOUT1/2/3 DIN1/2/3/4 |
| | Right-Justified | | 24, 16 | | | 64, 48 (slave) ⁽¹⁾ , 32 (slave, 16 bit) ⁽¹⁾ | |
| | I ² S/Left-Justified DSP | | 24 | 64 | | | |
| | I ² S/ Left-Justified TDM | | 24 | 48 kHz | 256, 512 | 256 | DOUT1, DIN1 |
| | High-Speed I ² S/Left-Justified TDM | Slave and DAC Only ⁽³⁾ | 24 | 96 kHz | 256 | 256 | DIN1 |
| 24 | | | 192 kHz | 128 | 128 | DIN1/2 | |
| Hardware control | I ² S | Master (ADC), Slave | 24 | 96 kHz (ADC) 192 kHz (DAC) | 256 to 768 (ADC) 128 to 768 (DAC) | 64, 48 (slave) ⁽¹⁾ | DOUT1/2/3 DIN1/2/3/4 |
| | I ² S TDM | | 24 | 48 kHz | 512 | 256 | DOUT1, DIN1 |
| | | | 24 | 96 kHz | 256 | 128 | DOUT1/2, DIN1/2 |

- (1) BCK = 48 f_s, 32 f_s is supported only in slave mode; BCK = 32 f_s is supported only for 16-bit data length.
- (2) SCKI = 128 f_s is supported only for DAC.
- (3) High-Speed I²S/Left-Justified TDM format is supported only for DAC operation in slave mode.

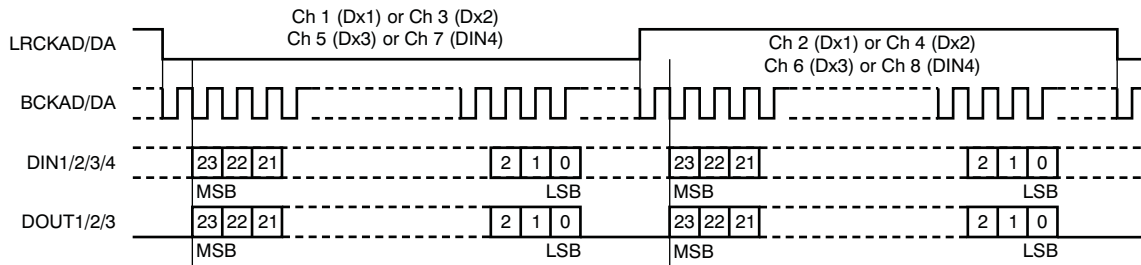


Figure 35. Audio Data Format: 24-Bit I²S

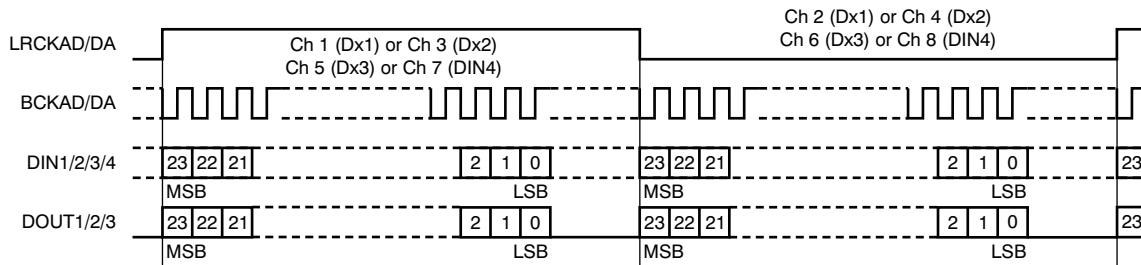


Figure 36. Audio Data Format: 24-Bit Left-Justified

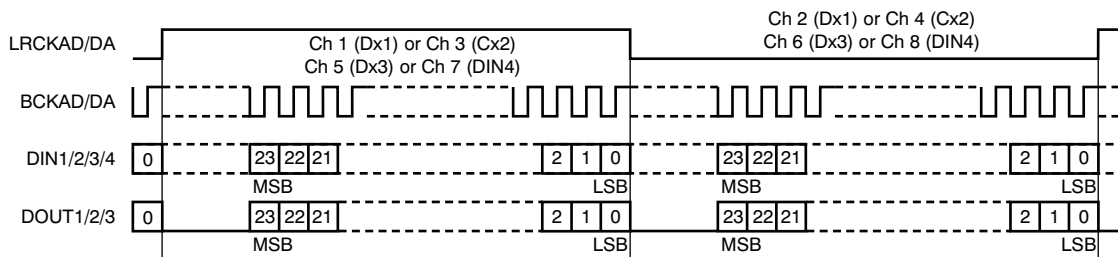


Figure 37. Audio Data Format: 24-Bit Right-Justified

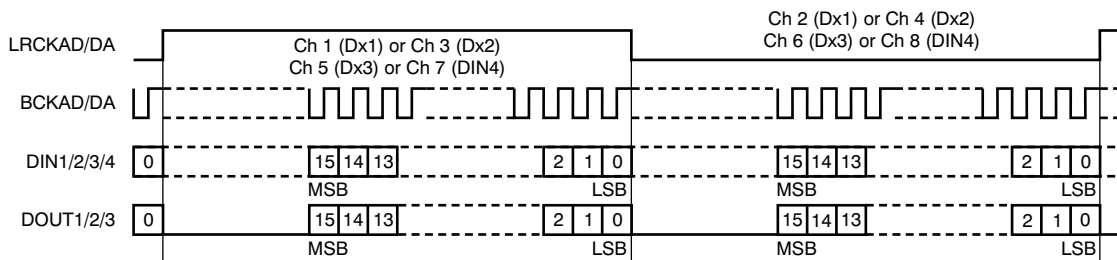


Figure 38. Audio Data Format: 16-Bit Right-Justified

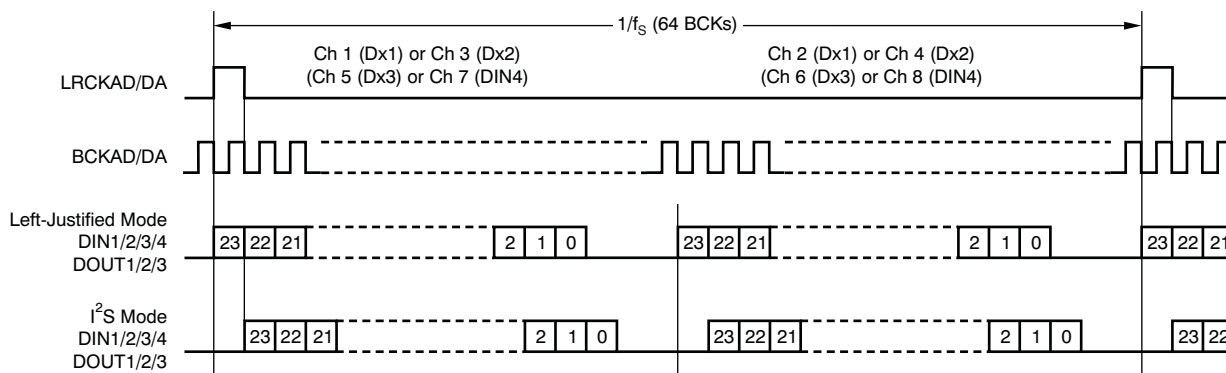


Figure 39. Audio Data Format: 24-Bit DSP Format

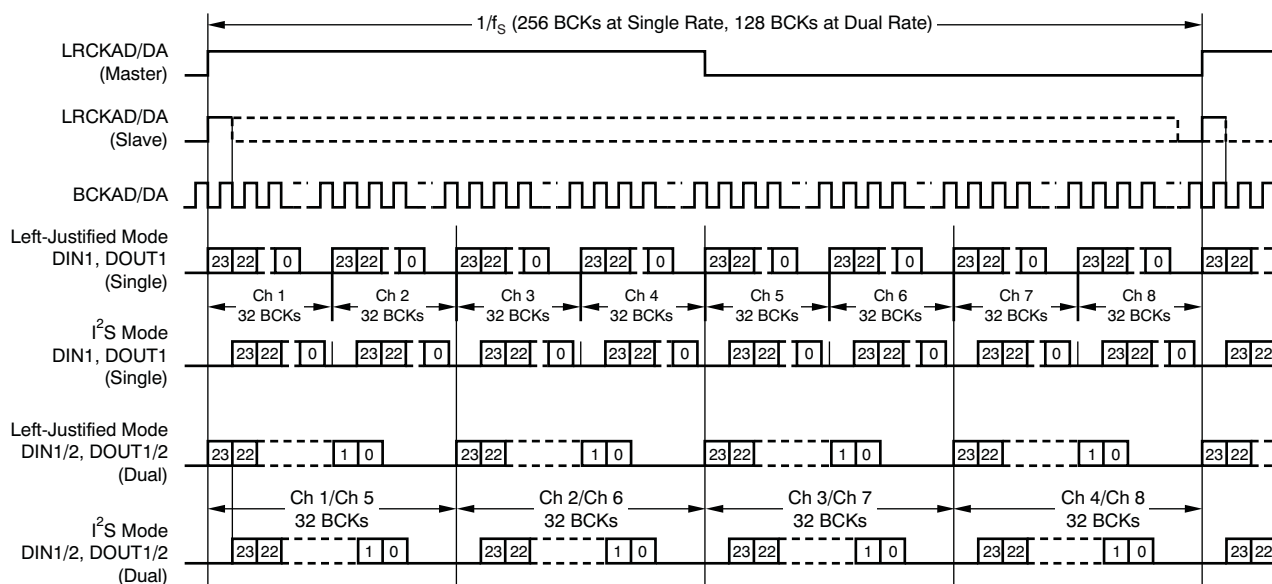


Figure 40. Audio Data Format: 24-Bit TDM Format (SCKI = 128 f_s , 256 f_s , and 512 f_s Only)

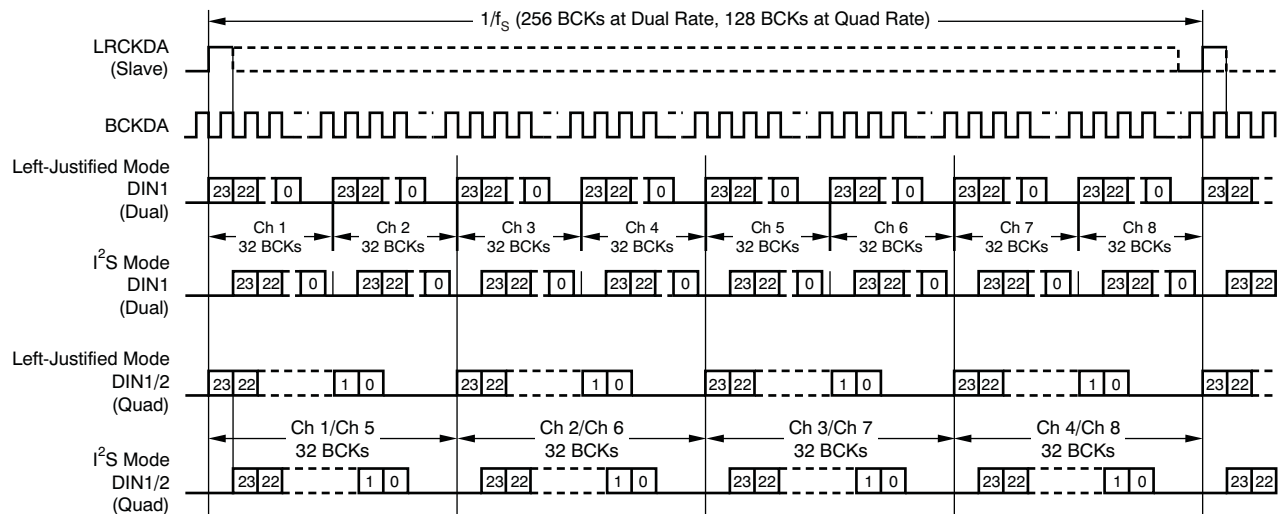


Figure 41. Audio Data Format: 24-Bit High-Speed TDM Format (SCKI = 128 f_s, 256 f_s, DAC, and Slave Mode Only)

AUDIO INTERFACE TIMING

Figure 42 through Figure 45 describe the detailed interface timing specifications.

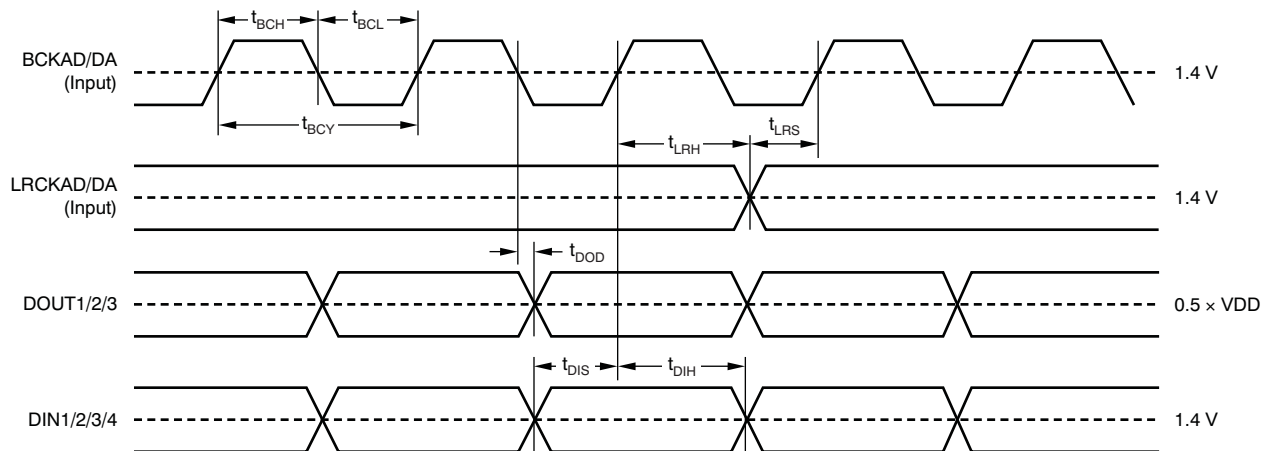


Figure 42. Audio Interface Timing Requirements for Left-Justified, Right-Justified, and I²S Data Formats (Slave Mode)

Table 9. Timing Requirements for Figure 42⁽¹⁾

| SYMBOL | DESCRIPTION | MIN | TYP | MAX | UNIT |
|------------------|--|-----|-----|-----|------|
| t _{BCY} | BCKAD/DA cycle time | 75 | | | ns |
| t _{BCH} | BCKAD/DA pulse width high | 35 | | | ns |
| t _{BCL} | BCKAD/DA pulse width low | 35 | | | ns |
| t _{LRS} | LRCKAD/DA setup time to BCKAD/DA rising edge | 10 | | | ns |
| t _{LRH} | LRCKAD/DA hold time to BCKAD/DA rising edge | 10 | | | ns |
| t _{DIS} | DIN1/2/3/4 setup time to BCKDA rising edge | 10 | | | ns |
| t _{DIH} | DIN1/2/3/4 hold time to BCKDA rising edge | 10 | | | ns |
| t _{DOD} | DOUT1/2/3 delay time from BCKAD falling edge | 0 | | 30 | ns |

(1) Load capacitance of output is 20 pF.

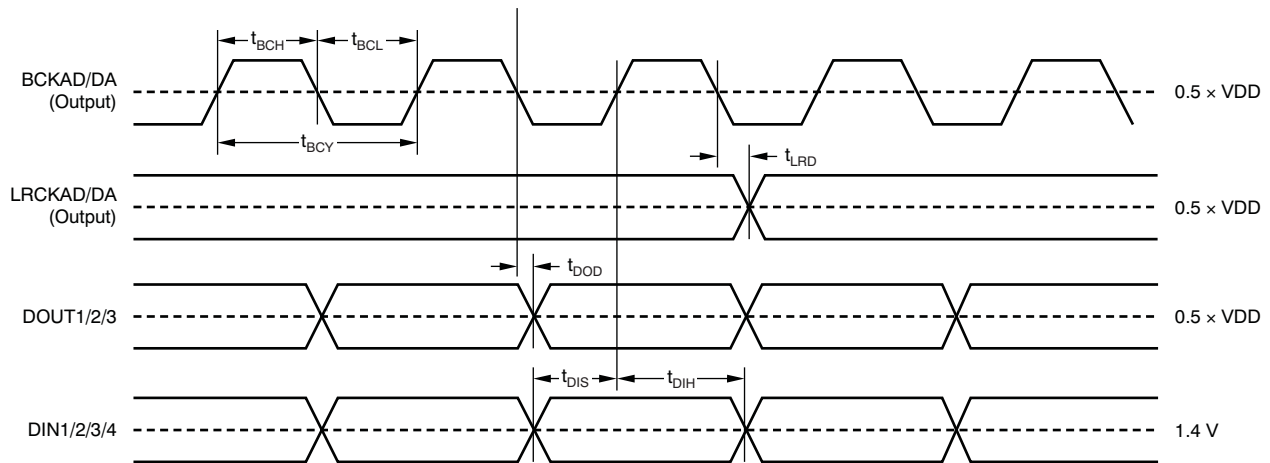


Figure 43. Audio Interface Timing Requirements for Left-Justified, Right-Justified, and I²S Data Formats (Master Mode)

Timing Requirements for Figure 43⁽¹⁾

| SYMBOL | DESCRIPTION | MIN | TYP | MAX | UNIT |
|-----------|---|---------------|------------------------|---------------|------|
| t_{BCY} | BCKAD/DA cycle time | | 1/(64 f _s) | | |
| t_{BCH} | BCKAD/DA pulse width high | 0.4 t_{BCY} | 0.5 t_{BCY} | 0.6 t_{BCY} | |
| t_{BCL} | BCKAD/DA pulse width low | 0.4 t_{BCY} | 0.5 t_{BCY} | 0.6 t_{BCY} | |
| t_{LRD} | LRCKAD/DA delay time from BCKAD/DA falling edge | -10 | | 20 | ns |
| t_{DIS} | DIN1/2/3/4 setup time to BCKDA rising edge | 10 | | | ns |
| t_{DIH} | DIN1/2/3/4 hold time to BCKDA rising edge | 10 | | | ns |
| t_{DOD} | DOUT1/2/3 delay time from BCKAD falling edge | -10 | | 20 | ns |

(1) Load capacitance of output is 20 pF.

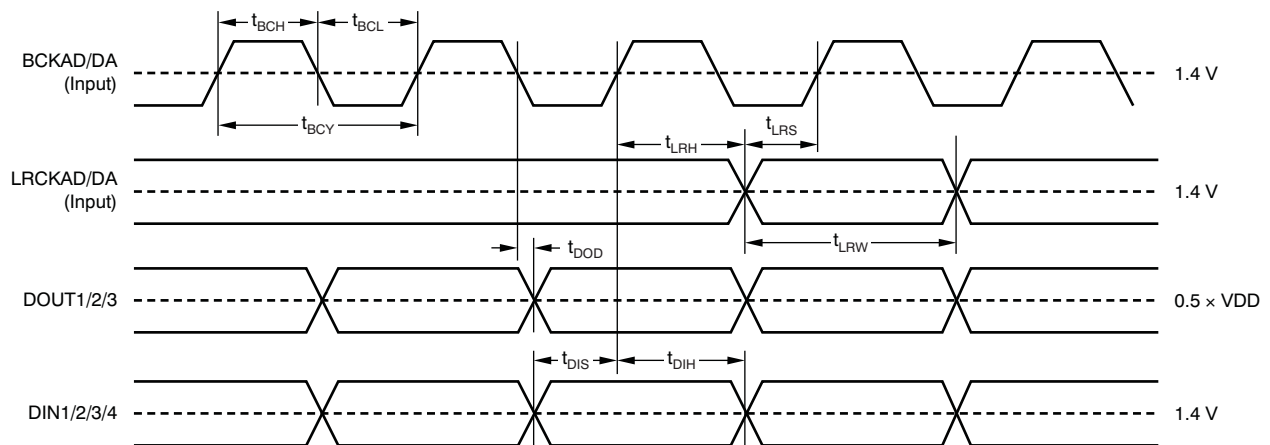


Figure 44. Audio Interface Timing Requirements for DSP and TDM Data Formats (Slave Mode)

Timing Requirements for Figure 44⁽¹⁾

| SYMBOL | DESCRIPTION | MIN | TYP | MAX | UNIT |
|-----------|---|-----------|-----|-------------------|------|
| t_{BCY} | BCKAD cycle time | 75 | | | ns |
| | BCKDA cycle time | 40 | | | ns |
| t_{BCH} | BCKAD pulse width high | 35 | | | ns |
| | BCKDA pulse width high | 15 | | | ns |
| t_{BCL} | BCKAD pulse width low | 35 | | | ns |
| | BCKDA pulse width low | 15 | | | ns |
| t_{LRW} | LRCKAD/DA pulse width high (DSP format) | t_{BCY} | | | |
| | LRCKAD/DA pulse width high (TDM format) | t_{BCY} | | $1/f_S - t_{BCY}$ | |
| t_{LRS} | LRCKAD/DA setup time to BCKAD/DA rising edge | 10 | | | ns |
| t_{LRH} | LRCKAD/DA hold time to BCKAD/DA rising edge | 10 | | | ns |
| t_{DIS} | DIN1/2/3/4 setup time to BCKDA rising edge | 10 | | | ns |
| t_{DIH} | DIN1/2/3/4 hold time to BCKDA rising edge | 10 | | | ns |
| t_{DOD} | DOOUT1/2/3 delay time from BCKAD falling edge | 0 | | 30 | ns |

(1) Load capacitance of output is 20 pF.

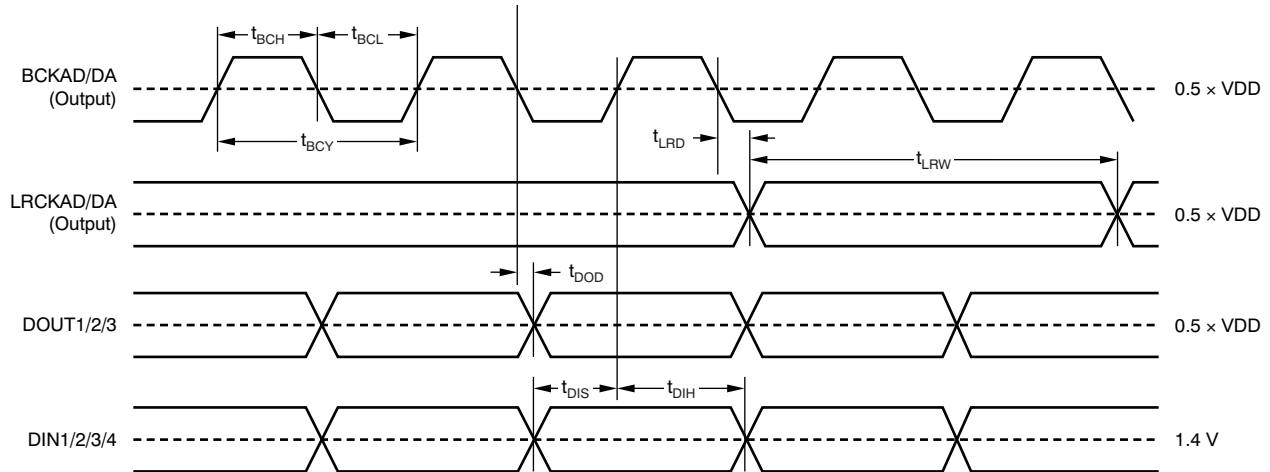


Figure 45. Audio Interface Timing Requirements for DSP and TDM Data Formats (Master Mode)

Timing Requirements for Figure 45⁽¹⁾

| SYMBOL | DESCRIPTION | MIN | TYP | MAX | UNIT |
|-----------|---|---------------|---------------|---------------|------|
| t_{BCY} | BCKAD/DA cycle time (DSP format) | | $1/(64 f_S)$ | | |
| | BCKAD/DA cycle time (TDM format, single rate) | | $1/(256 f_S)$ | | |
| | BCKAD/DA cycle time (TDM format, dual rate) | | $1/(128 f_S)$ | | |
| t_{BCH} | BCKAD/DA pulse width high | $0.4 t_{BCY}$ | $0.5 t_{BCY}$ | $0.6 t_{BCY}$ | |
| t_{BCL} | BCKAD/DA pulse width low | $0.4 t_{BCY}$ | $0.5 t_{BCY}$ | $0.6 t_{BCY}$ | |
| t_{LRW} | LRCKAD/DA pulse width high (DSP format) | | t_{BCY} | | |
| | LRCKAD/DA pulse width high (TDM format) | | $1/(2 f_S)$ | | |
| t_{LRD} | LRCKAD/DA delay time from BCKAD/DA falling edge | -10 | | 20 | ns |
| t_{DIS} | DIN1/2/3/4 setup time to BCKDA rising edge | 10 | | | ns |
| t_{DIH} | DIN1/2/3/4 hold time to BCKDA rising edge | 10 | | | ns |
| t_{DOD} | DOOUT1/2/3 delay time from BCKAD falling edge | -10 | | 20 | ns |

(1) Load capacitance of output is 20 pF.

SYNCHRONIZATION WITH THE DIGITAL AUDIO SYSTEM

The PCM3168A and PCM3168A-Q1 operate under the system clock (SCKI) and the audio sampling rate (LRCKAD/DA). Therefore, SCKI and LRCKAD/DA must have a specific relationship in slave mode. The PCM3168A and PCM3168A-Q1 do not need a specific phase relationship between the audio interface clocks (LRCKAD/DA, BCKAD/DA) and the system clock (SCKI), but does require a specific frequency relationship (ratiometric) between LRCKAD/DA, BCKAD/DA, and SCKI.

If the relationship between SCKI and LRCKDA changes more than ± 2 BCKDA clocks because of jitter, sampling frequency change, etc., the DAC internal operation halts within $1/f_s$, and the analog output is forced into VCOMDA ($0.5 V_{CCDA1}$) until re-synchronization between SCKI, LRCKDA, and BCKDA is completed and then $t_{DACDLY3}$ passes. If the relationship between SCKI and LRCKAD changes more than ± 2 BCKADs because of jitter, sampling frequency change, etc., the ADC internal operation halts within $1/f_s$, and the digital output is forced into a '0' code until re-synchronization between SCKI, LRCKAD, and BCKAD is completed and then $t_{ADC DLY3}$ passes. In the event the change is less than ± 2 BCKAD/DAs, re-synchronization does not occur, and this analog/digital output control and discontinuity do not occur.

Figure 46 shows the DAC analog output and ADC digital output for loss of synchronization. During undefined data periods, some noise may be generated in the audio signal. Also, the transition of normal to undefined data and undefined (or zero) data to normal data creates a discontinuity of data on the analog and digital outputs, which then may generate some noise in the audio signal.

Both ADC outputs (DOUTx) and DAC outputs (VOUTx) hold the previous state if the system clock halts, but the asynchronous and re-synchronization processes would occur after the system clock resumes. Figure 46 shows DAC outputs and ADC outputs for loss of synchronization.

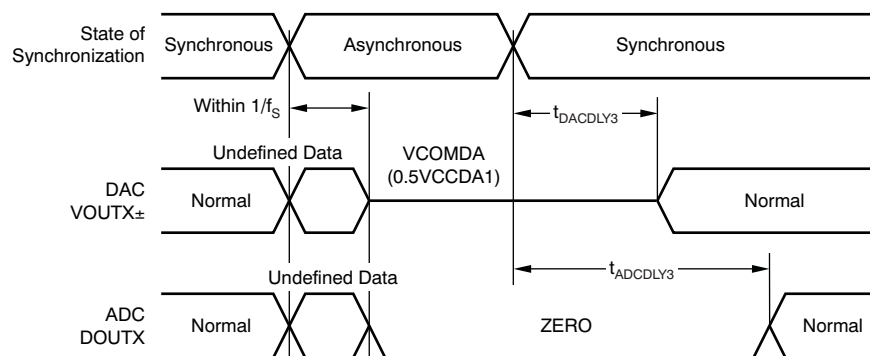


Figure 46. DAC Outputs and ADC Outputs for Loss of Synchronization

Timing Requirements for Figure 46

| SYMBOL | DESCRIPTION | SINGLE | DUAL | QUAD | UNIT |
|----------------|---|--------|------|------|------------------|
| $t_{DACDLY3}$ | DAC delay synchronization detect to normal data | 38 | 38 | 29 | Period of LRCKDA |
| $t_{ADC DLY3}$ | ADC delay synchronization detect to normal data | 60 | 60 | N/A | Period of LRCKAD |

HIGH-PASS FILTER (HPF)

The PCM3168A and PCM3168A-Q1 include a high-pass filter (HPF) for all ADC channels in order to remove the dc component of the digitized input signal. The filter is located at the output of the digital decimation filter. The -3 dB corner frequency for the HPF scales with the output sampling rate, where $f_{-3\text{ dB}} = 0.020 \times f_s/1000$. When $f_s = 48$ kHz, $f_{-3\text{ dB}}$ is 0.96 Hz. The HPF function can be disabled (bypassed) by the BYP bits in two channels.

OVERFLOW FLAG

The PCM3168A and PCM3168A-Q1 include an overflow flag output for all ADC channels. As soon as any of the six-channel ADC digital outputs exceed the full-scale range, an overflow flag is forced high on the OVF pin. The overflow flag is held high for 1024 LRCKAD clock cycles. In parallel, overflow flag information is stored in the OVF bits of the mode control register, and the OVF bit is held until the mode control register is read. The overflow flag polarity can be changed by the OVFP bit. The OVF pin also indicates internal reset completion by transmitting a 4096 SCKI width pulse.

ZERO FLAG

The PCM3168A and PCM3168A-Q1 include a zero flag output for all DAC channels. When all of the eight-channel DACs digital inputs have continued as zero data for 1024 LRCKDA clock cycles, the zero flag is forced high on ZERO. In parallel, zero flag information is stored in the ZERO bits according to channel. The zero flag polarity can be changed by the ZREV bit. Also, the zero flag function can be selected by the AZRO bits. AND or OR logic for stereo, six channels, and eight channels can be selected.

MODE CONTROL

The PCM3168A and PCM3168A-Q1 include four-way mode control selectable by MODE pin, as shown in [Table 10](#). The pull-up and pull-down resistors must be 220 kΩ ±5%. This mode control selection is sampled only when the internal reset is released by a power-on reset or by a low-to-high transition of the external reset (RST pin); a system clock is also required.

Table 10. Mode Control Selection

| MODE | MODE CONTROL INTERFACE |
|-------------------------------------|---|
| Tied to DGND | Two-wire (I ² C) serial control, selectable analog input configuration |
| Tied to DGND via pull-down resistor | H/W (hardware control), differential analog input |
| Tied to VDD via pull-up resistor | H/W (hardware control), single-ended analog input |
| Tied to VDD | Four-wire (SPI) serial control, selectable analog input configuration |

From the mode control selection described in [Table 10](#), the functions of four pins are changed, as shown in [Table 11](#).

Table 11. Pin Functions

| PIN | PIN ASSIGNMENTS | | |
|--------------|-----------------|------------------|------|
| | SPI | I ² C | H/W |
| MS/ADR0/MD0 | MS | ADR0 | MD0 |
| MDO/ADR1/MD1 | MDO | ADR1 | MD1 |
| MDI/SDA/DEMP | MDI | SDA | DEMP |
| MC/SCL/FMT | MC | SCL | FMT |

Both serial controls are available while RST = high and after internal reset completion, which is indicated as a negative transition (high ≥ low) of a 4096 × SCKI width pulse on the OVF pin.

HARDWARE CONTROL MODE CONFIGURATION

The data format is selected by the MC/SCL/FMT pin between I²S format and I²S mode in TDM format, as shown in [Table 12](#).

Table 12. Data Format Selection

| FMT | MODE CONTROL INTERFACE |
|------|---|
| Low | I ² S audio data format |
| High | I ² S mode, TDM audio data format (supported only for SCKI = 128 f _S , 256 f _S , or 512 f _S) |

The de-emphasis filter is enabled by the MDI/SDA/DEMP pin. The de-emphasis frequency is fixed at 44.1 kHz in hardware control mode, as shown in Table 13. The software mode provides full selections of 32 kHz, 44.1 kHz, and 48 kHz.

Table 13. Hardware Control Mode

| DEMP (DE-EMPHASIS FILTER ENABLE) | DESCRIPTION |
|----------------------------------|--------------------------------|
| Low | 44.1 kHz, de-emphasis disabled |
| High | 44.1 kHz, de-emphasis enabled |

The audio interface and the sampling mode are selected by the MS/ADR0/MD0 and MDO/ADR1/MD1 pins. The selectable multiple of the master mode audio interface is limited between 256 f_S, 384 f_S, and 512 f_S; the selectable sampling mode is limited as shown in Table 14. The software mode provides full selections.

Table 14. Selectable Sampling Mode

| MD1 | MD0 | DESCRIPTION | | | |
|------|------|----------------------------|----------------------|---------------------|---------------------|
| | | INTERFACE MODE | | SAMPLING MODE | |
| | | ADC | DAC | ADC | DAC |
| Low | Low | Slave ⁽¹⁾ | Slave ⁽¹⁾ | Auto ⁽²⁾ | Auto ⁽²⁾ |
| Low | High | Master, 512 f _S | Slave ⁽¹⁾ | Single rate | Auto ⁽²⁾ |
| High | Low | Master, 384 f _S | Slave ⁽¹⁾ | Dual rate | Auto ⁽²⁾ |
| High | High | Master, 256 f _S | Slave ⁽¹⁾ | Dual rate | Auto ⁽²⁾ |

- (1) The multiples between system clock and sampling frequency are automatically detected; 256 f_S, 384 f_S, 512 f_S, and 768 f_S are acceptable for ADC operation, and 128 f_S, 192 f_S, 256 f_S, 384 f_S, 512 f_S, and 768 f_S are acceptable for DAC operation.
- (2) The sampling mode is automatically set as single rate for 512 f_S and 768 f_S, dual rate for 256 f_S and 384 f_S, and quad rate for 128 f_S and 198 f_S, according to the detected multiples between the system clock and sampling clock.

FOUR-WIRE (SPI) SERIAL CONTROL

The PCM3168A and PCM3168A-Q1 include an SPI-compatible serial port that operates asynchronously with the audio serial interface. The control interface consists of MDI/SDA/DEMP, MDO/ADR1/MD1, MC/SCL/FMT, and MS/ADR0/MD0. MDI is the serial data input to program the mode control registers. MDO is the serial data output to read back register settings and some flags. MDO is inactive (Hi-Z, high impedance) during MS = high. MC is the serial bit clock that shifts the data into the control port. MS is the select input to enable the mode control port.

CONTROL DATA WORD FORMAT

All single write/read operations via the serial control port use 16-bit data words. Figure 47 shows the control data word format. The first bit is for read/write controls; '0' indicates a write operation and '1' indicates a read operation. Following the first bit are seven other bits, labeled ADR[6:0] that set the register address for the write/read operation. The eight least significant bits (LSBs), D[7:0] on MDI or MDO, contain the data to be written to the register specified by ADR[6:0], or the data read from the register specified by ADR[6:0].

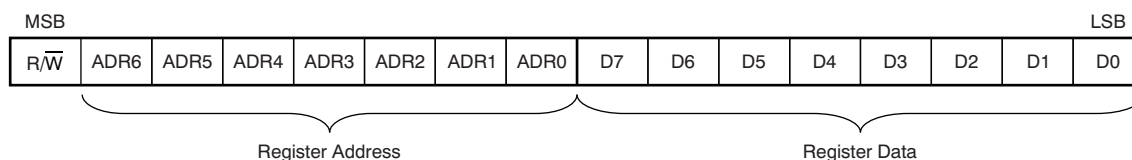
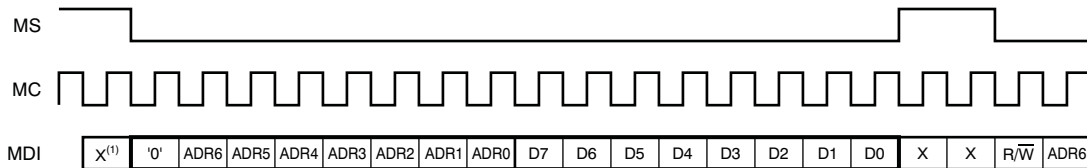


Figure 47. Control Data Word Format for MDI

REGISTER WRITE OPERATION

Figure 48 shows the functional timing diagram for single write operations on the serial control port. MS is held at a high state until a register must be written. To start the register write cycle, MS is set to a low state. 16 clocks are then provided on MC, corresponding to the 16 bits of the control data word on MDI. After the 16th clock cycle has been completed, MS is set high to latch the data into the indexed mode control register.

Also, the PCM3168A and PCM3168A-Q1 support multiple write operations in addition to single write operations, which can be performed by sending the following N-times of the 8-bit register data after the first 16-bit register address and register data while keeping the MC clocks and MS at a low state. Closing a multiple write operation can be accomplished by setting MS to a high state.

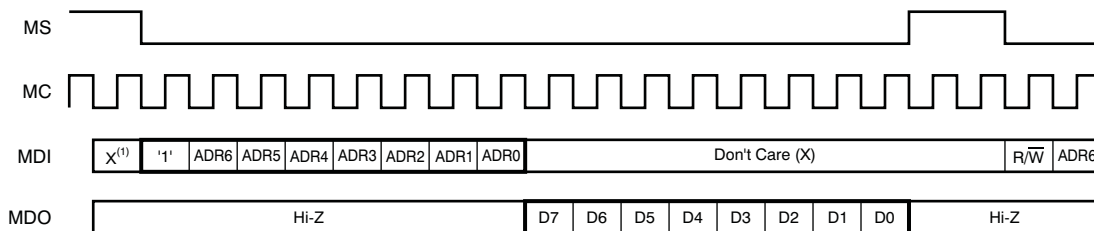


(1) X = Don't care.

Figure 48. Register Write Operation

REGISTER READ OPERATION

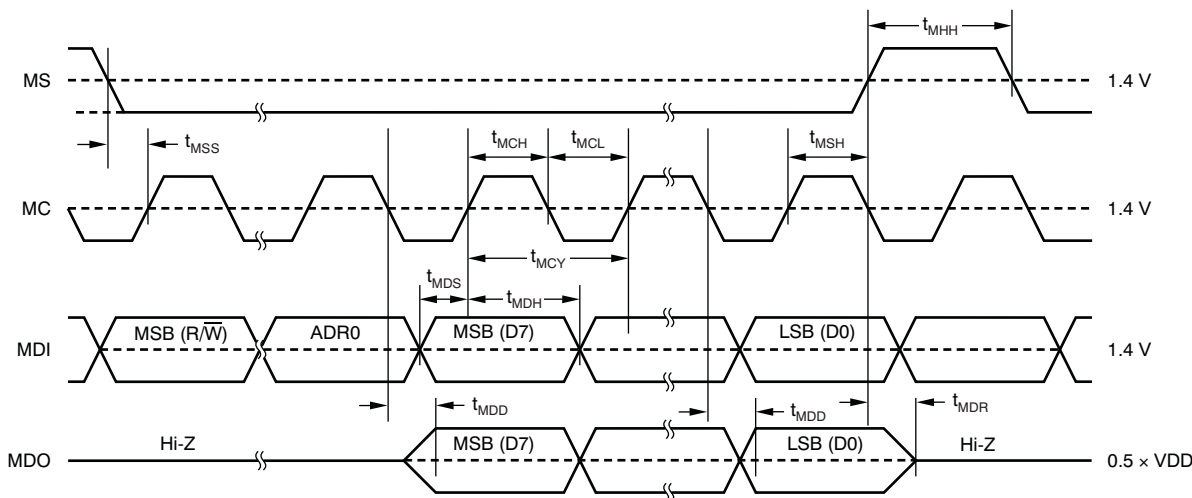
Figure 49 shows the functional timing diagram for single read operations on the serial control port. MS is held at a high state until a register must be read. To start the register read cycle, MS is set to a low state. 16 clocks are then provided on MC, corresponding to the first eight bits of the control data word on MDI and the second eight bits of the read-back data word from MDO. After the 16th clock cycle has been completed, MS is held high for the next write or read operation. MDO remains in a high impedance state except during the eight MC clock periods of the actual data transfer.



(1) X = Don't care.

Figure 49. Register Read Operation

TIMING CHARACTERISTICS: FOUR-WIRE



(1) These timing parameters are critical for proper control port operation.

Figure 50. Four-Wire Serial Control Interface Timing⁽¹⁾

Timing Requirements for Figure 50⁽¹⁾

| SYMBOL | PARAMETER | PCM3168A, PCM3168A-Q1 | | UNIT |
|------------------|---|-----------------------|-----|------|
| | | MIN | MAX | |
| t _{MCY} | MC pulse cycle time | 100 | | ns |
| t _{MCL} | MC low-level time | 40 | | ns |
| t _{MCH} | MC high-level time | 40 | | ns |
| t _{MHH} | MS high-level time | t _{MCY} | | ns |
| t _{MSS} | MS falling edge to MC rising edge | 30 | | ns |
| t _{MSH} | MS rising edge from MC rising edge for LSB | 15 | | ns |
| t _{MDH} | MDI hold time | 15 | | ns |
| t _{MDS} | MDI setup time | 15 | | ns |
| t _{MDD} | MDO enable or delay time from MC falling edge | 0 | 30 | ns |
| t _{MDR} | MDO disable time from MS rising edge | 0 | 30 | ns |

(1) These timing parameters are critical for proper control port operation.

TWO-WIRE (I²C) SERIAL CONTROL

The PCM3168A and PCM3168A-Q1 support an I²C-compatible serial bus and data transmission protocol for fast mode configured as a slave device. This protocol is explained in the I²C specification, version 2.0.

The PCM3168A and PCM3168A-Q1 have a 7-bit slave address, as shown in Figure 51. The first five bits are the most significant bits (MSB) of the slave address and are factory-preset to 10001. The next two bits of the address byte are selectable bits that can be set by MS/ADR0/MDO and MDO/ADR1/MD1. A maximum of four PCM3168A and PCM3168A-Q1s can be connected on the same bus at any one time. Each PCM3168A and PCM3168A-Q1 respond when it receives its own slave address.

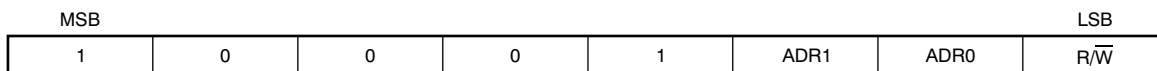
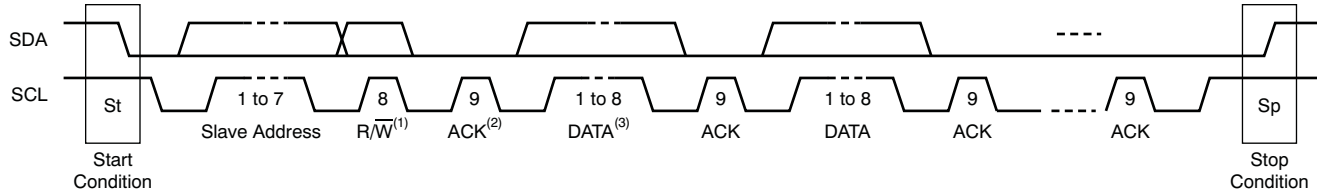


Figure 51. Slave Address

PACKET PROTOCOL

A master device must control the packet protocol, which consists of the start condition, slave address with the read/write bit, data if a write operation is required, acknowledgement if a read operation is required, and stop condition. The PCM3168A and PCM3168A-Q1 support both slave receiver and transmitter functions. Details about DATA for both write and read operations are described in [Figure 52](#).



- (1) R/\bar{W} : Read operation if '1'; write operation otherwise.
- (2) ACK: Acknowledgement of a byte if '0', not Acknowledgement of a byte if '1'.
- (3) DATA: Eight bits (byte); details are described in the [Write Operation](#) and [Read Operation](#) sections.

Figure 52. DATA Operation

WRITE OPERATION

The PCM3168A and PCM3168A-Q1 support a receiver function. A master device can write to any PCM3168A and PCM3168A-Q1 register using single or multiple accesses. The master sends a PCM3168A and PCM3168A-Q1 slave address with a write bit, a register address, and the data. If multiple access is required, the address is that of the starting register, followed by the data to be transferred. When the data are received properly, the index register is incremented by one automatically. When the index register reaches ^h5E, the next value is ^h40. When undefined registers are accessed, the PCM3168A and PCM3168A-Q1 do not send an acknowledgement. [Figure 53](#) illustrates a diagram of the write operation. The register address and write data are in 8-bit, MSB-first format.

| | | | | | | | | | | | | | |
|-------------|----|---------------|-----------|-----|-------------|-----|--------------|-----|--------------|-----|-------|-----|----|
| Transmitter | M | M | M | S | M | S | M | S | M | S | | S | M |
| Data Type | St | Slave Address | \bar{W} | ACK | Reg Address | ACK | Write Data 1 | ACK | Write Data 2 | ACK | | ACK | Sp |

- (1) M = Master device, S = Slave device, St = Start condition, \bar{W} = Write, ACK = Acknowledge, and Sp = Stop condition.

Figure 53. Framework for Write Operation

READ OPERATION

A master device can read the registers from ^h40 to ^h5E of the PCM3168A and PCM3168A-Q1. The value of the register address is stored in an indirect index register in advance. The master sends the PCM3168A and PCM3168A-Q1 slave address with a read bit after storing the register address. Then the PCM3168A and PCM3168A-Q1 transfer the data that the index register points to. [Figure 54](#) shows a diagram of the read operation.

| | | | | | | | | | | | | | |
|-------------|----|---------------|-----------|-----|-------------|-----|----|---------------|---|-----|-----------|------|----|
| Transmitter | M | M | M | S | M | S | M | M | M | S | S | M | M |
| Data Type | St | Slave Address | \bar{W} | ACK | Reg Address | ACK | Sr | Slave Address | R | ACK | Read Data | NACK | Sp |

- (1) M = Master device, S = Slave device, St = Start condition, Sr = Repeated start condition, \bar{W} = Write, R = Read, ACK = Acknowledge, NACK = Not acknowledge, and Sp = Stop condition.

NOTE: The slave address after the repeated start condition must be the same as the previous address.

Figure 54. Framework for Read Operation

TIMING REQUIREMENTS: SCL AND SDA

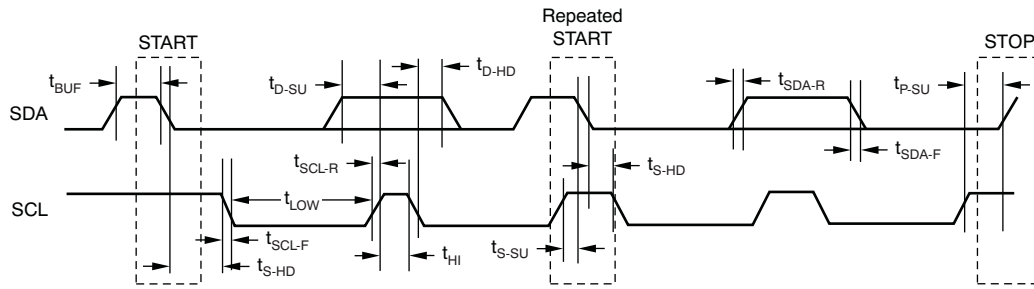


Figure 55. SCL and SDA Control Interface Timing

Timing Requirements for Figure 55

| SYMBOL | PARAMETER | PCM3168A, PCM3168A-Q1 | | | | UNIT |
|-------------|---|-----------------------|------|----------------------|-----|---------|
| | | STANDARD MODE | | FAST MODE | | |
| | | MIN | MAX | MIN | MAX | |
| f_{SCL} | SCL clock frequency | | 100 | | 400 | kHz |
| t_{BUF} | Bus free time between STOP and START condition | 4.7 | | 1.3 | | μ s |
| t_{LOW} | Low period of the SCL clock | 4.7 | | 1.3 | | μ s |
| t_{HI} | High period of the SCL clock | 4.0 | | 0.6 | | μ s |
| t_{S-SU} | Setup time for START/Repeated START condition | 4.7 | | 0.6 | | μ s |
| t_{S-HD} | Hold time for START/Repeated START condition | 4.0 | | 0.6 | | μ s |
| t_{D-SU} | Data setup time | 250 | | 100 | | ns |
| t_{D-HD} | Data hold time | 0 | 3450 | 0 | 900 | ns |
| t_{SCL-R} | Rise time of SCL signal | | 1000 | $20 + 0.1 C_B$ | 300 | ns |
| t_{SCL-F} | Fall time of SCL signal | | 1000 | $20 + 0.1 C_B$ | 300 | ns |
| t_{SDA-R} | Rise time of SDA signal | | 1000 | $20 + 0.1 C_B$ | 300 | ns |
| t_{SDA-F} | Fall time of SDA signal | | 1000 | $20 + 0.1 C_B$ | 300 | ns |
| t_{P-SU} | Setup time for STOP condition | 4.0 | | 0.6 | | μ s |
| t_{GW} | Allowable glitch width | | N/A | | 50 | |
| C_B | Capacitive load for SDA and SCL line | | 400 | | 100 | pF |
| V_{NH} | Noise margin at high level for each connected device (including hysteresis) | $0.2 \times V_{DD}$ | | $0.2 \times V_{DD}$ | | V |
| V_{NL} | Noise margin at low level for each connected device (including hysteresis) | $0.1 \times V_{DD}$ | | $0.1 \times V_{DD}$ | | V |
| V_{HYS} | Hysteresis of Schmitt-trigger input | N/A | | $0.05 \times V_{DD}$ | | V |

CONTROL REGISTER DEFINITIONS (SOFTWARE MODE ONLY)

The PCM3168A and PCM3168A-Q1 have many user-programmable functions that are accessed via control registers, and are programmed through the SPI or I²C serial control port. Table 15 shows the available mode control functions along with reset default conditions and associated register address. Table 16 lists the register map.

Table 15. User-Programmable Mode Control Functions

| FUNCTION | RESET DEFAULT | REGISTER | LABEL |
|---|------------------------------|----------|------------|
| Mode control register reset for ADC and DAC operation | Normal operation | 64 | MRST |
| System reset for ADC and DAC operation | Normal operation | 64 | SRST |
| DAC sampling mode selection | Auto | 64 | SRDA[1:0] |
| DAC power-save mode selection | Power save | 65 | PSMDA |
| DAC master/slave mode selection | Slave | 65 | MSDA[2:0] |
| DAC audio interface format selection | I ² S | 65 | FMTDA[3:0] |
| DAC operation control | Normal operation | 66 | OPEDA[3:0] |
| DAC digital filter roll-off control | Sharp roll-off | 66 | FLT[3:0] |
| DAC output phase selection | Normal | 67 | REVDA[8:1] |
| DAC soft mute control | Mute disabled | 68 | MUTDA[8:1] |
| DAC zero flag | Not detected | 69 | ZERO[8:1] |
| DAC digital attenuation mode | Channel independent | 70 | ATMDDA |
| DAC digital attenuation speed | $N \times 2048/f_S$ | 70 | ATSPDA |
| DAC digital de-emphasis function control | Disabled | 70 | DEMP[1:0] |
| DAC zero flag function selection | Independent | 70 | AZRO[2:0] |
| DAC zero flag polarity selection | High for detection | 70 | ZREV |
| DAC digital attenuation level shifting | 0 dB, no attenuation | 71–79 | ATDAx[7:0] |
| ADC sampling mode selection | Auto | 80 | SRAD[1:0] |
| ADC master/slave mode selection | Slave | 81 | MSAD[2:0] |
| ADC audio interface format selection | I ² S | 81 | FMTAD[2:0] |
| ADC power-save control | Normal operation | 82 | PSVAD[2:0] |
| ADC HPF bypass control | Normal output, HPF enabled | 82 | BYP[2:0] |
| ADC input configuration control | Differential | 83 | SEAD[6:1] |
| ADC input phase selection | Normal | 84 | REVAD[6:1] |
| ADC soft mute control | Mute disabled | 85 | MUTAD[6:1] |
| ADC overflow flag | Not detected | 86 | OVF[6:1] |
| ADC digital attenuation mode | Channel independent | 87 | ATMDAD |
| ADC digital attenuation speed | $N \times 2048/f_S$ | 87 | ATSPAD |
| ADC overflow flag polarity selection | High for detection | 87 | OVFP |
| ADC digital attenuation level setting | 0 dB, no gain or attenuation | 88–94 | ATADx[7:0] |

Table 16. Register Map

| ADDRESS | | DATA | | | | | | | |
|---------|-----|--------|--------|--------|--------|--------|--------|--------|--------|
| DAC | HEX | B7 | B6 | B5 | B4 | B3 | B2 | B1 | B0 |
| 64 | 40 | MRST | SRST | — | — | — | — | SRDA1 | SRDA0 |
| 65 | 41 | PSMDA | MSDA2 | MSDA1 | MSDA0 | FMTDA3 | FMTDA2 | FMTDA1 | FMTDA0 |
| 66 | 42 | OPEDA3 | OPEDA2 | OPEDA1 | OPEDA0 | FLT3 | FLT2 | FLT1 | FLT0 |
| 67 | 43 | REVDA8 | REVDA7 | REVDA6 | REVDA5 | REVDA4 | REVDA3 | REVDA2 | REVDA1 |
| 68 | 44 | MUTDA8 | MUTDA7 | MUTDA6 | MUTDA5 | MUTDA4 | MUTDA3 | MUTDA2 | MUTDA1 |
| 69 | 45 | ZERO8 | ZERO7 | ZERO6 | ZERO5 | ZERO4 | ZERO3 | ZERO2 | ZERO1 |
| 70 | 46 | ATMDDA | ATSPDA | DEMP1 | DEMP0 | AZRO2 | AZRO1 | AZRO0 | ZREV |
| 71 | 47 | ATDA07 | ATDA06 | ATDA05 | ATDA04 | ATDA03 | ATDA02 | ATDA01 | ATDA00 |
| 72 | 48 | ATDA17 | ATDA16 | ATDA15 | ATDA14 | ATDA13 | ATDA12 | ATDA11 | ATDA10 |
| 73 | 49 | ATDA27 | ATDA26 | ATDA25 | ATDA24 | ATDA23 | ATDA22 | ATDA21 | ATDA20 |
| 74 | 4A | ATDA37 | ATDA36 | ATDA35 | ATDA34 | ATDA33 | ATDA32 | ATDA31 | ATDA30 |
| 75 | 4B | ATDA47 | ATDA46 | ATDA45 | ATDA44 | ATDA43 | ATDA42 | ATDA41 | ATDA40 |
| 76 | 4C | ATDA57 | ATDA56 | ATDA55 | ATDA54 | ATDA53 | ATDA52 | ATDA51 | ATDA50 |
| 77 | 4D | ATDA67 | ATDA66 | ATDA65 | ATDA64 | ATDA63 | ATDA62 | ATDA61 | ATDA60 |
| 78 | 4E | ATDA77 | ATDA76 | ATDA75 | ATDA74 | ATDA73 | ATDA72 | ATDA71 | ATDA70 |
| 79 | 4F | ATDA87 | ATDA86 | ATDA85 | ATDA84 | ATDA83 | ATDA82 | ATDA81 | ATDA80 |
| 80 | 50 | — | — | — | — | — | — | SRAD1 | SRAD0 |
| 81 | 51 | — | MSAD2 | MSAD1 | MSAD0 | — | FMTAD2 | FMTAD1 | FMTAD0 |
| 82 | 52 | — | PSVAD2 | PSVAD1 | PSVAD0 | — | BYP2 | BYP1 | BYP0 |
| 83 | 53 | — | — | SEAD6 | SEAD5 | SEAD4 | SEAD3 | SEAD2 | SEAD1 |
| 84 | 54 | — | — | REVAD6 | REVAD5 | REVAD4 | REVAD3 | REVAD2 | REVAD1 |
| 85 | 55 | — | — | MUTAD6 | MUTAD5 | MUTAD4 | MUTAD3 | MUTAD2 | MUTAD1 |
| 86 | 56 | — | — | OVF6 | OVF5 | OVF4 | OVF3 | OVF2 | OVF1 |
| 87 | 57 | ATMDAD | ATSPAD | — | — | — | — | — | OVFP |
| 88 | 58 | ATAD07 | ATAD06 | ATAD05 | ATAD04 | ATAD03 | ATAD02 | ATAD01 | ATAD00 |
| 89 | 59 | ATAD17 | ATAD16 | ATAD15 | ATAD14 | ATAD13 | ATAD12 | ATAD11 | ATAD10 |
| 90 | 5A | ATAD27 | ATAD26 | ATAD25 | ATAD24 | ATAD23 | ATAD22 | ATAD21 | ATAD20 |
| 91 | 5B | ATAD37 | ATAD36 | ATAD35 | ATAD34 | ATAD33 | ATAD32 | ATAD31 | ATAD30 |
| 92 | 5C | ATAD47 | ATAD46 | ATAD45 | ATAD44 | ATAD43 | ATAD42 | ATAD41 | ATAD40 |
| 93 | 5D | ATAD57 | ATAD56 | ATAD55 | ATAD54 | ATAD53 | ATAD52 | ATAD51 | ATAD50 |
| 94 | 5E | ATAD67 | ATAD66 | ATAD65 | ATAD64 | ATAD63 | ATAD62 | ATAD61 | ATAD60 |

REGISTER DEFINITIONS

| DEC | HEX | B7 | B6 | B5 | B4 | B3 | B2 | B1 | B0 |
|-----|-----|------|------|----|----|----|----|-------|-------|
| 64 | 40 | MRST | SRST | — | — | — | — | SRDA1 | SRDA0 |

MRST Mode control register reset for the ADC and DAC

This bit sets the mode control register reset to the default value. Pop-noise may be generated. Returning the MRST bit to '1' is unnecessary, because it is automatically set to '1' after the mode control register is reset.

Default value = 1.

| MRST | Mode control register reset |
|------|-----------------------------|
| 0 | Set default value |
| 1 | Normal operation (default) |

SRST System reset for the ADC and DAC

This bit controls system reset, the relation between system clock and sampling clock re-synchronization, and ADC operation and DAC operation restart. The mode control register is not reset and the PCM3168A and PCM3168A-Q1 do not go into a power-down state. The fade-in sequence is supported in the resume process, but pop-noise may be generated. Returning the SRST bit to '1' is unnecessary; it is automatically set to '1' after triggering a system reset.

Default value = 1.

| SRST | System reset |
|------|----------------------------|
| 0 | Resynchronization |
| 1 | Normal operation (default) |

SRDA[1:0] DAC Sampling mode select

These bits control the sampling mode of DAC operation. In Auto mode, the sampling mode is automatically set according to multiples between the system clock and sampling clock, single rate for 512 f_S and 768 f_S, dual rate for 256 f_S or 384 f_S, and quad rate for 128 f_S and 192 f_S.

Default value = 00.

| SRDA | DAC Sampling mode select |
|------|--------------------------|
| 00 | Auto (default) |
| 01 | Single rate |
| 10 | Dual rate |
| 11 | Quad rate |

| DEC | HEX | B7 | B6 | B5 | B4 | B3 | B2 | B1 | B0 |
|-----|-----|-------|-------|-------|-------|--------|--------|--------|--------|
| 65 | 41 | PSMDA | MSDA2 | MSDA1 | MSDA0 | FMTDA3 | FMTDA2 | FMTDA1 | FMTDA0 |

PSMDA DAC Power-save mode select

This bit selects the power-save mode for the OPEDA[3:0] function. OPEDA[3:0] is the control of power-save mode and normal operation for PSMDA = 0, or OPEDA[3:0] works as the control of DAC disable (not power-save mode) and normal operation for PSMDA = 1.

Default value: 0.

PSMDA DAC Power-save mode select

| | |
|---|----------------------------------|
| 0 | Power-save enable mode (default) |
| 1 | Power-save disable mode |

MSDA[2:0] DAC Master/slave mode select

These bits control the audio interface mode for DAC operation.

Default value: 000 (slave mode).

MSDA DAC Master/slave mode select

| | |
|-----|---------------------------------|
| 000 | Slave mode (default) |
| 001 | Master mode, 768 f _S |
| 010 | Master mode, 512 f _S |
| 011 | Master mode, 384 f _S |
| 100 | Master mode, 256 f _S |
| 101 | Master mode, 192 f _S |
| 110 | Master mode, 128 f _S |
| 111 | Reserved |

FMTDA[3:0] DAC Audio interface format select

These bits control the audio interface format for DAC operation. Details of the format, and any related restrictions with the system clock and master/slave mode, are described in the [Audio Data Interface Formats and Timing](#) section.

Default value: 0000 (24-bit I²S format).

FMTDA DAC Audio interface format select

| | |
|------|--|
| 0000 | 24-bit I ² S format (default) |
| 0001 | 24-bit left-justified format |
| 0010 | 24-bit right-justified format |
| 0011 | 16-bit right-justified format |
| 0100 | 24-bit I ² S mode DSP format |
| 0101 | 24-bit left-justified mode DSP format |
| 0110 | 24-bit I ² S mode TDM format |
| 0111 | 24-bit left-justified mode TDM format |
| 1000 | 24-bit high-speed I ² S mode TDM format |
| 1001 | 24-bit high-speed left-justified mode TDM format |
| 101x | Reserved |
| 11xx | Reserved |

| DEC | HEX | B7 | B6 | B5 | B4 | B3 | B2 | B1 | B0 |
|-----|-----|--------|--------|--------|--------|------|------|------|------|
| 66 | 42 | OPEDA3 | OPEDA2 | OPEDA1 | OPEDA0 | FLT3 | FLT2 | FLT1 | FLT0 |

OPEDA[3:0] DAC Operation control

These bits control the DAC operation mode. In operation disable mode, the DAC output is cut off from DIN with a fade-out sequence, and the internal DAC data is reset. DAC output is forced into VCOMDA if PSMDA = 1, or DAC output is forced into AGNDDA and goes into a power-down state if PSMDA = 0. For normal operating mode, a fade-in sequence is applied on the DAC output in resume process. The serial mode control is effective during operation disable mode. A wait time greater than $t_{DACDLY2}$ is required for the status change because of power-save control turning on/off.

Default value: 0000.

OPEDA DAC Operation control

- xxx0 DAC1/2 normal operation
- xxx1 DAC1/2 operation disable with or without power save
- xx0x DAC3/4 normal operation
- xx1x DAC3/4 operation disable with or without power save
- x0xx DAC5/6 normal operation
- x1xx DAC5/6 operation disable with or without power save
- 0xxx DAC7/8 normal operation
- 1xxx DAC7/8 operation disable with or without power save

FLT[3:0] DAC Digital filter roll-off control

The FLT[3:0] bits allow users to select the digital filter roll-off that is best suited to their applications. Sharp and Slow filter roll-off selections are available. The filter responses for these selections are shown in the [Typical Characteristics](#) section of this data sheet.

Default value: 0000.

FLT DAC Digital filter roll-off control

- xxx0 DAC1/2 sharp roll-off
- xxx1 DAC1/2 slow roll-off
- xx0x DAC3/4 sharp roll-off
- xx1x DAC3/4 slow roll-off
- x0xx DAC5/6 sharp roll-off
- x1xx DAC5/6 slow roll-off
- 0xxx DAC7/8 sharp roll-off
- 1xxx DAC7/8 slow roll-off

| DEC | HEX | B7 | B6 | B5 | B4 | B3 | B2 | B1 | B0 |
|-----|-----|--------|--------|--------|--------|--------|--------|--------|--------|
| 67 | 43 | REVDA8 | REVDA7 | REVDA6 | REVDA5 | REVDA4 | REVDA3 | REVDA2 | REVDA1 |

REVDA[8:1] DAC Output phase select

The REVDA[8:1] bits are used to control the phase of DAC analog signal outputs.

Default value: 0000 0000.

| REVDA | DAC Output phase select |
|-----------|-------------------------|
| xxxx xxx0 | DAC1 normal output |
| xxxx xxx1 | DAC1 inverted output |
| xxxx xx0x | DAC2 normal output |
| xxxx xx1x | DAC2 inverted output |
| xxxx x0xx | DAC3 normal output |
| xxxx x1xx | DAC3 inverted output |
| xxxx 0xxx | DAC4 normal output |
| xxxx 1xxx | DAC4 inverted output |
| xxx0 xxxx | DAC5 normal output |
| xxx1 xxxx | DAC5 inverted output |
| xx0x xxxx | DAC6 normal output |
| xx1x xxxx | DAC6 inverted output |
| x0xx xxxx | DAC7 normal output |
| x1xx xxxx | DAC7 inverted output |
| 0xxx xxxx | DAC8 normal output |
| 1xxx xxxx | DAC8 inverted output |

| DEC | HEX | B7 | B6 | B5 | B4 | B3 | B2 | B1 | B0 |
|-----|-----|--------|--------|--------|--------|--------|--------|--------|--------|
| 68 | 44 | MUTDA8 | MUTDA7 | MUTDA6 | MUTDA5 | MUTDA4 | MUTDA3 | MUTDA2 | MUTDA1 |

MUTDA[8:1] DAC Soft Mute control

These bits are used to enable or disable the Soft Mute function for the corresponding DAC outputs, VOUT. The Soft Mute function is incorporated into the digital attenuators. When Mute is disabled (MUTDA[8:1] = 0), the attenuator and DAC operate normally. When Mute is enabled by setting MUTDA[8:1] = 1, the digital attenuator for the corresponding output decreases from the current setting to infinite attenuation with an s-curve response and time set by ATSPDA.

By setting MUTDA[8:1] = 0, the attenuator increases to the last attenuation level with s-curve response in the same manner as it is for decreasing levels. This configuration provides *pop and zipper noise-free* muting of the DAC output.

The Soft Mute control uses the same digital attenuation level resource setting as the DAC. Mute control has priority over the digital attenuation level setting.

Default value: 0000 0000.

MUTDA DAC Soft Mute control

| | |
|-----------|--------------------|
| xxxx xxx0 | DAC1 Mute disabled |
| xxxx xxx1 | DAC1 Mute enabled |
| xxxx xx0x | DAC2 Mute disabled |
| xxxx xx1x | DAC2 Mute enabled |
| xxxx x0xx | DAC3 Mute disabled |
| xxxx x1xx | DAC3 Mute enabled |
| xxxx 0xxx | DAC4 Mute disabled |
| xxxx 1xxx | DAC4 Mute enabled |
| xxx0 xxxx | DAC5 Mute disabled |
| xxx1 xxxx | DAC5 Mute enabled |
| xx0x xxxx | DAC6 Mute disabled |
| xx1x xxxx | DAC6 Mute enabled |
| x0xx xxxx | DAC7 Mute disabled |
| x1xx xxxx | DAC7 Mute enabled |
| 0xxx xxxx | DAC8 Mute disabled |
| 1xxx xxxx | DAC8 Mute enabled |

| DEC | HEX | B7 | B6 | B5 | B4 | B3 | B2 | B1 | B0 |
|-----|-----|-------|-------|-------|-------|-------|-------|-------|-------|
| 69 | 45 | ZERO8 | ZERO7 | ZERO6 | ZERO5 | ZERO4 | ZERO3 | ZERO2 | ZERO1 |

ZERO[8:1] DAC Zero flag (read-only)

These bits indicate the present status of the zero detect circuit for each DAC channel; these bits are read-only.

| ZERO | DAC Zero flag |
|-------------|------------------------------|
| xxxx xxx0 | DAC1 zero input not detected |
| xxxx xxx1 | DAC1 zero input detected |
| xxxx xx0x | DAC2 zero input not detected |
| xxxx xx1x | DAC2 zero input detected |
| xxxx x0xx | DAC3 zero input not detected |
| xxxx x1xx | DAC3 zero input detected |
| xxxx 0xxx | DAC4 zero input not detected |
| xxxx 1xxx | DAC4 zero input detected |
| xxx0 xxxx | DAC5 zero input not detected |
| xxx1 xxxx | DAC5 zero input detected |
| xx0x xxxx | DAC6 zero input not detected |
| xx1x xxxx | DAC6 zero input detected |
| x0xx xxxx | DAC7 zero input not detected |
| x1xx xxxx | DAC7 zero input detected |
| 0xxx xxxx | DAC8 zero input not detected |
| 1xxx xxxx | DAC8 zero input detected |

| DEC | HEX | B7 | B6 | B5 | B4 | B3 | B2 | B1 | B0 |
|-----|-----|--------|--------|-------|-------|-------|-------|-------|------|
| 70 | 46 | ATMDDA | ATSPDA | DEMP1 | DEMP0 | AZRO2 | AZRO1 | AZRO0 | ZREV |

ATMDDA DAC Attenuation mode

This bit controls the DAC attenuation mode. ATDA1[7:0] to ATDA8[7:0] are simply used for ATMDDA = 0, and ATDA0[7:0] + ATDA1[7:0] to ATDA0[7:0] + ATDA8[7:0] in decibel number are used for ATMDDA = 1.

Default value: 0.

ATMDDA DAC Attenuation mode

- 0 Each channel with independent data (default)
- 1 All channels with preset (independent) data + master (common) data in decibel number

ATSPDA DAC Attenuation speed

This bit controls the DAC attenuation speed. $N \times 2048/f_S$ for ATSPDA = 0 and $N \times 4096/f_S$ for ATSPDA = 1. N is automatically selected according to the DAC sampling mode, SRDA, N = 1 for single rate, N = 2 for dual rate, and N = 4 for quad rate.

Default value: 0.

ATSPDA DAC Attenuation speed

- 0 $N \times 2048/f_S$ (default)
- 1 $N \times 4096/f_S$

DEMP[1:0] DAC Digital de-emphasis function/sampling rate control

These bits are used to control the enable/disable and sampling frequency of the digital de-emphasis function.

Default value: 00.

DEMP DAC Digital de-emphasis function/sampling rate control

- 00 Disable (default)
- 01 48 kHz enable
- 10 44.1 kHz enable
- 11 32 kHz enable

AZRO[2:0] DAC Zero flag function select

The AZRO[2:0] bits are used to select the function of the zero flag pin.

Default value: 000.

AZRO DAC Zero flag function select

- 000 DAC1/2/3/4/5/6/7/8 (8 channel) zero input detect with AND logic (default)
- 001 DAC1/2/3/4/5/6/7/8 (8 channel) zero input detect with OR logic
- 010 DAC1/2/3/4/5/6 (6 channel) zero input detect with AND logic
- 011 DAC1/2/3/4/5/6 (6 channel) zero input detect with OR logic
- 100 DAC7/8 (2 channel) zero input detect with AND logic
- 101 DAC7/8 (2 channel) zero input detect with OR logic
- 11x Reserved

ZREV DAC Zero flag polarity select

This bit controls the polarity of the zero flag pin.

Default value: 0.

ZREV DAC Zero flag polarity select

| | |
|---|--------------------------------|
| 0 | High for zero detect (default) |
| 1 | Low for zero detect |

| DEC | HEX | B7 | B6 | B5 | B4 | B3 | B2 | B1 | B0 |
|-----|-----|--------|--------|--------|--------|--------|--------|--------|--------|
| 71 | 47 | ATDA07 | ATDA06 | ATDA05 | ATDA04 | ATDA03 | ATDA02 | ATDA01 | ATDA00 |
| 72 | 48 | ATDA17 | ATDA16 | ATDA15 | ATDA14 | ATDA13 | ATDA12 | ATDA11 | ATDA10 |
| 73 | 49 | ATDA27 | ATDA26 | ATDA25 | ATDA24 | ATDA23 | ATDA22 | ATDA21 | ATDA20 |
| 74 | 4A | ATDA37 | ATDA36 | ATDA35 | ATDA34 | ATDA33 | ATDA32 | ATDA31 | ATDA30 |
| 75 | 4B | ATDA47 | ATDA46 | ATDA45 | ATDA44 | ATDA43 | ATDA42 | ATDA41 | ATDA40 |
| 76 | 4C | ATDA57 | ATDA56 | ATDA55 | ATDA54 | ATDA53 | ATDA52 | ATDA51 | ATDA50 |
| 77 | 4D | ATDA67 | ATDA66 | ATDA65 | ATDA64 | ATDA63 | ATDA62 | ATDA61 | ATDA60 |
| 78 | 4E | ATDA77 | ATDA76 | ATDA75 | ATDA74 | ATDA73 | ATDA72 | ATDA71 | ATDA70 |
| 79 | 4F | ATDA87 | ATDA86 | ATDA85 | ATDA84 | ATDA83 | ATDA82 | ATDA81 | ATDA80 |

ATDAx[7:0] DAC Digital attenuation level setting

Where $x = 0$ and 1 to 8, corresponding to the DAC channel, DAC $_x$ ($x = 1$ to 8).

Each DAC channel (VOU $_x$) has a digital attenuator function. The attenuation level can be set from 0 dB to –100 dB in 0.5-dB steps, and also can be set to infinite attenuation (mute). The attenuation level change from current value to target value is performed by incrementing or decrementing with s-curve responses and a time set by ATSPDA. While an attenuation level change sequence is in progress, new processing of the attenuation level change for new commands are ignored; any new commands are overwritten into the command buffer. The last command for the attenuation level change is performed after the present attenuation level change sequence is finished.

The attenuation level for each channel can be set individually using the following formula; the table below shows attenuation levels for various settings.

Attenuation level (dB) = $0.5 \times (\text{ATDA}_x[7:0]\text{DEC} - 255)$, where $\text{ATDA}_x[7:0]\text{DEC} = 0$ through 255 for $\text{ATDA}_x[7:0]\text{DEC} = 0$ through 54, attenuation is set to infinite attenuation (Mute).

ATDA0[7:0] are used to control all channels at the same time with attenuation data of ATDA0[7:0] + ATDA $_x$ [7:0] in decibel number, when ATMDDA is set to '1'. This scheme provides preset and master volume operation.

Default value: 1111 1111.

| ATDAx | Decimal value | Attenuation level setting |
|-----------|---------------|--------------------------------|
| 1111 1111 | 255 | 0 dB, no attenuation (default) |
| 1111 1110 | 254 | –0.5 dB |
| 1111 1101 | 253 | –1.0 dB |
| ... | ... | ... |
| 1000 0001 | 129 | –63.0 dB |
| 1000 0000 | 128 | –63.5 dB |
| 0111 1111 | 127 | –64 dB |
| ... | ... | ... |
| 0011 1000 | 56 | –99.5 dB |
| 0011 0111 | 55 | –100 dB |
| 0011 0110 | 54 | Mute |
| ... | ... | ... |
| 0000 0000 | 0 | Mute |

| DEC | HEX | B7 | B6 | B5 | B4 | B3 | B2 | B1 | B0 |
|-----|-----|----|----|----|----|----|----|-------|-------|
| 80 | 50 | — | — | — | — | — | — | SRAD1 | SRAD0 |

SRAD[1:0] ADC Sampling mode select

These bits control the sampling mode of ADC operation. In Auto mode, the sampling mode is automatically set according to multiples between system clock and sampling clock, single rate for $512 f_s$ and $768 f_s$, and dual rate for $256 f_s$ and $384 f_s$.

Default value: 00.

| SRAD | ADC Sampling mode select |
|------|--------------------------|
| 00 | Auto (default) |
| 01 | Single rate |
| 10 | Dual rate |
| 11 | Reserved |

| DEC | HEX | B7 | B6 | B5 | B4 | B3 | B2 | B1 | B0 |
|-----|-----|----|-------|-------|-------|----|--------|--------|--------|
| 81 | 51 | — | MSAD2 | MSAD1 | MSAD0 | — | FMTAD2 | FMTAD1 | FMTAD0 |

MSAD[2:0] ADC Master/slave mode select

These bits control the audio interface mode for ADC operation.

Default value: 000 (slave mode).

MSAD ADC Master/slave mode select

| | |
|-----|------------------------|
| 000 | Slave mode (default) |
| 001 | Master mode, $768 f_S$ |
| 010 | Master mode, $512 f_S$ |
| 011 | Master mode, $384 f_S$ |
| 100 | Master mode, $256 f_S$ |
| 101 | Reserved |
| 110 | Reserved |
| 111 | Reserved |

FMTAD[2:0] ADC Audio interface format select

These bits control the audio interface format for ADC operation. The format details and restrictions related to the system clock and master/slave mode are described in the [Audio Data Interface Formats and Timing](#) section.

Default value: 000 (24-bit I²S format).

FMTAD ADC Audio interface format select

| | |
|-----|--|
| 000 | 24-bit I ² S format (default) |
| 001 | 24-bit left-justified format |
| 010 | 24-bit right-justified format |
| 011 | 16-bit right-justified format |
| 100 | 24-bit I ² S mode DSP format |
| 101 | 24-bit left-justified mode DSP format |
| 110 | 24-bit I ² S mode TDM format |
| 111 | 24-bit left-justified mode TDM format |

| DEC | HEX | B7 | B6 | B5 | B4 | B3 | B2 | B1 | B0 |
|-----|-----|----|--------|--------|--------|----|------|------|------|
| 82 | 52 | — | PSVAD2 | PSVAD1 | PSVAD0 | — | BYP2 | BYP1 | BYP0 |

PSVAD[2:0] ADC Power-save control

These bits control the ADC power-save mode. In power-save mode, DOUT is forced into ZERO with a fade-out sequence, the internal ADC data are reset, and the ADC goes into a power-down state. For power-save mode release, a fade-in sequence is applied on DOUT in resume process. The serial mode control is enabled during this mode. Wait times greater than $t_{ADCDLY2}$ are required for the status change because of the power-save control turning on/off.

Default value: 000.

PSVAD ADC Power-save control

- xx0 ADC1/2 normal operation
- xx1 ADC1/2 power-save mode
- x0x ADC3/4 normal operation
- x1x ADC3/4 power-save mode
- 0xx ADC5/6 normal operation
- 1xx ADC5/6 power-save mode

BYP[2:0] ADC HPF bypass control

These bits control the HPF function and dc components of the input signal; internal dc offset is converted in bypass mode.

Default value: 000.

BYP ADC HPF bypass control

- xx0 ADC1/2 normal output, HPF enabled
- xx1 ADC1/2 bypassed output, HPF disabled
- x0x ADC3/4 normal output, HPF enabled
- x1x ADC3/4 bypassed output, HPF disabled
- 0xx ADC5/6 normal output, HPF enabled
- 1xx ADC5/6 bypassed output, HPF disabled

| DEC | HEX | B7 | B6 | B5 | B4 | B3 | B2 | B1 | B0 |
|-----|-----|----|----|-------|-------|-------|-------|-------|-------|
| 83 | 53 | — | — | SEAD6 | SEAD5 | SEAD4 | SEAD3 | SEAD2 | SEAD1 |

SEAD[6:1] ADC Input configuration control

These bits control the input configuration of each ADC channel, differential or single-ended.

Default value: 00 0000 (all ADC channels have differential inputs).

| SEAD | ADC Input configuration |
|-------------|--------------------------------|
| xx xxx0 | ADC1 differential input |
| xx xxx1 | ADC1 single-ended input |
| xx xx0x | ADC2 differential input |
| xx xx1x | ADC2 single-ended input |
| xx x0xx | ADC3 differential input |
| xx x1xx | ADC3 single-ended input |
| xx 0xxx | ADC4 differential input |
| xx 1xxx | ADC4 single-ended input |
| x0 xxxx | ADC5 differential input |
| x1 xxxx | ADC5 single-ended input |
| 0x xxxx | ADC6 differential input |
| 1x xxxx | ADC6 single-ended input |

| DEC | HEX | B7 | B6 | B5 | B4 | B3 | B2 | B1 | B0 |
|-----|-----|----|----|--------|--------|--------|--------|--------|--------|
| 84 | 54 | — | — | REVAD6 | REVAD5 | REVAD4 | REVAD3 | REVAD2 | REVAD1 |

REVAD[6:1] ADC Input phase select

These bits are used to control the phase of analog signal inputs.

Default value: 00 0000.

REVAD ADC Input phase select

- xx xxx0 ADC1 normal input
- xx xxx1 ADC1 inverted input
- xx xx0x ADC2 normal input
- xx xx1x ADC2 inverted input
- xx x0xx ADC3 normal input
- xx x1xx ADC3 inverted input
- xx 0xxx ADC4 normal input
- xx 1xxx ADC4 inverted input
- x0 xxxx ADC5 normal input
- x1 xxxx ADC5 inverted input
- 0x xxxx ADC6 normal input
- 1x xxxx ADC6 inverted input

| DEC | HEX | B7 | B6 | B5 | B4 | B3 | B2 | B1 | B0 |
|-----|-----|----|----|--------|--------|--------|--------|--------|--------|
| 85 | 55 | — | — | MUTAD6 | MUTAD5 | MUTAD4 | MUTAD3 | MUTAD2 | MUTAD1 |

MUTAD[6:1] ADC Soft Mute control

These bits are used to enable or disable the Soft Mute function for the corresponding ADC outputs, DOUT. The Soft Mute function is incorporated into the digital attenuators. When Mute is disabled (MUTAD[6:1] = 0), the attenuator and ADC operate normally. When Mute is enabled by setting MUTAD[6:1] = 1, the digital attenuator for the corresponding output decreases from the current setting to infinite attenuation with an s-curve responses and time set by ATSPAD.

By setting MUTAD[6:1] = 0, the attenuator increases to the last attenuation level with the s-curve response in same manner as for decreasing levels. This provides *pop and zipper noise-free* muting for the ADC input.

The Soft Mute control uses the same digital attenuation level resource setting as the ADC. Mute control has priority over the digital attenuation level setting.

Default value: 00 0000.

MUTAD ADC Soft Mute control

| | |
|---------|--------------------|
| xx xxx0 | ADC1 Mute disabled |
| xx xxx1 | ADC1 Mute enabled |
| xx xx0x | ADC2 Mute disabled |
| xx xx1x | ADC2 Mute enabled |
| xx x0xx | ADC3 Mute disabled |
| xx x1xx | ADC3 Mute enabled |
| xx 0xxx | ADC4 Mute disabled |
| xx 1xxx | ADC4 Mute enabled |
| x0 xxxx | ADC5 Mute disabled |
| x1 xxxx | ADC5 Mute enabled |
| 0x xxxx | ADC6 Mute disabled |
| 1x xxxx | ADC6 Mute enabled |

| DEC | HEX | B7 | B6 | B5 | B4 | B3 | B2 | B1 | B0 |
|-----|-----|----|----|------|------|------|------|------|------|
| 86 | 56 | — | — | OVF6 | OVF5 | OVF4 | OVF3 | OVF2 | OVF1 |

OVF[6:1] ADC Overflow flag (read-only)

These bits indicate the status information of an overflow detect circuit for each ADC channel; these bits are read only. '1' means an overflow has been detected in the past, and reading this register resets all OVF bits.

| OVF | ADC Overflow flag |
|------------|----------------------------------|
| xx xxx0 | ADC1 overflow input not detected |
| xx xxx1 | ADC1 overflow input detected |
| xx xx0x | ADC2 overflow input not detected |
| xx xx1x | ADC2 overflow input detected |
| xx x0xx | ADC3 overflow input not detected |
| xx x1xx | ADC3 overflow input detected |
| xx 0xxx | ADC4 overflow input not detected |
| xx 1xxx | ADC4 overflow input detected |
| x0 xxxx | ADC5 overflow input not detected |
| x1 xxxx | ADC5 overflow input detected |
| 0x xxxx | ADC6 overflow input not detected |
| 1x xxxx | ADC6 overflow input detected |

| DEC | HEX | B7 | B6 | B5 | B4 | B3 | B2 | B1 | B0 |
|-----|-----|--------|--------|----|----|----|----|----|------|
| 87 | 57 | ATMDAD | ATSPAD | — | — | — | — | — | OVFP |

ATMDAD ADC Attenuation mode

This bit controls the ADC attenuation mode. ATAD1[7:0] to ATAD6[7:0] are simply used for ATMDAD = 0, and ATAD0[7:0] + ATAD1[7:0] to ATAD0[7:0] + ATAD6[7:0] in decibel number are used for ATMDAD = 1.

Default value: 0.

ATMDAD ADC Attenuation mode

- 0 Each channel with independent data (default)
- 1 All channels with preset (independent) data + master (common) data in decibel number

ATSPAD ADC Attenuation speed

This bit controls the ADC attenuation Speed, $N \times 2048/f_s$ for ATSPAD = 0 and $N \times 4096/f_s$ for ATSPAD = 1. N is automatically selected according to the ADC sampling mode, SRAD: N = 1 for single and N = 2 for dual rate.

Default value: 0.

ATSPAD ADC Attenuation speed

- 0 $N \times 2048/f_s$ (default)
- 1 $N \times 4096/f_s$

OVFP ADC Overflow flag polarity select

This bit controls the polarity of the overflow flag pin.

Default value: 0.

OVFP ADC Overflow flag polarity select

- 0 High for overflow detect (default)
- 1 Low for overflow detect

| DEC | HEX | B7 | B6 | B5 | B4 | B3 | B2 | B1 | B0 |
|-----|-----|--------|--------|--------|--------|--------|--------|--------|--------|
| 88 | 58 | ATAD07 | ATAD06 | ATAD05 | ATAD04 | ATAD03 | ATAD02 | ATAD01 | ATAD00 |
| 89 | 59 | ATAD17 | ATAD16 | ATAD15 | ATAD14 | ATAD13 | ATAD12 | ATAD11 | ATAD10 |
| 90 | 5A | ATAD27 | ATAD26 | ATAD25 | ATAD24 | ATAD23 | ATAD22 | ATAD21 | ATAD20 |
| 91 | 5B | ATAD37 | ATAD36 | ATAD35 | ATAD34 | ATAD33 | ATAD32 | ATAD31 | ATAD30 |
| 92 | 5C | ATAD47 | ATAD46 | ATAD45 | ATAD44 | ATAD43 | ATAD42 | ATAD41 | ATAD40 |
| 93 | 5D | ATAD57 | ATAD56 | ATAD55 | ATAD54 | ATAD53 | ATAD52 | ATAD51 | ATAD50 |
| 94 | 5E | ATAD67 | ATAD66 | ATAD65 | ATAD64 | ATAD63 | ATAD62 | ATAD61 | ATAD60 |

ATADx[7:0] ADC Digital attenuation level setting

Where x = 0 and 1 to 6, corresponding to the ADC channel, ADCx (x = 1 to 6).

Each ADC channel has a digital attenuator function with 20-dB gain. The attenuation level can be set from 20 dB to –100 dB in 0.5-dB steps, and also can be set to infinite attenuation (mute). The attenuation level change from current value to target value is performed by increment or decrement with s-curve response and time set by ATSPAD. While the attenuation level change sequence is in progress, new processing of an attenuation level change for a new command is ignored; the new command is overwritten into the command buffer. The last command for an attenuation level change is performed after the present attenuation level change sequence is finished.

The attenuation level for each channel can be set individually using the following formula, and the above table shows attenuation levels for various settings.

Attenuation level (dB) = 0.5 × (ATADx[7:0]DEC – 215), where ATADx[7:0]DEC = 0 through 255 for ATADx[7:0]DEC = 0 through 14, attenuation is set to infinite attenuation (Mute).

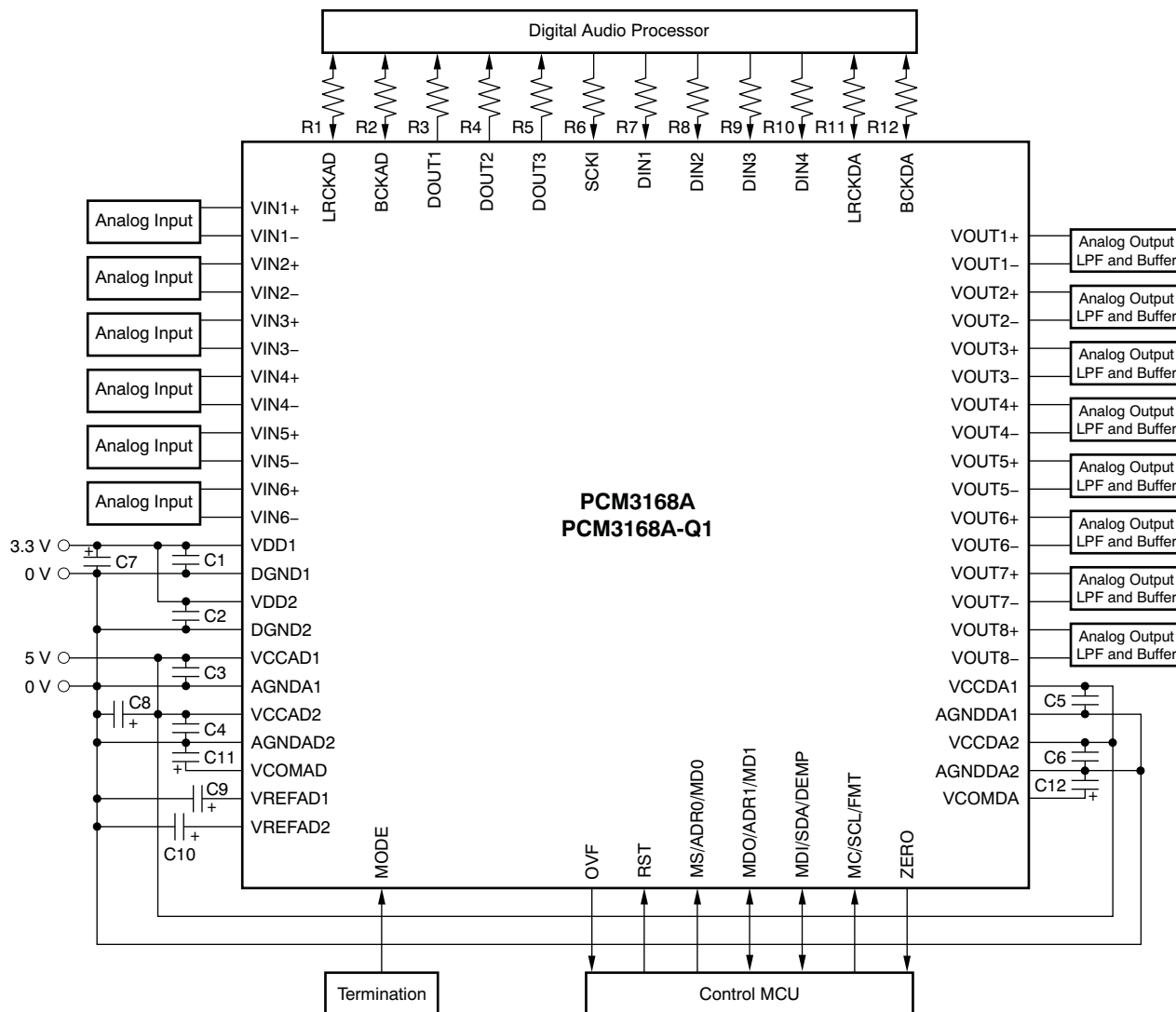
ATAD0[7:0] is used to control all channels at the same time with attenuation data of ATAD0[7:0] + ATADx[7:0] in decibel number, though maximum level is limited within +20 dB, when ATMDAD is set to '1'. This scheme provides preset and master volume operation.

Default value: 1101 0111.

| ATADx | Decimal value | Attenuation level setting |
|-----------|---------------|--------------------------------|
| 1111 1111 | 255 | +20.0 dB |
| 1111 1110 | 254 | +19.5 dB |
| 1111 1101 | 253 | +19.0 dB |
| ... | ... | ... |
| 1101 1000 | 216 | +0.5 dB |
| 1101 0111 | 215 | 0 dB, no attenuation (default) |
| 1101 0110 | 214 | –0.5 dB |
| ... | ... | ... |
| 0001 0000 | 16 | –99.5 dB |
| 0000 1111 | 15 | –100.0 dB |
| 0000 1110 | 14 | Mute |
| ... | ... | ... |
| 0000 0000 | 0 | Mute |

APPLICATION INFORMATION

A typical circuit connection for six-channel analog in and eight-channel analog out is shown in Figure 56.



C₁ through C₆ are 1-μF ceramic capacitors dependent on power-supply quality. C₇ and C₈ are 10-μF electrolytic capacitors dependent on power-supply quality. C₉ and C₁₀ are 10-μF electrolytic capacitors. C₁₁ and C₁₂ are 10-μF electrolytic capacitors. R₁ through R₁₂ are 22-Ω to 100-Ω resistors.

Figure 56. Example Board Layout

Termination for mode control: any one of the circuits shown in Figure 57 must be applied according to the necessary mode/configuration. Resistor value must be 220 kΩ, ±5% tolerant. The PowerPAD must be tied to the ground plane with enough electrical and thermal conductivity; see the example board layout in Figure 56.

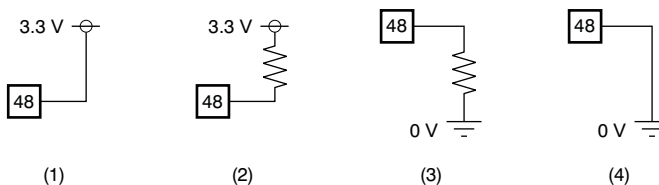
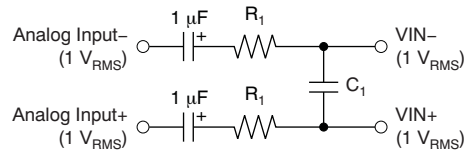


Figure 57. Typical Circuit Connections

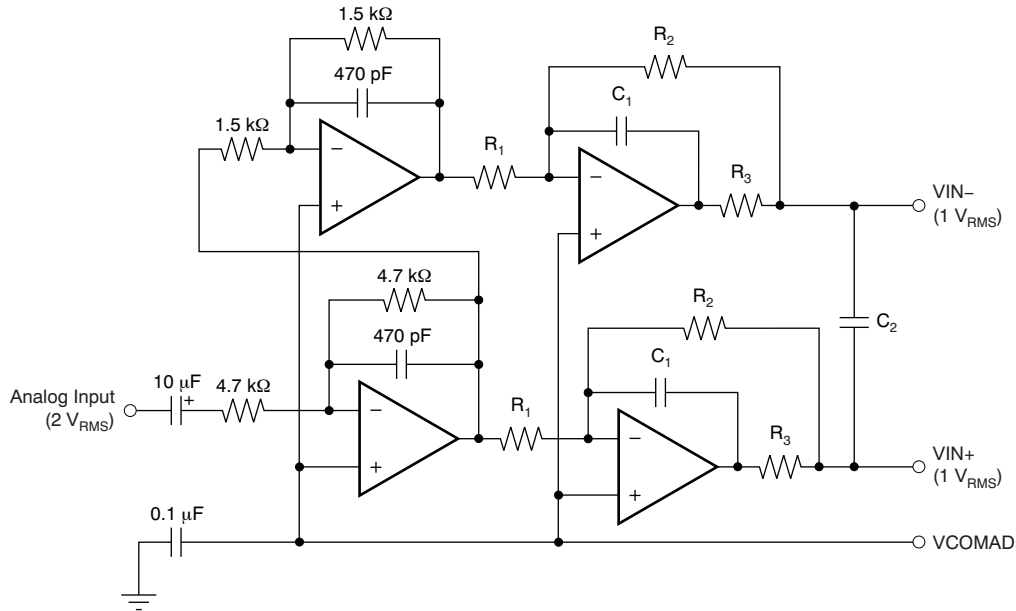
Typical interface circuits for analog input and analog output are shown in Figure 58 through Figure 63.



$R_1 = 47\text{-}\Omega$ to $470\text{-}\Omega$ resistor, $C_1 = 0.01\text{-pF}$ to $0.001\text{-}\mu\text{F}$ capacitor; $f_{-3\text{ dB}} = 160\text{ kHz}$.

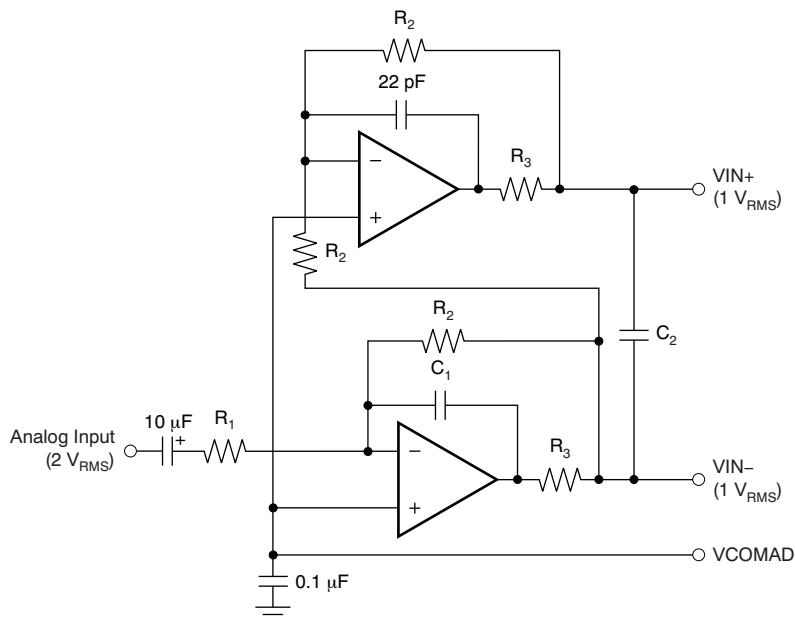
NOTE: The signal source impedance must be low enough to apply this configuration.

Figure 58. Basic Differential Input Circuit with Anti-Aliasing LPF for Differential ADC Input



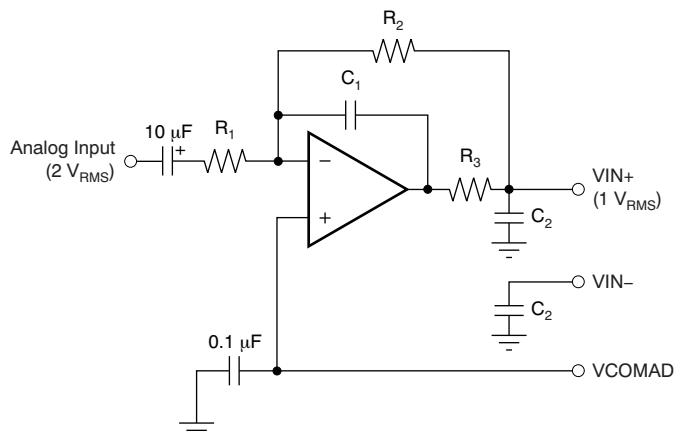
Amplifier is an NE5532A x2 or OPA2134 x2; $R_1 = 1.5\text{-k}\Omega$ resistor; $R_2 = 750\text{-}\Omega$ resistor; $R_3 = 47\text{-}\Omega$ resistor; $C_1 = 3300\text{-pF}$ capacitor; $C_2 = 0.01\text{-}\mu\text{F}$ capacitor; Gain = 1; $f_{-3\text{ dB}} = 45\text{ kHz}$.

Figure 59. Single-Ended to Differential Buffer and Anti-Aliasing LPF for Differential ADC Input



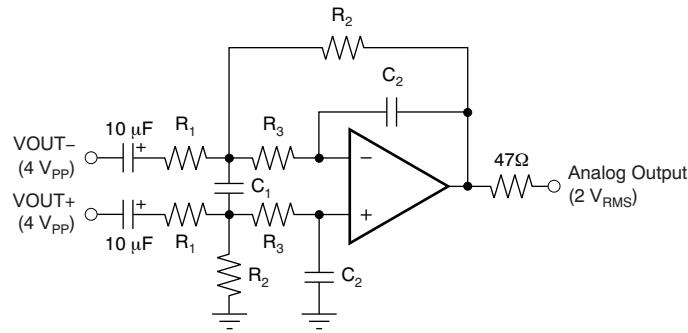
Amplifier is an NE5532A x1 or OPA2134 x1; $R_1 = 3\text{-k}\Omega$ resistor; $R_2 = 1.5\text{-k}\Omega$ resistor; $R_3 = 47\text{-}\Omega$ resistor; $C_1 = 2200\text{-pF}$ capacitor; $C_2 = 0.01\text{-}\mu\text{F}$ capacitor; Gain = 1; $f_{-3\text{ dB}} = 48\text{ kHz}$.

Figure 60. Single-Ended to Differential Buffer and Anti-Aliasing LPF for Differential ADC Input



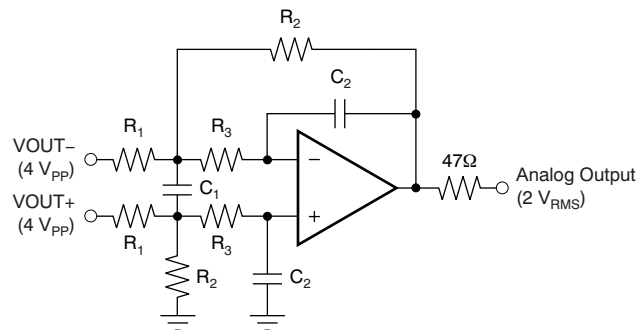
Amplifier is an NE5532A x1 or OPA2134 x1; $R_1 = 3\text{-k}\Omega$ resistor; $R_2 = 1.5\text{-k}\Omega$ resistor; $R_3 = 47\text{-}\Omega$ resistor; $C_1 = 2200\text{-pF}$ capacitor; $C_2 = 0.022\text{-}\mu\text{F}$ capacitor; Gain = 0.5; $f_{-3\text{ dB}} = 48\text{ kHz}$.

Figure 61. Buffer and Anti-Aliasing LPF for Single-Ended ADC Input



Amplifier is an NE5532A x1/2 or OPA2134 x1/2; $R_1 = 7.5\text{-k}\Omega$ resistor; $R_2 = 5.6\text{-k}\Omega$ resistor; $R_3 = 360\text{-}\Omega$ resistor; $C_1 = 3300\text{-pF}$ capacitor; $C_2 = 680\text{-pF}$ capacitor; Gain = 0.747; $f_{-3\text{ dB}} = 53\text{ kHz}$.

Figure 62. Post-LPF and Differential to Single-Ended Buffer for DAC Output (AC-Coupled)



Amplifier is an NE5532A x1/2 or OPA2134 x1/2; $R_1 = 15\text{-k}\Omega$ resistor; $R_2 = 11\text{-k}\Omega$ resistor; $R_3 = 820\text{-}\Omega$ resistor; $C_1 = 1500\text{-pF}$ capacitor; $C_2 = 330\text{-pF}$ capacitor; Gain = 0.733; $f_{-3\text{ dB}} = 54\text{ kHz}$.

Figure 63. Post-LPF and Differential to Single-Ended Buffer for DAC Output (DC-Coupled)

DESIGN AND LAYOUT CONSIDERATIONS

Power-Supply Pins (VCCAD1/2, VCCDA1/2, and VDD1/2)

The digital and analog power-supply pins of the PCM3168A and PCM3168A-Q1 should be bypassed to the corresponding ground pins with 1- μ F ceramic capacitors placed as close to the pins as possible. Each power-supply line (VCC, VDD) to the PCM3168A and PCM3168A-Q1 should be bypassed to the corresponding ground pins with 10- μ F electrolytic capacitors to maximize the dynamic performance of the ADC and DAC.

Although the PCM3168A and PCM3168A-Q1 have two power lines to maximize the potential of dynamic performance, using one common source (for instance, a +5-V power supply for VCC and a +3.3-V power supply for VDD generated from one common source) is recommended to avoid unexpected power-supply trouble such as latch-up or incorrect power-supply conditions. Also, simultaneous power-on/off of VCC and VDD is recommended to avoid unexpected transient responses in outputs, though the power-supply sequence of VCC and VDD is not specified in the operation and absolute maximum ratings point of view.

Grounding (AGNDAD1/2, AGNDDA1/2, and DGND1/2)

To maximize the dynamic performance of the PCM3168A and PCM3168A-Q1, the analog and digital grounds are not connected internally. These pins should have very low impedances to avoid digital noise and signal components feeding back into the analog ground. All ground pins should be connected directly to each other under the part, and the device should be connected to the analog ground of the application, as with acceptable analog layout practices; this layout reduces the potential of noise problems.

VIN1 \pm , VIN2 \pm , VIN3 \pm , VIN4 \pm , VIN5 \pm , and VIN6 \pm Pins

In case of direct interface to VINx \pm , 1- μ F electrolytic capacitors are recommended because the ac-coupling capacitor (which gives a 2-Hz HPF corner frequency and 47 Ω and 0.1 μ F to 470 Ω and 0.001 μ F differential LPF) is recommended as the anti-aliasing filter that gives a 160-kHz LPF corner frequency. If signal source impedance is not enough (too low) or input line length to the VINx \pm is not enough (too short), insertion of an analog front-end buffer (see [Figure 59](#) to [Figure 61](#)) is recommended to maximize the dynamic performance. The voltage coefficient of the capacitor for an anti-aliasing filter should be considered to maximize the THD performance. A film-type capacitor is recommended; if a ceramic capacitor is used, a relatively higher voltage type is recommended.

There are three ways to terminate any unused input pins. First, terminate these pins to AGNDAD with 0.001- μ F to 1- μ F capacitors. This termination is applied on unused pins whose channels are configured in single-ended mode. The second form of termination is to connect the positive (+) pin and negative (–) pins together and terminating these to AGNDAD with 0.001- μ F to 1- μ F capacitors. This option applies to unused pins with channels that are configured in differential mode. The last termination method is to terminate the pins directly to VCOMAD; this option can be applied on unused pins with unused channels combined into two channels that are then configured in power-save mode.

VCOMAD and VCOMDA Pins

10- μ F electrolytic capacitors are recommended between VCOMAD and AGNDAD, and VCOMDA and AGNDDA to ensure a low source impedance of ADC and DAC common voltages. These capacitors should be located as close to each pin as possible to reduce dynamic errors on the ADC and DAC common voltages.

VREFAD1/2 Pins

10- μ F electrolytic capacitors are recommended between VREFAD1/2 and AGNDAD to ensure low source impedances of ADC references. These capacitors should be located as close to each pin as possible to reduce dynamic errors on ADC references.

VOUT1 \pm , VOU2 \pm , VOUT3 \pm , VOUT4 \pm , VOUT5 \pm , VOUT6 \pm , VOU7 \pm , and VOUT8 \pm Pins

The differential to single-ended buffer with post LPF can be directly connected (without capacitors) to these output pins (see [Figure 63](#)), thereby minimizing the use of coupling capacitors for the 2- V_{RMS} outputs. The op amp and resistors should be determined with consideration of degrading some performance through this differential to single-ended and LPF buffer; there is about 1.5-dB degradation seen in the examples of [Figure 62](#) and [Figure 63](#).

MODE Pin

This pin is a logic input with quad-state input capability. The MODE pin is high when connected to VDD, low when connected to DGND, and pulled up or pulled down through an external resistor and for the two mid-states in order to distinguish the four input states. The pull-up or pull-down resistor must be 220 k Ω , $\pm 5\%$ in tolerance. Note that the state of the MODE pin is only sampled by a power-on or a low-to-high transition of the RST pin.

RST Pin

When the MODE pin setting changes to change the operating mode, the new mode setting does not take effect immediately; a RST pin toggle is required to make the new mode setting valid, and for the new mode to take effect.

OVF Pin

The OVF pin has two functions. It is primarily the flag for ADC overflow occurrence detection. It is also used to indicate that the internal reset sequence is complete and that the device is ready to enter serial mode control.

System Clock and Audio Interface Clocks

The quality of SCKI may influence dynamic performance, because the PCM3168A and PCM3168A-Q1 (both the ADC and DAC) operate based on SCKI. Therefore, it may be required to consider the jitter, duty, and rise/fall time of the system clock.

In slave mode, the PCM3168A and PCM3168A-Q1 do not require a specific timing relationship between BCKAD/LRCKAD and SCKI, and BCKDA/LRCKDA and SCKI; however, there is a possibility of performance degradation with a certain timing relationship between them. In that case, specific timing relationship control might resolve this performance degradation.

In master mode, there is a possibility of performance degradation because of heavy loads on BCKAD/LRCKAD, BCKDA/LRCKDA, and DOUT1/2/3. It is recommended to load these pins as lightly as possible. Note that all output clocks and signals go low; they do not go into a high-impedance state during power-save mode.

PowerPAD

The PowerPAD of the PCM3168A and PCM3168A-Q1 is internally connected to the substrate of the silicon. It should be connected to the ground plane with sufficient low conductance in electrical and thermal; see [Figure 56](#). The PowerPAD size is 7,25 mm x 7,00 mm (0,725 cm x 0.7 cm).

External Mute Control

For power-down ON/OFF control without the pop-noise that is generated by a dc level change on the DAC output, the external mute control is generally required. Use of the following control sequence is recommended: external mute ON, codec power-down ON, SCKI stop and resume if necessary, codec power-down OFF, and external mute OFF control.

PACKAGING INFORMATION

| Orderable Device | Status (1) | Package Type | Package Drawing | Pins | Package Qty | Eco Plan (2) | Lead/Ball Finish | MSL Peak Temp (3) | Op Temp (°C) | Top-Side Markings (4) | Samples |
|-------------------|---------------|--------------|-----------------|------|-------------|-------------------------|------------------|----------------------|--------------|--------------------------|-------------------------|
| PCM3168APAP | ACTIVE | HTQFP | PAP | 64 | 160 | Green (RoHS & no Sb/Br) | CU NIPDAU | Level-3-260C-168 HR | -40 to 85 | PCM3168A | Samples |
| PCM3168APAPR | ACTIVE | HTQFP | PAP | 64 | 1000 | Green (RoHS & no Sb/Br) | CU NIPDAU | Level-3-260C-168 HR | -40 to 85 | PCM3168A | Samples |
| PCM3168APAPRG4 | ACTIVE | HTQFP | PAP | 64 | 1000 | Green (RoHS & no Sb/Br) | CU NIPDAU | Level-3-260C-168 HR | -40 to 85 | PCM3168A | Samples |
| PCM3168ATPAPQ1 | ACTIVE | HTQFP | PAP | 64 | 160 | Green (RoHS & no Sb/Br) | CU NIPDAU | Level-3-260C-168 HR | -40 to 105 | PCM3168AQ1 | Samples |
| PCM3168ATPAPRQ1 | ACTIVE | HTQFP | PAP | 64 | 1000 | Green (RoHS & no Sb/Br) | CU NIPDAU | Level-3-260C-168 HR | -40 to 105 | PCM3168AQ1 | Samples |
| PCM3168ATPAPRQ1G4 | ACTIVE | HTQFP | PAP | 64 | 1000 | Green (RoHS & no Sb/Br) | CU NIPDAU | Level-3-260C-168 HR | -40 to 105 | PCM3168AQ1 | Samples |

(1) The marketing status values are defined as follows:

ACTIVE: Product device recommended for new designs.

LIFEBUY: TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.

NRND: Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.

PREVIEW: Device has been announced but is not in production. Samples may or may not be available.

OBSELETE: TI has discontinued the production of the device.

(2) Eco Plan - The planned eco-friendly classification: Pb-Free (RoHS), Pb-Free (RoHS Exempt), or Green (RoHS & no Sb/Br) - please check <http://www.ti.com/productcontent> for the latest availability information and additional product content details.

TBD: The Pb-Free/Green conversion plan has not been defined.

Pb-Free (RoHS): TI's terms "Lead-Free" or "Pb-Free" mean semiconductor products that are compatible with the current RoHS requirements for all 6 substances, including the requirement that lead not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, TI Pb-Free products are suitable for use in specified lead-free processes.

Pb-Free (RoHS Exempt): This component has a RoHS exemption for either 1) lead-based flip-chip solder bumps used between the die and package, or 2) lead-based die adhesive used between the die and leadframe. The component is otherwise considered Pb-Free (RoHS compatible) as defined above.

Green (RoHS & no Sb/Br): TI defines "Green" to mean Pb-Free (RoHS compatible), and free of Bromine (Br) and Antimony (Sb) based flame retardants (Br or Sb do not exceed 0.1% by weight in homogeneous material)

(3) MSL, Peak Temp. -- The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.

(4) Multiple Top-Side Markings will be inside parentheses. Only one Top-Side Marking contained in parentheses and separated by a "~" will appear on a device. If a line is indented then it is a continuation of the previous line and the two combined represent the entire Top-Side Marking for that device.

Important Information and Disclaimer: The information provided on this page represents TI's knowledge and belief as of the date that it is provided. TI bases its knowledge and belief on information provided by third parties, and makes no representation or warranty as to the accuracy of such information. Efforts are underway to better integrate information from third parties. TI has taken and continues to take reasonable steps to provide representative and accurate information but may not have conducted destructive testing or chemical analysis on incoming materials and chemicals. TI and TI suppliers consider certain information to be proprietary, and thus CAS numbers and other limited information may not be available for release.

In no event shall TI's liability arising out of such information exceed the total purchase price of the TI part(s) at issue in this document sold by TI to Customer on an annual basis.

OTHER QUALIFIED VERSIONS OF PCM3168A, PCM3168A-Q1 :

- Catalog: [PCM3168A](#)
- Automotive: [PCM3168A-Q1](#)

NOTE: Qualified Version Definitions:

- Catalog - TI's standard catalog product
- Automotive - Q100 devices qualified for high-reliability automotive applications targeting zero defects

TAPE AND REEL INFORMATION

QUADRANT ASSIGNMENTS FOR PIN 1 ORIENTATION IN TAPE


*All dimensions are nominal

| Device | Package Type | Package Drawing | Pins | SPQ | Reel Diameter (mm) | Reel Width W1 (mm) | A0 (mm) | B0 (mm) | K0 (mm) | P1 (mm) | W (mm) | Pin1 Quadrant |
|-----------------|--------------|-----------------|------|------|--------------------|--------------------|---------|---------|---------|---------|--------|---------------|
| PCM3168APAPR | HTQFP | PAP | 64 | 1000 | 330.0 | 24.4 | 13.0 | 13.0 | 1.5 | 16.0 | 24.0 | Q2 |
| PCM3168ATPAPRQ1 | HTQFP | PAP | 64 | 1000 | 330.0 | 24.4 | 13.0 | 13.0 | 1.5 | 16.0 | 24.0 | Q2 |

TAPE AND REEL BOX DIMENSIONS



*All dimensions are nominal

| Device | Package Type | Package Drawing | Pins | SPQ | Length (mm) | Width (mm) | Height (mm) |
|-----------------|--------------|-----------------|------|------|-------------|------------|-------------|
| PCM3168APAPR | HTQFP | PAP | 64 | 1000 | 367.0 | 367.0 | 45.0 |
| PCM3168ATPAPRQ1 | HTQFP | PAP | 64 | 1000 | 367.0 | 367.0 | 45.0 |

PAP (S-PQFP-G64)

PowerPAD™ PLASTIC QUAD FLATPACK



- NOTES:
- A. All linear dimensions are in millimeters.
 - B. This drawing is subject to change without notice.
 - C. Body dimensions do not include mold flash or protrusion
 - D. This package is designed to be soldered to a thermal pad on the board. Refer to Technical Brief, PowerPad Thermally Enhanced Package, Texas Instruments Literature No. SLMA002 for information regarding recommended board layout. This document is available at www.ti.com <<http://www.ti.com>>.
 - E. Falls within JEDEC MS-026

PowerPAD is a trademark of Texas Instruments.

THERMAL PAD MECHANICAL DATA

PAP (S-PQFP-G64)

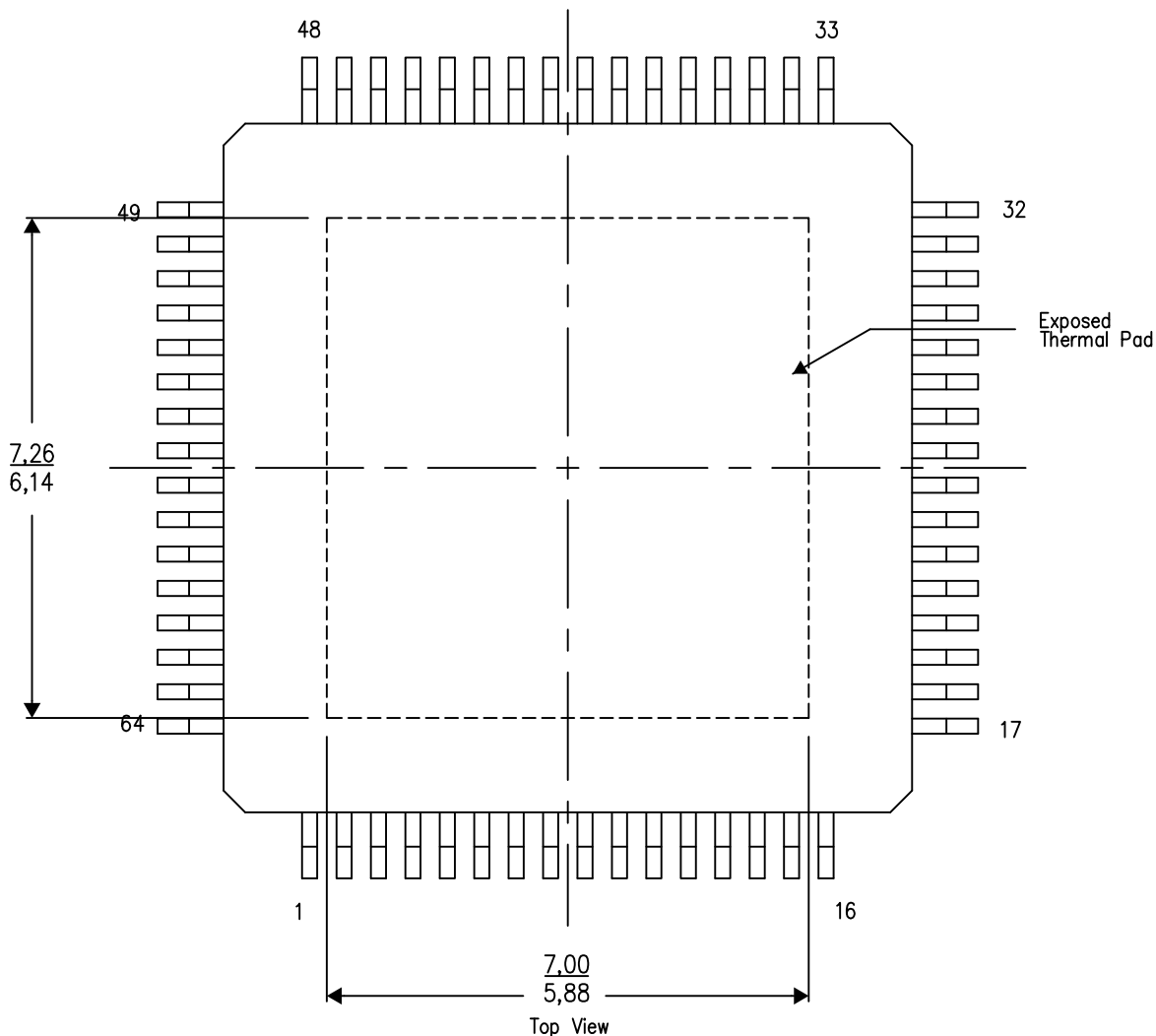
PowerPAD™ PLASTIC QUAD FLATPACK

THERMAL INFORMATION

This PowerPAD™ package incorporates an exposed thermal pad that is designed to be attached to a printed circuit board (PCB). The thermal pad must be soldered directly to the PCB. After soldering, the PCB can be used as a heatsink. In addition, through the use of thermal vias, the thermal pad can be attached directly to the appropriate copper plane shown in the electrical schematic for the device, or alternatively, can be attached to a special heatsink structure designed into the PCB. This design optimizes the heat transfer from the integrated circuit (IC).

For additional information on the PowerPAD package and how to take advantage of its heat dissipating abilities, refer to Technical Brief, PowerPAD Thermally Enhanced Package, Texas Instruments Literature No. SLMA002 and Application Brief, PowerPAD Made Easy, Texas Instruments Literature No. SLMA004. Both documents are available at www.ti.com.

The exposed thermal pad dimensions for this package are shown in the following illustration.



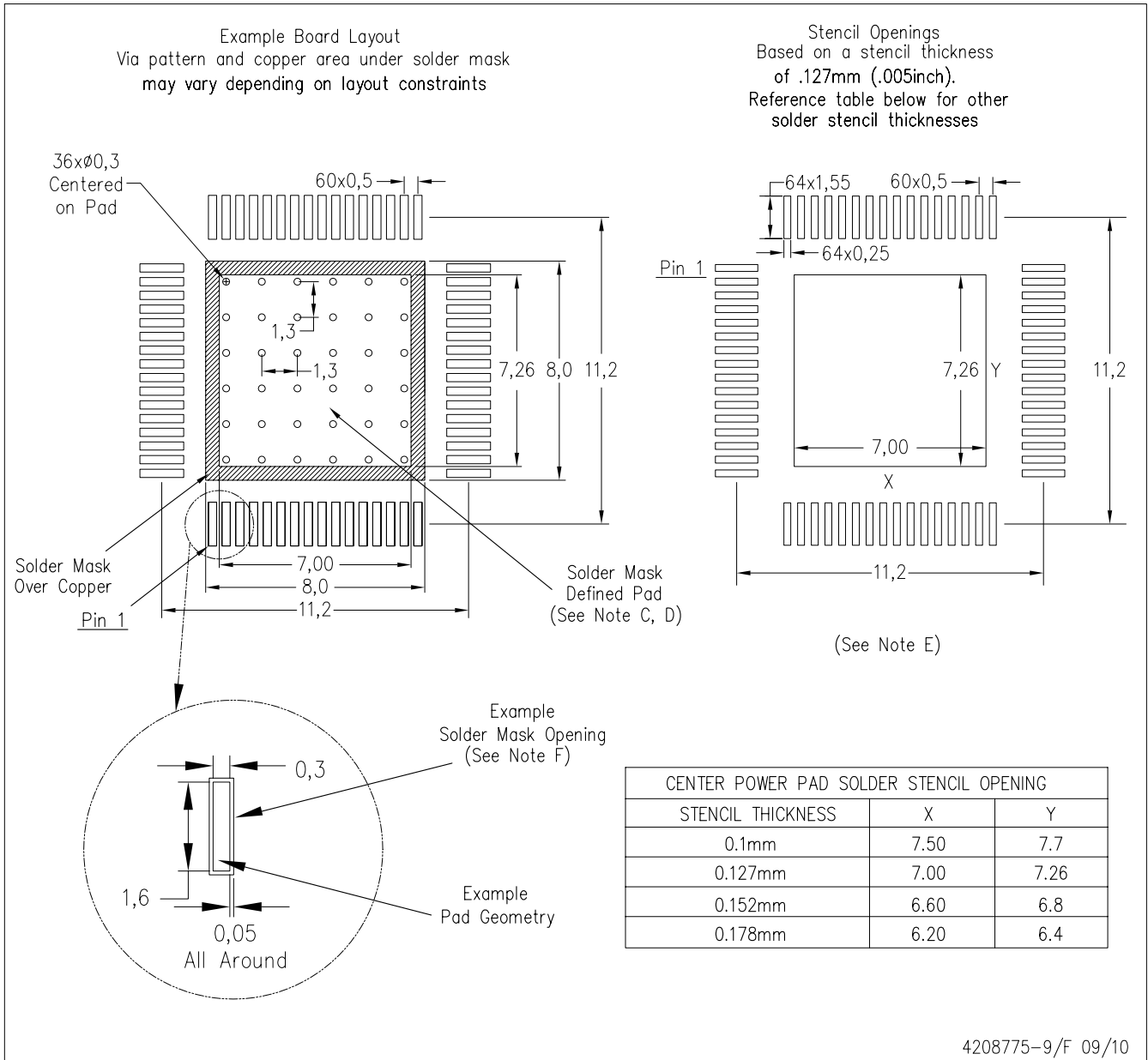
4206326-9/0 10/13

NOTES: A. All linear dimensions are in millimeters

PowerPAD is a trademark of Texas Instruments

PAP (S-PQFP-G64)

PowerPAD™ PLASTIC QUAD FLATPACK



- NOTES:
- A. All linear dimensions are in millimeters.
 - B. This drawing is subject to change without notice.
 - C. Customers should place a note on the circuit board fabrication drawing not to alter the center solder mask defined pad.
 - D. This package is designed to be soldered to a thermal pad on the board. Refer to Technical Brief, PowerPAD Thermally Enhanced Package, Texas Instruments Literature No. SLMA002, SLMA004, and also the Product Data Sheets for specific thermal information, via requirements, and recommended board layout. These documents are available at www.ti.com <<http://www.ti.com>>. Publication IPC-7351 is recommended for alternate designs.
 - E. Laser cutting apertures with trapezoidal walls and also rounding corners will offer better paste release. Customers should contact their board assembly site for stencil design recommendations. Example stencil design based on a 50% volumetric metal load solder paste. Refer to IPC-7525 for other stencil recommendations.
 - F. Customers should contact their board fabrication site for solder mask tolerances between and around signal pads.

PowerPAD is a trademark of Texas Instruments

IMPORTANT NOTICE

Texas Instruments Incorporated and its subsidiaries (TI) reserve the right to make corrections, enhancements, improvements and other changes to its semiconductor products and services per JESD46, latest issue, and to discontinue any product or service per JESD48, latest issue. Buyers should obtain the latest relevant information before placing orders and should verify that such information is current and complete. All semiconductor products (also referred to herein as "components") are sold subject to TI's terms and conditions of sale supplied at the time of order acknowledgment.

TI warrants performance of its components to the specifications applicable at the time of sale, in accordance with the warranty in TI's terms and conditions of sale of semiconductor products. Testing and other quality control techniques are used to the extent TI deems necessary to support this warranty. Except where mandated by applicable law, testing of all parameters of each component is not necessarily performed.

TI assumes no liability for applications assistance or the design of Buyers' products. Buyers are responsible for their products and applications using TI components. To minimize the risks associated with Buyers' products and applications, Buyers should provide adequate design and operating safeguards.

TI does not warrant or represent that any license, either express or implied, is granted under any patent right, copyright, mask work right, or other intellectual property right relating to any combination, machine, or process in which TI components or services are used. Information published by TI regarding third-party products or services does not constitute a license to use such products or services or a warranty or endorsement thereof. Use of such information may require a license from a third party under the patents or other intellectual property of the third party, or a license from TI under the patents or other intellectual property of TI.

Reproduction of significant portions of TI information in TI data books or data sheets is permissible only if reproduction is without alteration and is accompanied by all associated warranties, conditions, limitations, and notices. TI is not responsible or liable for such altered documentation. Information of third parties may be subject to additional restrictions.

Resale of TI components or services with statements different from or beyond the parameters stated by TI for that component or service voids all express and any implied warranties for the associated TI component or service and is an unfair and deceptive business practice. TI is not responsible or liable for any such statements.

Buyer acknowledges and agrees that it is solely responsible for compliance with all legal, regulatory and safety-related requirements concerning its products, and any use of TI components in its applications, notwithstanding any applications-related information or support that may be provided by TI. Buyer represents and agrees that it has all the necessary expertise to create and implement safeguards which anticipate dangerous consequences of failures, monitor failures and their consequences, lessen the likelihood of failures that might cause harm and take appropriate remedial actions. Buyer will fully indemnify TI and its representatives against any damages arising out of the use of any TI components in safety-critical applications.

In some cases, TI components may be promoted specifically to facilitate safety-related applications. With such components, TI's goal is to help enable customers to design and create their own end-product solutions that meet applicable functional safety standards and requirements. Nonetheless, such components are subject to these terms.

No TI components are authorized for use in FDA Class III (or similar life-critical medical equipment) unless authorized officers of the parties have executed a special agreement specifically governing such use.

Only those TI components which TI has specifically designated as military grade or "enhanced plastic" are designed and intended for use in military/aerospace applications or environments. Buyer acknowledges and agrees that any military or aerospace use of TI components which have **not** been so designated is solely at the Buyer's risk, and that Buyer is solely responsible for compliance with all legal and regulatory requirements in connection with such use.

TI has specifically designated certain components as meeting ISO/TS16949 requirements, mainly for automotive use. In any case of use of non-designated products, TI will not be responsible for any failure to meet ISO/TS16949.

Products

| | |
|------------------------------|--|
| Audio | www.ti.com/audio |
| Amplifiers | amplifier.ti.com |
| Data Converters | dataconverter.ti.com |
| DLP® Products | www.dlp.com |
| DSP | dsp.ti.com |
| Clocks and Timers | www.ti.com/clocks |
| Interface | interface.ti.com |
| Logic | logic.ti.com |
| Power Mgmt | power.ti.com |
| Microcontrollers | microcontroller.ti.com |
| RFID | www.ti-rfid.com |
| OMAP Applications Processors | www.ti.com/omap |
| Wireless Connectivity | www.ti.com/wirelessconnectivity |

Applications

| | |
|-------------------------------|--|
| Automotive and Transportation | www.ti.com/automotive |
| Communications and Telecom | www.ti.com/communications |
| Computers and Peripherals | www.ti.com/computers |
| Consumer Electronics | www.ti.com/consumer-apps |
| Energy and Lighting | www.ti.com/energy |
| Industrial | www.ti.com/industrial |
| Medical | www.ti.com/medical |
| Security | www.ti.com/security |
| Space, Avionics and Defense | www.ti.com/space-avionics-defense |
| Video and Imaging | www.ti.com/video |

TI E2E Community

e2e.ti.com